

PIC16F87XA Data Sheet

28/40/44-Pin Enhanced Flash Microcontrollers

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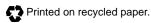
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28/40/44-Pin Enhanced Flash Microcontrollers

Devices Included in this Data Sheet:

- PIC16F873A
- PIC16F876A
- PIC16F874A PIC16F877A

High-Performance RISC CPU:

- Only 35 single-word instructions to learn
- All single-cycle instructions except for program branches, which are two-cycle
- Operating speed: DC 20 MHz clock input DC – 200 ns instruction cycle
- Up to 8K x 14 words of Flash Program Memory, Up to 368 x 8 bytes of Data Memory (RAM), Up to 256 x 8 bytes of EEPROM Data Memory
- Pinout compatible to other 28-pin or 40/44-pin PIC16CXXX and PIC16FXXX microcontrollers

Peripheral Features:

- Timer0: 8-bit timer/counter with 8-bit prescaler
- Timer1: 16-bit timer/counter with prescaler, can be incremented during Sleep via external crystal/clock
- Timer2: 8-bit timer/counter with 8-bit period register, prescaler and postscaler
- Two Capture, Compare, PWM modules
 - Capture is 16-bit, max. resolution is 12.5 ns
 - Compare is 16-bit, max. resolution is 200 ns
 - PWM max. resolution is 10-bit
- Synchronous Serial Port (SSP) with SPI™ (Master mode) and I²C[™] (Master/Slave)
- Universal Synchronous Asynchronous Receiver Transmitter (USART/SCI) with 9-bit address detection
- Parallel Slave Port (PSP) 8 bits wide with external RD, WR and CS controls (40/44-pin only)
- Brown-out detection circuitry for Brown-out Reset (BOR)

Analog Features:

- 10-bit, up to 8-channel Analog-to-Digital Converter (A/D)
- Brown-out Reset (BOR)
- Analog Comparator module with:
 - Two analog comparators
 - Programmable on-chip voltage reference (VREF) module
 - Programmable input multiplexing from device inputs and internal voltage reference
 - Comparator outputs are externally accessible

Special Microcontroller Features:

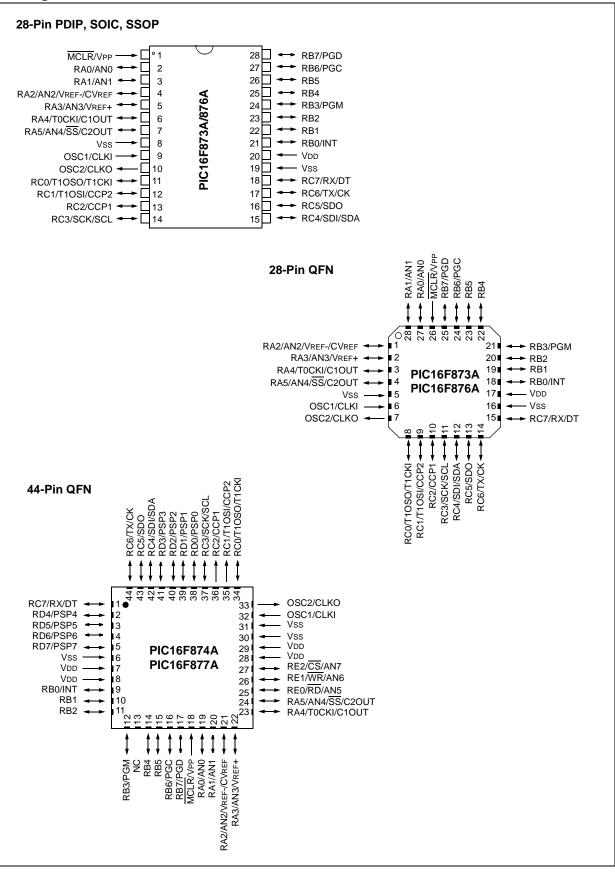
- 100,000 erase/write cycle Enhanced Flash
 program memory typical
- 1,000,000 erase/write cycle Data EEPROM memory typical
- Data EEPROM Retention > 40 years
- Self-reprogrammable under software control
- In-Circuit Serial Programming[™] (ICSP[™]) via two pins
- Single-supply 5V In-Circuit Serial Programming
- Watchdog Timer (WDT) with its own on-chip RC oscillator for reliable operation
- Programmable code protection
- Power saving Sleep mode
- · Selectable oscillator options
- · In-Circuit Debug (ICD) via two pins

CMOS Technology:

- Low-power, high-speed Flash/EEPROM technology
- Fully static design
- Wide operating voltage range (2.0V to 5.5V)
- Commercial and Industrial temperature ranges
- Low-power consumption

	Prog	ram Memory	Data	Data EEPROM		10 hit	10-bit CCP		ISSP		Timers	
Device	Bytes	# Single Word Instructions	SRAM (Bytes)	(Bytes)	1/()		(PWM)	SPI	Master I ² C	USART	8/16-bit	Comparators
PIC16F873A	7.2K	4096	192	128	22	5	2	Yes	Yes	Yes	2/1	2
PIC16F874A	7.2K	4096	192	128	33	8	2	Yes	Yes	Yes	2/1	2
PIC16F876A	14.3K	8192	368	256	22	5	2	Yes	Yes	Yes	2/1	2
PIC16F877A	14.3K	8192	368	256	33	8	2	Yes	Yes	Yes	2/1	2

Pin Diagrams



Pin Diagrams (Continued)

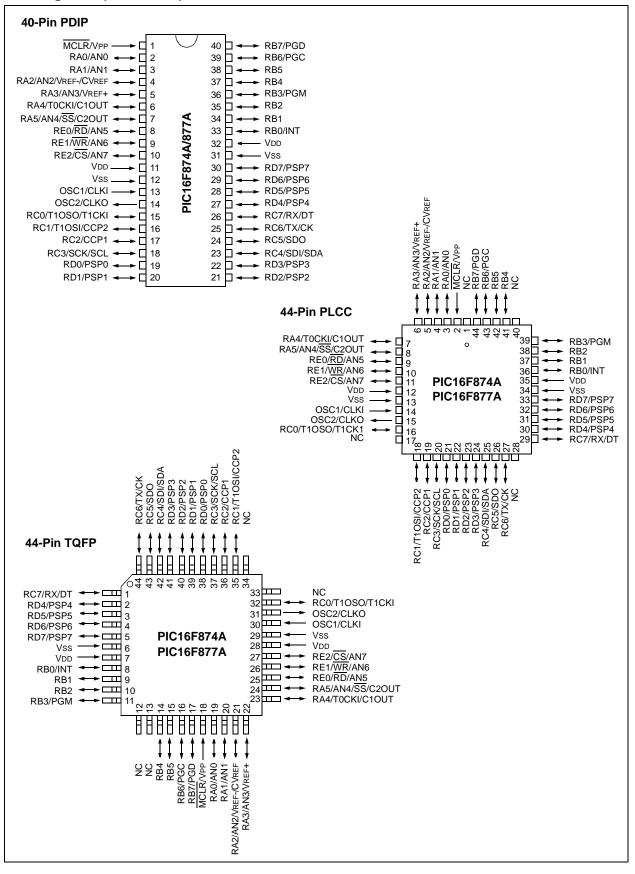


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1.0 DEVICE OVERVIEW

This document contains device specific information about the following devices:

- PIC16F873A
- PIC16F874A
- PIC16F876A
- PIC16F877A

PIC16F873A/876A devices are available only in 28-pin packages, while PIC16F874A/877A devices are available in 40-pin and 44-pin packages. All devices in the PIC16F87XA family share common architecture with the following differences:

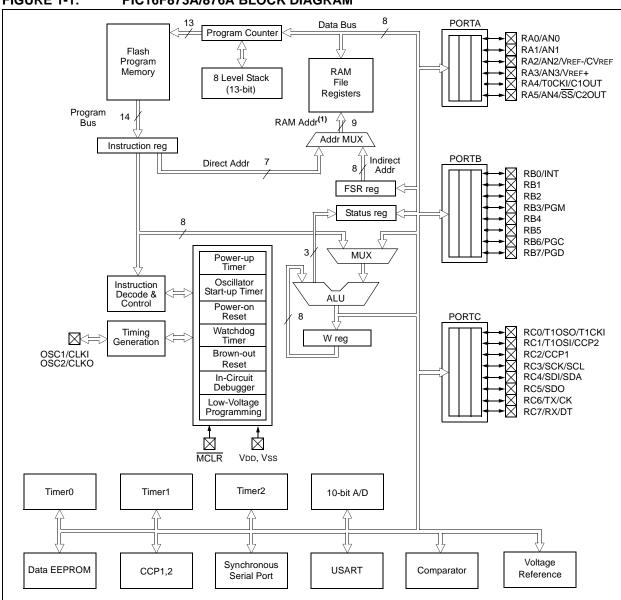
- The PIC16F873A and PIC16F874A have one-half of the total on-chip memory of the PIC16F876A and PIC16F877A
- The 28-pin devices have three I/O ports, while the 40/44-pin devices have five
- The 28-pin devices have fourteen interrupts, while the 40/44-pin devices have fifteen
- The 28-pin devices have five A/D input channels, while the 40/44-pin devices have eight
- The Parallel Slave Port is implemented only on the 40/44-pin devices

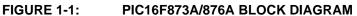
The available features are summarized in Table 1-1. Block diagrams of the PIC16F873A/876A and PIC16F874A/877A devices are provided in Figure 1-1 and Figure 1-2, respectively. The pinouts for these device families are listed in Table 1-2 and Table 1-3.

Additional information may be found in the PICmicro[®] Mid-Range Reference Manual (DS33023), which may be obtained from your local Microchip Sales Representative or downloaded from the Microchip web site. The Reference Manual should be considered a complementary document to this data sheet and is highly recommended reading for a better understanding of the device architecture and operation of the peripheral modules.

Key Features	PIC16F873A	PIC16F874A	PIC16F876A	PIC16F877A
Operating Frequency	DC – 20 MHz			
Resets (and Delays)	POR, BOR (PWRT, OST)	POR, BOR (PWRT, OST)	POR, BOR (PWRT, OST)	POR, BOR (PWRT, OST)
Flash Program Memory (14-bit words)	4K	4K	8K	8K
Data Memory (bytes)	192	192	368	368
EEPROM Data Memory (bytes)	128	128	256	256
Interrupts	14	15	14	15
I/O Ports	Ports A, B, C	Ports A, B, C, D, E	Ports A, B, C	Ports A, B, C, D, E
Timers	3	3	3	3
Capture/Compare/PWM modules	2	2	2	2
Serial Communications	MSSP, USART	MSSP, USART	MSSP, USART	MSSP, USART
Parallel Communications	—	PSP	—	PSP
10-bit Analog-to-Digital Module	5 input channels	8 input channels	5 input channels	8 input channels
Analog Comparators	2	2	2	2
Instruction Set	35 Instructions	35 Instructions	35 Instructions	35 Instructions
Packages	28-pin PDIP	40-pin PDIP	28-pin PDIP	40-pin PDIP
	28-pin SOIC	44-pin PLCC	28-pin SOIC	44-pin PLCC
	28-pin SSOP	44-pin TQFP	28-pin SSOP	44-pin TQFP
	28-pin QFN	44-pin QFN	28-pin QFN	44-pin QFN

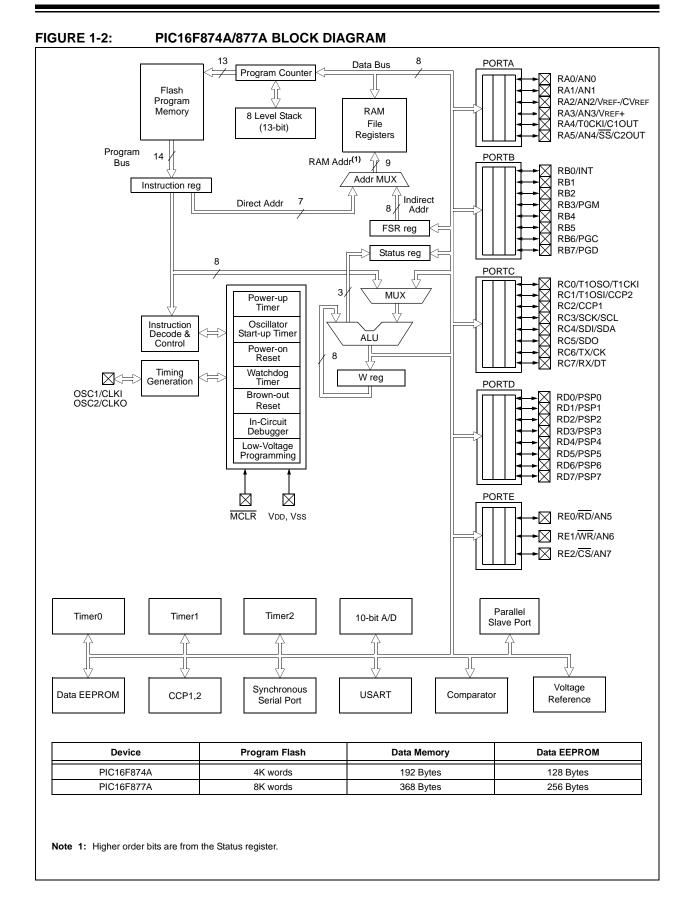
TABLE 1-1:PIC16F87XA DEVICE FEATURES





Device	Program Flash	Data Memory	Data EEPROM		
PIC16F873A	4K words	192 Bytes	128 Bytes		
PIC16F876A	8K words	368 Bytes	256 Bytes		

Note 1: Higher order bits are from the Status register.



Pin Name	PDIP, SOIC, SSOP Pin#	QFN Pin#	I/O/P Type	Buffer Type	Description
OSC1/CLKI OSC1 CLKI	9	6	I	ST/CMOS ⁽³⁾	Oscillator crystal or external clock input. Oscillator crystal input or external clock source input. ST buffer when configured in RC mode; otherwise CMOS. External clock source input. Always associated with pin
					function OSC1 (see OSC1/CLKI, OSC2/CLKO pins).
OSC2/CLKO OSC2	10	7	Ο	-	Oscillator crystal or clock output. Oscillator crystal output. Connects to crystal or resonator in Crystal Oscillator mode.
CLKO			0		In RC mode, OSC2 pin outputs CLKO, which has 1/4 the frequency of OSC1 and denotes the instruction cycle rate.
MCLR/VPP MCLR	1	26	I	ST	Master Clear (input) or programming voltage (output). Master Clear (Reset) input. This pin is an active low Reset to the device.
VPP			Р		Programming voltage input.
					PORTA is a bidirectional I/O port.
RA0/AN0 RA0 AN0	2	27	I/O I	TTL	Digital I/O. Analog input 0.
RA1/AN1 RA1 AN1	3	28	I/O I	TTL	Digital I/O. Analog input 1.
RA2/AN2/VREF-/ CVREF RA2 AN2 VREF- CVREF	4	1	I/O I I O	TTL	Digital I/O. Analog input 2. A/D reference voltage (Low) input. Comparator VREF output.
RA3/AN3/VREF+ RA3 AN3 VREF+	5	2	I/O I I	TTL	Digital I/O. Analog input 3. A/D reference voltage (High) input.
RA4/T0CKI/C1OUT RA4 T0CKI C1OUT	6	3	I/O I O	ST	Digital I/O – Open-drain when configured as output. Timer0 external clock input. Comparator 1 output.
RA5/AN4/SS/C2OUT RA5 AN4 SS C2OUT	7	4	I/O I I O	TTL	Digital I/O. Analog input 4. SPI slave select input. Comparator 2 output.

TABLE 1-2:PIC16F873A/876A PINOUT DESCRIPTION

- = Not used TTL = TTL input ST = Schmitt Trigger input

Note 1: This buffer is a Schmitt Trigger input when configured as the external interrupt.

2: This buffer is a Schmitt Trigger input when used in Serial Programming mode.

Pin Name	PDIP, SOIC, SSOP Pin#	QFN Pin#	I/O/P Type	Buffer Type	Description
					PORTB is a bidirectional I/O port. PORTB can be software
					programmed for internal weak pull-ups on all inputs.
RB0/INT	21	18		TTL/ST ⁽¹⁾	
RB0			I/O		Digital I/O.
INT			I		External interrupt.
RB1	22	19	I/O	TTL	Digital I/O.
RB2	23	20	I/O	TTL	Digital I/O.
RB3/PGM	24	21		TTL	
RB3			I/O		Digital I/O.
PGM			I		Low-voltage (single-supply) ICSP programming enable pir
RB4	25	22	I/O	TTL	Digital I/O.
RB5	26	23	I/O	TTL	Digital I/O.
-	27	24	., 0	TTL/ST ⁽²⁾	Digital I/O.
RB6/PGC RB6	21	24	I/O	111/31	Digital I/O.
PGC			1		In-circuit debugger and ICSP programming clock.
RB7/PGD	28	25		TTL/ST ⁽²⁾	
RB7	20	25	I/O	111/31	Digital I/O.
PGD			I/O		In-circuit debugger and ICSP programming data.
					PORTC is a bidirectional I/O port.
	11	8		ST	
RC0/T1OSO/T1CKI RC0		0	I/O	31	Digital I/O.
T1OSO			0		Timer1 oscillator output.
T1CKI			Ĩ		Timer1 external clock input.
RC1/T1OSI/CCP2	12	9		ST	
RC1	.=	Ũ	I/O	0.	Digital I/O.
T1OSI			I		Timer1 oscillator input.
CCP2			I/O		Capture2 input, Compare2 output, PWM2 output.
RC2/CCP1	13	10		ST	
RC2			I/O		Digital I/O.
CCP1			I/O		Capture1 input, Compare1 output, PWM1 output.
RC3/SCK/SCL	14	11		ST	
RC3			I/O		Digital I/O.
SCK			I/O		Synchronous serial clock input/output for SPI mode.
SCL			I/O		Synchronous serial clock input/output for I ² C mode.
RC4/SDI/SDA	15	12		ST	
RC4			I/O		Digital I/O.
SDI					SPI data in.
SDA	10	40	I/O	07	l ² C data I/O.
RC5/SDO RC5	16	13	I/O	ST	
SDO			0		Digital I/O. SPI data out.
RC6/TX/CK	17	14	Ŭ	ST	
RC6	17	14	I/O	51	Digital I/O.
TX			0		USART asynchronous transmit.
CK			1/0		USART1 synchronous clock.
RC7/RX/DT	18	15		ST	
RC7	-	-	I/O		Digital I/O.
RX			I		USART asynchronous receive.
DT			I/O		USART synchronous data.
Vss	8, 19	5, 6	Р	—	Ground reference for logic and I/O pins.
		17	Р	1	Positive supply for logic and I/O pins.
Vdd	20	17			

TABLE 1-2:	PIC16F873A/876A PINOUT DESCRIPTION (CONTINUED)	ł
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2: This buffer is a Schmitt Trigger input when used in Serial Programming mode.

Pin Name	PDIP Pin#	PLCC Pin#	TQFP Pin#	QFN Pin#	I/O/P Type	Buffer Type	Description
OSC1/CLKI OSC1	13	14	30	32	1	ST/CMOS ⁽⁴⁾	Oscillator crystal or external clock input. Oscillator crystal input or external clock source
					'		input. ST buffer when configured in RC mode; otherwise CMOS.
CLKI					I		External clock source input. Always associated with pin function OSC1 (see OSC1/CLKI, OSC2/CLKO pins).
OSC2/CLKO OSC2	14	15	31	33	0	_	Oscillator crystal or clock output. Oscillator crystal output. Connects to crystal or resonator in Crystal
CLKO					0		Oscillator mode. In RC mode, OSC2 pin outputs CLKO, which has 1/4 the frequency of OSC1 and denotes the instruction cycle rate.
MCLR/VPP MCLR	1	2	18	18	I	ST	Master Clear (input) or programming voltage (output Master Clear (Reset) input. This pin is an active low Reset to the device.
Vpp					Р		Programming voltage input.
							PORTA is a bidirectional I/O port.
RA0/AN0 RA0 AN0	2	3	19	19	I/O I	TTL	Digital I/O. Analog input 0.
RA1/AN1 RA1 AN1	3	4	20	20	I/O I	TTL	Digital I/O. Analog input 1.
RA2/AN2/VREF-/CVREF RA2 AN2 VREF-	4	5	21	21	I/O I	TTL	Digital I/O. Analog input 2.
CVREF					0		A/D reference voltage (Low) input. Comparator VREF output.
RA3/AN3/VREF+ RA3	5	6	22	22	I/O	TTL	Digital I/O.
AN3 VREF+					1 		Analog input 3. A/D reference voltage (High) input.
RA4/T0CKI/C1OUT RA4	6	7	23	23	I/O	ST	Digital I/O – Open-drain when configured as
T0CKI C1OUT					I O		output. Timer0 external clock input. Comparator 1 output.
RA5/AN4/SS/C2OUT	7	8	24	24		TTL	
RA5 <u>AN</u> 4 SS					I/O I I		Digital I/O. Analog input 4. SPI slave select input.
C2OUT					0		Comparator 2 output.

TABLE 1-3: PIC16F874A/877A PINOUT DESCRIPTION

— = Not used TTL = TTL input ST = Schmitt Trigger input

Note 1: This buffer is a Schmitt Trigger input when configured as the external interrupt.

2: This buffer is a Schmitt Trigger input when used in Serial Programming mode.

Pin Name	PDIP Pin#	PLCC Pin#	TQFP Pin#	QFN Pin#	I/O/P Type	Buffer Type	Description
							PORTB is a bidirectional I/O port. PORTB can be
							software programmed for internal weak pull-up on all
							inputs.
RB0/INT	33	36	8	9		TTL/ST ⁽¹⁾	
RB0					I/O		Digital I/O.
INT					I		External interrupt.
RB1	34	37	9	10	I/O	TTL	Digital I/O.
RB2	35	38	10	11	I/O	TTL	Digital I/O.
RB3/PGM	36	39	11	12		TTL	
RB3					I/O		Digital I/O.
PGM					I		Low-voltage ICSP programming enable pin.
RB4	37	41	14	14	I/O	TTL	Digital I/O.
RB5	38	42	15	15	I/O	TTL	Digital I/O.
RB6/PGC	39	43	16	16		TTL/ST ⁽²⁾	
RB6					I/O		Digital I/O.
PGC					I		In-circuit debugger and ICSP programming clock.
RB7/PGD	40	44	17	17		TTL/ST ⁽²⁾	
RB7					I/O		Digital I/O.
PGD					I/O		In-circuit debugger and ICSP programming data.

TABLE 1-3: PIC16F874A/877A PINOUT DESCRIPTION (CONTINUED)

— = Not used TTL = TTL input ST = Schmitt Trigger input

Note 1: This buffer is a Schmitt Trigger input when configured as the external interrupt.

2: This buffer is a Schmitt Trigger input when used in Serial Programming mode.

Pin Name	PDIP Pin#	PLCC Pin#	TQFP Pin#	QFN Pin#	I/O/P Type	Buffer Type	Description
							PORTC is a bidirectional I/O port.
RC0/T1OSO/T1CKI RC0 T1OSO T1CKI	15	16	32	34	I/O O I	ST	Digital I/O. Timer1 oscillator output. Timer1 external clock input.
RC1/T1OSI/CCP2 RC1 T1OSI CCP2	16	18	35	35	I/O I I/O	ST	Digital I/O. Timer1 oscillator input. Capture2 input, Compare2 output, PWM2 output.
RC2/CCP1 RC2 CCP1	17	19	36	36	I/O I/O	ST	Digital I/O. Capture1 input, Compare1 output, PWM1 output.
RC3/SCK/SCL RC3 SCK	18	20	37	37	I/O I/O	ST	Digital I/O. Synchronous serial clock input/output for SPI mode.
SCL					I/O		Synchronous serial clock input/output for I ² C mode.
RC4/SDI/SDA RC4 SDI SDA	23	25	42	42	I/O I I/O	ST	Digital I/O. SPI data in. I ² C data I/O.
RC5/SDO RC5 SDO	24	26	43	43	I/O O	ST	Digital I/O. SPI data out.
RC6/TX/CK RC6 TX CK	25	27	44	44	I/O O I/O	ST	Digital I/O. USART asynchronous transmit. USART1 synchronous clock.
RC7/RX/DT RC7 RX DT	26	29	1	1	I/O I I/O	ST	Digital I/O. USART asynchronous receive. USART synchronous data.

TABLE 1-3: PIC16F874A/877A PINOUT DESCRIPTION (CONTINUED)

= Not used TTL = TTL input ST = Schmitt Trigger input

Note 1: This buffer is a Schmitt Trigger input when configured as the external interrupt.

2: This buffer is a Schmitt Trigger input when used in Serial Programming mode.

TABLE 1-3:	PIC16F	374A/87			JESCR		ONTINUED)
Pin Name	PDI Pin		TQFP Pin#	QFN Pin#	I/O/P Type	Buffer Type	Description
	1116	· · · · · · · · · · · · · · · · · · ·	1 111#	r 111#	туре	туре	
							PORTD is a bidirectional I/O port or Parallel Slave Port when interfacing to a microprocessor bus.
	19	21	38	38		ST/TTL ⁽³⁾	For when interfacing to a microprocessor bus.
RD0/PSP0 RD0	19	21	30	30	I/O	51/11L*/	Digital I/O.
PSP0					1/O		Parallel Slave Port data.
RD1/PSP1	20	22	39	39		ST/TTL ⁽³⁾	
RD1	20	~~~	00	00	I/O	ONTE	Digital I/O.
PSP1					I/O		Parallel Slave Port data.
RD2/PSP2	21	23	40	40		ST/TTL ⁽³⁾	
RD2					I/O		Digital I/O.
PSP2					I/O		Parallel Slave Port data.
RD3/PSP3	22	24	41	41		ST/TTL ⁽³⁾	
RD3					I/O		Digital I/O.
PSP3					I/O		Parallel Slave Port data.
RD4/PSP4	27	30	2	2		ST/TTL ⁽³⁾	
RD4					I/O		Digital I/O.
PSP4					I/O	(0)	Parallel Slave Port data.
RD5/PSP5	28	31	3	3		ST/TTL ⁽³⁾	
RD5 PSP5					I/O I/O		Digital I/O. Parallel Slave Port data.
					1/0	o= /== (3)	Parallel Slave Port data.
RD6/PSP6 RD6	29	32	4	4	I/O	ST/TTL ⁽³⁾	Digital I/O.
PSP6					1/O		Parallel Slave Port data.
RD7/PSP7	30	33	5	5	1/0	ST/TTL ⁽³⁾	
RD7	30	- 33	5	5	I/O	31/112.7	Digital I/O.
PSP7					I/O		Parallel Slave Port data.
							PORTE is a bidirectional I/O port.
RE0/RD/AN5	8	9	25	25		ST/TTL ⁽³⁾	
RE0	0	5	25	20	I/O	01/11L	Digital I/O.
RD					I		Read control for Parallel Slave Port.
AN5					I		Analog input 5.
RE1/WR/AN6	9	10	26	26		ST/TTL ⁽³⁾	
RE1					I/O		Digital I/O.
WR					I		Write control for Parallel Slave Port.
AN6					I	(0)	Analog input 6.
RE2/CS/AN7	10	11	27	27		ST/TTL ⁽³⁾	
RE2 CS					I/O		Digital I/O. Chip select control for Parallel Slove Part
AN7							Chip select control for Parallel Slave Port. Analog input 7.
Vss	12, 3	1 13, 34	6, 29	6, 30,	P	_	Ground reference for logic and I/O pins.
	, -	.,		31			
Vdd	11, 3	2 12, 35	7, 28	7, 8, 28, 29	Р		Positive supply for logic and I/O pins.
NC	-	1, 17, 28, 40	12,13, 33, 34	13	—	—	These pins are not internally connected. These pins should be left unconnected.

PIC16F874A/877A PINOLIT DESCRIPTION (CONTINUED) **TABLE 1-3**

— = Not used TTL = TTL input

ST = Schmitt Trigger input

Note 1: This buffer is a Schmitt Trigger input when configured as the external interrupt.

2: This buffer is a Schmitt Trigger input when used in Serial Programming mode.

NOTES:

2.0 MEMORY ORGANIZATION

There are three memory blocks in each of the PIC16F87XA devices. The program memory and data memory have separate buses so that concurrent access can occur and is detailed in this section. The EEPROM data memory block is detailed in **Section 3.0 "Data EEPROM and Flash Program Memory"**.

Additional information on device memory may be found in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

FIGURE 2-1:

PIC16F876A/877A

PROGRAM MEMORY MAP

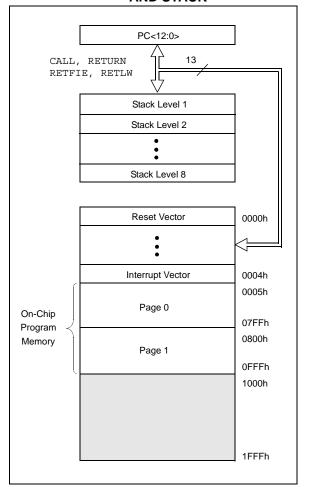
AND STACK PC<12:0> CALL, RETURN 13 RETFIE, RETLW Stack Level 1 Stack Level 2 Stack Level 8 Reset Vector 0000h Interrupt Vector 0004h 0005h Page 0 07FFh 0800h Page 1 On-Chip 0FFFh Program 1000h Memory Page 2 17FFh 1800h Page 3 1FFFh

2.1 **Program Memory Organization**

The PIC16F87XA devices have a 13-bit program counter capable of addressing an 8K word x 14 bit program memory space. The PIC16F876A/877A devices have 8K words x 14 bits of Flash program memory, while PIC16F873A/874A devices have 4K words x 14 bits. Accessing a location above the physically implemented address will cause a wraparound.

The Reset vector is at 0000h and the interrupt vector is at 0004h.

FIGURE 2-2: PIC16F873A/874A PROGRAM MEMORY MAP AND STACK



2.2 Data Memory Organization

The data memory is partitioned into multiple banks which contain the General Purpose Registers and the Special Function Registers. Bits RP1 (Status<6>) and RP0 (Status<5>) are the bank select bits.

RP1:RP0	Bank
00	0
01	1
10	2
11	3

Each bank extends up to 7Fh (128 bytes). The lower locations of each bank are reserved for the Special Function Registers. Above the Special Function Registers are General Purpose Registers, implemented as static RAM. All implemented banks contain Special Function Registers. Some frequently used Special Function Registers from one bank may be mirrored in another bank for code reduction and quicker access.

Note:	The EEPROM data memory description can
	be found in Section 3.0 "Data EEPROM
	and Flash Program Memory" of this data
	sheet.

2.2.1 GENERAL PURPOSE REGISTER FILE

The register file can be accessed either directly, or indirectly, through the File Select Register (FSR).

```
FIGURE 2-3:
```

PIC16F876A/877A REGISTER FILE MAP

A	File ddress	A	File Address		File Address		File Addro
Indirect addr.(*)	00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)	180
TMR0	01h	OPTION_REG	81h	TMR0	101h	OPTION_REG	18
PCL	02h	PCL	82h	PCL	102h	PCL	182
STATUS	03h	STATUS	83h	STATUS	103h	STATUS	183
FSR	04h	FSR	84h	FSR	104h	FSR	184
PORTA	05h	TRISA	85h		105h		185
PORTB	06h	TRISB	86h	PORTB	106h	TRISB	186
PORTC	07h	TRISC	87h		107h		187
PORTD ⁽¹⁾	08h	TRISD ⁽¹⁾	88h		108h		188
PORTE ⁽¹⁾	09h	TRISE ⁽¹⁾	89h		109h		189
PCLATH	0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18/
INTCON	0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18
PIR1	0Ch	PIE1	8Ch	EEDATA	10Ch	EECON1	180
PIR2	0Dh	PIE2	8Dh	EEADR	10Dh	EECON2	18[
TMR1L	0Eh	PCON	8Eh	EEDATH	10Eh	Reserved ⁽²⁾	18
TMR1H	0Fh		8Fh	EEADRH	10Fh	Reserved ⁽²⁾	18
T1CON	10h		90h		110h		190
TMR2	11h	SSPCON2	91h		111h		19 ⁻
T2CON	12h	PR2	92h		112h		192
SSPBUF	13h	SSPADD	93h		113h		193
SSPCON	14h	SSPSTAT	94h		114h		194
CCPR1L	15h		95h		115h		195
CCPR1H	16h		96h		116h		196
CCP1CON	17h		97h	General	117h	General	197
RCSTA	18h	TXSTA	98h	Purpose Register	118h	Purpose Register	198
TXREG	19h	SPBRG	99h	16 Bytes	119h	16 Bytes	199
RCREG	1Ah		9Ah		11Ah		19/
CCPR2L	1Bh		9Bh		11Bh		19
CCPR2H	1Ch	CMCON	9Ch		11Ch		190
CCP2CON	1Dh	CVRCON	9Dh		11Dh		19[
ADRESH	1Eh	ADRESL	9Eh		11Eh		19
ADCON0	1Fh	ADCON1	9Fh		11Fh		19
	20h		A0h		120h		1A
		General		General		General	
General		Purpose Register		Purpose		Purpose Register	
Purpose		-		Register		°,	
Register		80 Bytes		80 Bytes		80 Bytes	
96 Bytes			EFh		16Fh		1E
		accesses	F0h	accesses	170h	accesses	1F(
		70h-7Fh		70h-7Fh		70h - 7Fh	
Bank 0	7Fh	Bank 1	FFh	Bank 2	17Fh	Bank 3	1FI
* Not a ph	iysical reg	ata memory location ister. e not implemented					



PIC16F873A/874A REGISTER FILE MAP

	A	File ddress		File Address		File Addres
00h	Indirect addr.(*)	80h	Indirect addr.(*)	100h	Indirect addr.(*)	180h
01h	OPTION_REG	81h	TMR0	101h	OPTION_REG	181h
02h	PCL	82h	PCL	102h	PCL	182h
03h	STATUS	83h	STATUS	103h	STATUS	183h
04h	FSR	84h	FSR	104h	FSR	184h
05h	TRISA	85h		105h		185h
06h	TRISB	86h	PORTB	106h	TRISB	186h
07h	TRISC	87h				187h
08h		88h				188h
09h	TRISE ⁽¹⁾	89h		109h		189h
0Ah	PCLATH	8Ah	PCLATH	10Ah	PCLATH	18Ał
0Bh	INTCON	8Bh	INTCON	10Bh	INTCON	18Bł
0Ch	PIE1	8Ch	EEDATA	10Ch	EECON1	18Cł
0Dh	PIE2	8Dh	EEADR	10Dh	EECON2	18DI
0Eh	PCON	8Eh	EEDATH	10Eh	Reserved ⁽²⁾	18Eł
0Fh		8Fh	EEADRH	10Fh	Reserved ⁽²⁾	18Fł
10h		90h		110h		190h
11h	SSPCON2	91h				
12h	PR2	92h				
13h	SSPADD	93h				
14h	SSPSTAT	94h				
15h		95h				
16h		96h				
17h		97h				
18h	TXSTA	98h				
19h	SPBRG	99h				
1Ah		9Ah				
1Bh		9Bh				
1Ch	CMCON	9Ch				
1Dh	CVRCON	9Dh				
1Eh	ADRESL	9Eh				
1Fh	ADCON1	9Fh		100		4 4 04
20h		A0h		120h		1A0h
	General Purpose Register		accesses 20h-7Fh		accesses A0h - FFh	
	96 Bytes			16Fh 170h		1EFh 1F0h
7Fh		FFh		17Fh		1FFh
	Bank 1		Bank 2		Bank 3	
	02h 03h 04h 05h 06h 07h 08h 02h 02h 02h 02h 02h 02h 02h 02h 02h 02	D2h PCL D3h STATUS D4h FSR D5h TRISA D6h TRISB D7h TRISC D8h TRISD(1) D9h TRISC(1) D8h TRISC(1) D9h TRISC(1) D8h TRISC(1) D9h TRISC(1) D0h PCLATH D8h INTCON DCh PIE1 D0h PIE2 D6h PCON D7h Registant 10h SSPCON2 12h PR2 13h SSPADD 14h SSPSTAT 15h 16h 17h 115h 18h TXSTA 19h SPBRG 1Ah 115h 16h 111 17h General 18h ADRESL 16h ADRESL 16h ADRESL	D2h PCL 82h D3h STATUS 83h D4h FSR 84h D5h TRISA 85h D6h TRISB 86h D7h TRISC 87h D8h TRISD(1) 88h D9h TRISE(1) 89h DAh PCLATH 8Ah D9h TRISE(1) 89h DAh PCLATH 8Ah D8h INTCON 8Bh DAh PCLATH 8Ah D9h PIE1 8Ch D0h PIE2 8Dh D6h PCON 8Eh D7h SSPCON2 91h 12h PR2 92h 13h SSPADD 93h 14h SSPSTAT 94h 15h 95h 16h 96h 17h 97h 18h TXSTA 19h SPBRG 99h Ah<	D2hPCL82hPCL03hSTATUS83hSTATUS03hSTATUS83hSTATUS04hFSR84hFSR05hTRISA85h006hTRISB86hPORTB07hTRISC87h008hTRISD(1)88h009hTRISE(1)89h000hPIE18ChEEDATA00hPIE18ChEEADR00hPIE28DhEEADR00hPIE28DhEEADR00hPIE28DhEEADR00hPR292h11hSSPCON291h12hPR292h13hSSPADD93h14hSSPSTAT94h15h95h16h96h17h97h18hTXSTA98h99h1Ah9Ah1Bh99h1ChCMCON9ChA0h1Bh96 Bytes20hA0hA0haccesses20h-7Fh96 BytesFFh	PCL 82h PCL 102h 03h STATUS 83h STATUS 103h 04h FSR 84h FSR 104h 05h TRISA 85h 105h 06h TRISA 85h 105h 07h TRISC 87h 107h 08h TRISC 87h 107h 08h TRISC 87h 109h 09h TRISE ⁽¹⁾ 89h 100h 09h PCLATH 8Ah PCLATH 00h PIE1 8Ch EEDATH 10Ch 0Dh PIE2 8Dh EEADR 10Dh 0Fh 8Fh EEADR 10Dh 10Dh 11h SSPCON2 91h 110h	D2h PCL 82h PCL 102h PCL 03h STATUS 83h STATUS 103h STATUS 103h 03h STATUS 83h STATUS 103h FSR 104h FSR 05h TRISA 85h 104h FSR 104h FSR 05h TRISB 86h PORTB 107h Image: Construct Struct Stru

2.2.2 SPECIAL FUNCTION REGISTERS

The Special Function Registers are registers used by the CPU and peripheral modules for controlling the desired operation of the device. These registers are implemented as static RAM. A list of these registers is given in Table 2-1. The Special Function Registers can be classified into two sets: core (CPU) and peripheral. Those registers associated with the core functions are described in detail in this section. Those related to the operation of the peripheral features are described in detail in the peripheral features section.

TABLE 2-1:SPECIAL FUNCTION REGISTER SUMMARY

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Details on page:
Bank 0											
00h ⁽³⁾	INDF	Addressing	g this locatio	n uses cont	ents of FSR t	o address da	ata memory (not a physic	cal register)	0000 0000	31, 150
01h	TMR0	Timer0 Mo	dule Registe	er						XXXX XXXX	55, 150
02h ⁽³⁾	PCL	Program C	ounter (PC)	Least Sign	ificant Byte					0000 0000	30, 150
03h ⁽³⁾	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	22, 150
04h ⁽³⁾	FSR	Indirect Da	ta Memory		xxxx xxxx	31, 150					
05h	PORTA	_	_	PORTA Da	ta Latch whe	en written: PO	ORTA pins w	hen read		0x 0000	43, 150
06h	PORTB	PORTB Da	ata Latch wh		xxxx xxxx	45, 150					
07h	PORTC	PORTC Da	ata Latch wh		xxxx xxxx	47, 150					
08h ⁽⁴⁾	PORTD	PORTD Da	ORTD Data Latch when written: PORTD pins when read								48, 150
09h ⁽⁴⁾	PORTE	_	_	_	_	_	RE2	RE1	RE0	xxx	49, 150
0Ah ^(1,3)	PCLATH	_	_	_	Write Buffer	for the uppe	er 5 bits of the	Program C	Counter	0 0000	30, 150
0Bh ⁽³⁾	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	24, 150
0Ch	PIR1	PSPIF ⁽³⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	26, 150
0Dh	PIR2	_	CMIF	—	EEIF	BCLIF	—	_	CCP2IF	-0-0 00	28, 150
0Eh	TMR1L	Holding Re	egister for th	e Least Sig	nificant Byte	of the 16-bit	TMR1 Regis	ter		xxxx xxxx	60, 150
0Fh	TMR1H	Holding Re	egister for th	e Most Sigr	nificant Byte c	of the 16-bit	TMR1 Regist	er		XXXX XXXX	60, 150
10h	T1CON	_		T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N	00 0000	57, 150
11h	TMR2	Timer2 Mo	dule Registe	er						0000 0000	62, 150
12h	T2CON		TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 0000	61, 150
13h	SSPBUF	Synchrono	us Serial Po	ort Receive	Buffer/Transr	nit Register				xxxx xxxx	79, 150
14h	SSPCON	WCOL	SSPOV	SSPEN	СКР	SSPM3	SSPM2	SSPM1	SSPM0	0000 0000	82, 82, 150
15h	CCPR1L	Capture/Co	ompare/PW	M Register	1 (LSB)					xxxx xxxx	63, 150
16h	CCPR1H	Capture/Co	ompare/PW	M Register	1 (MSB)					xxxx xxxx	63, 150
17h	CCP1CON	_	_	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00 0000	64, 150
18h	RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	112, 150
19h	TXREG	USART Tra	ansmit Data	Register						0000 0000	118, 150
1Ah	RCREG	USART Re	eceive Data	Register						0000 0000	118, 150
1Bh	CCPR2L	Capture/C	ompare/PW	M Register	2 (LSB)					xxxx xxxx	63, 150
1Ch	CCPR2H	Capture/C	ompare/PW	M Register	2 (MSB)					xxxx xxxx	63, 150
1Dh	CCP2CON	_		CCP2X	CCP2Y	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 0000	64, 150
1Eh	ADRESH	A/D Result	Register Hi	gh Byte						xxxx xxxx	133, 150
1Fh	ADCON0	ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/DONE	_	ADON	0000 00-0	127, 150

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8>, whose contents are transferred to the upper byte of the program counter.

2: Bits PSPIE and PSPIF are reserved on PIC16F873A/876A devices; always maintain these bits clear.

3: These registers can be addressed from any bank.

4: PORTD, PORTE, TRISD and TRISE are not implemented on PIC16F873A/876A devices, read as '0'.

5: Bit 4 of EEADRH implemented only on the PIC16F876A/877A devices.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Details on page:
Bank 1											
80h ⁽³⁾	INDF	Addressing	g this locatio	n uses cont	ents of FSR t	o address d	ata memory (not a physic	al register)	0000 0000	31, 150
81h	OPTION_REG	RBPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	23, 150
82h ⁽³⁾	PCL	Program C	Counter (PC)	Least Sign	ificant Byte					0000 0000	30, 150
83h ⁽³⁾	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	22, 150
84h ⁽³⁾	FSR	Indirect Da	ata Memory	Address Po	inter					xxxx xxxx	31, 150
85h	TRISA	_	_	PORTA Da	ta Direction F	Register				11 1111	43, 150
86h	TRISB	PORTB Da	ata Direction	Register						1111 1111	45, 150
87h	TRISC	PORTC Da	ata Directior	Register						1111 1111	47, 150
88h ⁽⁴⁾	TRISD	PORTD Da	ORTD Data Direction Register						1111 1111	48, 151	
89h ⁽⁴⁾	TRISE	IBF	OBF	IBOV	PSPMODE		PORTE Dat	a Direction I	oits	0000 -111	50, 151
8Ah ^(1,3)	PCLATH	—	—		Write Buffer	for the uppe	er 5 bits of the	Program C	ounter	0 0000	30, 150
8Bh ⁽³⁾	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	24, 150
8Ch	PIE1	PSPIE ⁽²⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	25, 151
8Dh	PIE2	_	CMIE		EEIE	BCLIE	—	_	CCP2IE	-0-0 00	27, 151
8Eh	PCON	—	—	_	—		—	POR	BOR	qq	29, 151
8Fh	—	Unimpleme	Unimplemented								
90h	—	Unimpleme	ented							—	_
91h	SSPCON2	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN	0000 0000	83, 151
92h	PR2		riod Registe							1111 1111	62, 151
93h	SSPADD	Synchrono	ous Serial Po	ort (I ² C mod	e) Address R	egister				0000 0000	79, 151
94h	SSPSTAT	SMP	CKE	D/A	Р	S	R/W	UA	BF	0000 0000	79, 151
95h	—	Unimpleme	ented							—	
96h	—	Unimpleme	ented							—	
97h	—	Unimpleme	ented							—	
98h	TXSTA	CSRC	TX9	TXEN	SYNC		BRGH	TRMT	TX9D	0000 -010	111, 151
99h	SPBRG	Baud Rate	Generator	Register						0000 0000	113, 151
9Ah	_	Unimpleme	ented							_	_
9Bh	_	Unimpleme	ented							—	_
9Ch	CMCON	C2OUT	C10UT	C2INV	C1INV	CIS	CM2	CM1	CM0	0000 0111	135, 151
9Dh	CVRCON	CVREN	CVROE	CVRR	—	CVR3	CVR2	CVR1	CVR0	000- 0000	141, 151
9Eh	ADRESL	A/D Result	t Register Lo	ow Byte						xxxx xxxx	133, 151
9Fh	ADCON1	ADFM	ADCS2	_	_	PCFG3	PCFG2	PCFG1	PCFG0	00 0000	128, 151

TABLE 2-1: SPECIAL FUNCTION REGISTER SUMMARY (CONTINUED)

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8>, whose contents are transferred to the upper byte of the program counter.

2: Bits PSPIE and PSPIF are reserved on PIC16F873A/876A devices; always maintain these bits clear.

3: These registers can be addressed from any bank.

4: PORTD, PORTE, TRISD and TRISE are not implemented on PIC16F873A/876A devices, read as '0'.

5: Bit 4 of EEADRH implemented only on the PIC16F876A/877A devices.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Details on page:
Bank 2											
100h ⁽³⁾	INDF	Addressing	g this locatio	n uses cont	ents of FSR t	to address da	ata memory (not a physic	al register)	0000 0000	31, 150
101h	TMR0	Timer0 Mo	dule Registe	ər						xxxx xxxx	55, 150
102h ⁽³⁾	PCL	Program C	Counter's (PC	C) Least Sig	nificant Byte					0000 0000	30, 150
103h ⁽³⁾	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	22, 150
104h ⁽³⁾	FSR	Indirect Da	ata Memory	Address Po	inter	•			•	xxxx xxxx	31, 150
105h	—	Unimplem	ented								_
106h	PORTB	PORTB Da	ata Latch wh	en written:	PORTB pins	when read				xxxx xxxx	45, 150
107h	_	Unimplem	ented		_						
108h	—	Unimplem	ented			_					
109h	_	Unimplem	ented		_						
10Ah ^(1,3)	PCLATH	_	— — Write Buffer for the upper 5 bits of the Program Counter								30, 150
10Bh ⁽³⁾	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	24, 150
10Ch	EEDATA	EEPROM	Data Regist		xxxx xxxx	39, 151					
10Dh	EEADR	EEPROM	EPROM Address Register Low Byte								
10Eh	EEDATH			EEPROM I	Data Registe	r High Byte				xx xxxx	39, 151
10Fh	EEADRH	_	_	_	(5)	EEPROM A	Address Regi	ster High By	/te	xxxx	39, 151
Bank 3											
180h ⁽³⁾	INDF	Addressing	g this locatio	n uses cont	ents of FSR I	to address da	ata memory (not a physic	al register)	0000 0000	31, 150
181h	OPTION_REG	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0	1111 1111	23, 150
182h ⁽³⁾	PCL	Program C	Counter (PC)	Least Sign	ificant Byte					0000 0000	30, 150
183h ⁽³⁾	STATUS	IRP	RP1	RP0	TO	PD	Z	DC	С	0001 1xxx	22, 150
184h ⁽³⁾	FSR	Indirect Da	ata Memory /	Address Po	inter					xxxx xxxx	31, 150
185h	_	Unimplem	ented								
186h	TRISB	PORTB Da	ata Direction	Register						1111 1111	45, 150
4071	_	Unimplem	ented	-							_
187h											
187h 188h	_	Unimplem	ented							—	—
		Unimplem Unimplem									_
188h 189h	— — PCLATH	•			Write Buffer	for the uppe	er 5 bits of the	Program C	Counter	 	— — 30, 150
188h 189h	 PCLATH INTCON	•		— TMR0IE	Write Buffer	for the uppe	er 5 bits of the	Program C	counter RBIF	 0000 000x	— — 30, 150 24, 150
188h 189h 18Ah ^(1,3)	-	Unimplem	ented	— TMR0IE —		· · ·					,
188h 189h 18Ah ^(1,3) 18Bh ⁽³⁾	INTCON	Unimplem — GIE EEPGD	ented — PEIE —			RBIE	TMR0IF	INTF	RBIF	0000 000x	24, 150
188h 189h 18Ah ^(1,3) 18Bh ⁽³⁾ 18Ch	INTCON EECON1	Unimplem — GIE EEPGD EEPROM	ented — PEIE —	— ister 2 (not a	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x x x000	24, 150 34, 151

TABLE 2-1:	SPECIAL FUNCTION REGISTER SUMMARY ((CONTINUED)

Legend: x = unknown, u = unchanged, q = value depends on condition, - = unimplemented, read as '0', r = reserved. Shaded locations are unimplemented, read as '0'.

Note 1: The upper byte of the program counter is not directly accessible. PCLATH is a holding register for the PC<12:8>, whose contents are transferred to the upper byte of the program counter.

2: Bits PSPIE and PSPIF are reserved on PIC16F873A/876A devices; always maintain these bits clear.

3: These registers can be addressed from any bank.

4: PORTD, PORTE, TRISD and TRISE are not implemented on PIC16F873A/876A devices, read as '0'.

5: Bit 4 of EEADRH implemented only on the PIC16F876A/877A devices.

2.2.2.1 Status Register

The Status register contains the arithmetic status of the ALU, the Reset status and the bank select bits for data memory.

The Status register can be the destination for any instruction, as with any other register. If the Status register is the destination for an instruction that affects the Z, DC or C bits, then the write to these three bits is disabled. These bits are set or cleared according to the device logic. Furthermore, the TO and PD bits are not writable, therefore, the result of an instruction with the Status register as destination may be different than intended.

For example, CLRF STATUS, will clear the upper three bits and set the Z bit. This leaves the Status register as 000u uluu (where u = unchanged).

It is recommended, therefore, that only BCF, BSF, SWAPF and MOVWF instructions are used to alter the Status register because these instructions do not affect the Z, C or DC bits from the Status register. For other instructions not affecting any status bits, see **Section 15.0 "Instruction Set Summary"**.

Note:	The C and DC bits operate as a borrow									
	and digit borrow bit, respectively, in sub-									
	traction. See the SUBLW and SUBWF									
	instructions for examples.									

REGISTER 2-1:	STATUS REGISTER (ADDRESS 03h, 83h, 103h, 183h)
---------------	--

	R/W-0	R/W-0	R/W-0	R-1	R-1	R/W-x	R/W-x	R/W-x					
	IRP	RP1	RP0	TO	PD	Z	DC	С					
	bit 7							bit 0					
bit 7	•	IRP: Register Bank Select bit (used for indirect addressing)											
	1 = Bank 2, 3 (100h-1FFh) 0 = Bank 0, 1 (00h-FFh)												
bit 6-5			<i>.</i>	ite (used for	direct addressi	na)							
bit 0-5	RP1:RP0 : Register Bank Select bits (used for direct addressing) 11 = Bank 3 (180h-1FFh)												
	11 = Bank 3 (100h-17Fh) 10 = Bank 2 (100h-17Fh)												
	01 = Bank 1 (80h-FFh)												
	00 = Bank 0 (00h-7Fh)												
bit 4		Each bank is 128 bytes. TO: Time-out bit											
	1 = After power-up, CLRWDT instruction or SLEEP instruction												
	•	r time-out oc											
bit 3	PD: Power-down bit												
	 1 = After power-up or by the CLRWDT instruction 0 = By execution of the SLEEP instruction 												
	-		SLEEP INS	truction									
bit 2	Z: Zero bit												
	 1 = The result of an arithmetic or logic operation is zero 0 = The result of an arithmetic or logic operation is not zero 												
bit 1				•		tructions)							
	DC : Digit carry/borrow bit (ADDWF, ADDLW, SUBLW, SUBWF instructions) (for borrow, the polarity is reversed)												
	 1 = A carry-out from the 4th low order bit of the result occurred 0 = No carry-out from the 4th low order bit of the result 												
bit 0	C: Carry/b	orrow bit (AD	DWF, ADDLI	W,SUBLW,S	UBWF instruction	ons)							
	•		•		f the result occu of the result occ								
	Note:	complemer	nt of the se	cond opera	d. A subtraction nd. For rotate (order bit of the	RRF, RLF)	instructions	•					
	Legend:]					
	R = Reada	able bit	VV = V	Vritable bit	U = Unimpl	emented b	it, read as '	0'					

'1' = Bit is set

- n = Value at POR

x = Bit is unknown

'0' = Bit is cleared

2.2.2.2 OPTION_REG Register

The OPTION_REG Register is a readable and writable register, which contains various control bits to configure the TMR0 prescaler/WDT postscaler (single assignable register known also as the prescaler), the external INT interrupt, TMR0 and the weak pull-ups on PORTB.

Note: To achieve a 1:1 prescaler assignment for the TMR0 register, assign the prescaler to the Watchdog Timer.

ER 2-2:	OPTION_REG REGISTER (ADDRESS 81h, 181h)											
	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1				
	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0				
	bit 7							bit 0				
bit 7	RBPU: PORTB Pull-up Enable bit											
		 1 = PORTB pull-ups are disabled 0 = PORTB pull-ups are enabled by individual port latch values 										
bit 6	INTEDG:	INTEDG: Interrupt Edge Select bit										
		pt on rising e pt on falling										
bit 5	T0CS : TM	R0 Clock So	ource Select	t bit								
	1 = Transition on RA4/T0CKI pin											
	0 = Internal instruction cycle clock (CLKO)											
bit 4		R0 Source E	-									
	 1 = Increment on high-to-low transition on RA4/T0CKI pin 0 = Increment on low-to-high transition on RA4/T0CKI pin 											
bit 3	PSA : Prescaler Assignment bit											
	1 = Prescaler is assigned to the WDT											
	0 = Prescaler is assigned to the Timer0 module											
bit 2-0	PS2:PS0: Prescaler Rate Select bits											
	Bit Value	TMR0 Rate	WDT Rate	Э								
	000	1:2	1:1									
	001 010	1:4 1:8	1:2									
	011	1:16	1:8									
	100 101	1:32	1 : 16 1 : 32									
	101	1:64 1:128	1:64									
	111	1 : 256	1 : 128									
	Legend:	legend:										
	R = Reada	able bit	W = V	Vritable bit	U = Unimpl	emented b	it, read as '	0'				
	- n = Value	e at POR	'1' = E	Bit is set	'0' = Bit is c		x = Bit is ur					
	Note:		3 in the TF	RISB registe	gramming (LVF r must be clear							

REGISTER 2-2: OPTION_REG REGISTER (ADDRESS 81h, 181h)

2.2.2.3 INTCON Register

The INTCON register is a readable and writable register, which contains various enable and flag bits for the TMR0 register overflow, RB port change and external RB0/INT pin interrupts. Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>). User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 2-3: INTCON REGISTER (ADDRESS 0Bh, 8Bh, 10Bh, 18Bh)

	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-x				
	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF				
	bit 7							bit 0				
bit 7	GIE: Globa	al Interrupt E	nable bit									
		s all unmas es all interru	ked interrupt pts	tS								
bit 6	PEIE: Perip	oheral Interr	upt Enable b	bit								
			ked peripher eral interrup		i							
bit 5	TMR0IE: T	MR0 Overfl	ow Interrupt	Enable bit								
	1 = Enables the TMR0 interrupt											
	0 = Disables the TMR0 interrupt											
bit 4	INTE: RB0/INT External Interrupt Enable bit											
	 1 = Enables the RB0/INT external interrupt 0 = Disables the RB0/INT external interrupt 											
	0 = Disable	es the RB0/I	NT external	interrupt								
bit 3	RBIE: RB I	Port Change	e Interrupt Ei	nable bit								
	1 = Enables the RB port change interrupt											
	0 = Disable	es the RB po	ort change ir	nterrupt								
bit 2			ow Interrupt	•								
		•	overflowed not overflow	•	eared in soft	ware)						
bit 1	INTF: RB0	INT Externation	al Interrupt F	lag bit								
			nal interrupt	•		red in softwa	are)					
bit 0	RBIF: RB F	Port Change	e Interrupt Fl	ag bit								
	the bit	 1 = At least one of the RB7:RB4 pins changed state; a mismatch condition will continue to set the bit. Reading PORTB will end the mismatch condition and allow the bit to be cleared (must be cleared in software). 										
	•		RB4 pins hav	ve changed	state							
	Legend:											
	R = Reada	ble bit	W = W	/ritable bit	U = Unim	nolemented	bit, read as	'O'				

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented I	bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

2.2.2.4 PIE1 Register

The PIE1 register contains the individual enable bits for the peripheral interrupts.

Note:	Bit PEIE (INTCON<6>) must be set to
	enable any peripheral interrupt.

REGISTER 2-4:	PIE1 REGI	STER (AD	DRESS 8	Ch)				
	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE
	bit 7							bit 0
					(4)			
bit 7					pt Enable bit ⁽¹⁾			
	1 = Enables 0 = Disable							
	Note 1:	PSPIE is re	eserved on	PIC16F873	A/876A devices	; always m	aintain this	bit clear.
bit 6	ADIE: A/D	Converter I	nterrupt Ena	able bit				
	1 = Enables							
	0 = Disable			•				
bit 5	RCIE: USA		-					
	1 = Enables 0 = Disable							
bit 4	TXIE: USAI	RT Transmi	t Interrupt E	nable bit				
	1 = Enables 0 = Disable							
bit 3					able bit			
	bit 3 SSPIE: Synchronous Serial Port Interrupt Enable bit 1 = Enables the SSP interrupt							
0 = Disables the SSP interrupt								
bit 2	CCP1IE: C	CP1 Interru	pt Enable b	it				
1 = Enables the CCP1 interrupt								
0 = Disables the CCP1 interrupt								
bit 1				errupt Enabl				
				tch interrupt atch interrup				
bit 0	TMR1IE: TI			•				
DILO	1 = Enables		•					
	0 = Disable			•				
				•				

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

2.2.2.5 PIR1 Register

The PIR1 register contains the individual flag bits for the peripheral interrupts.

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>). User software should ensure the appropriate interrupt bits are clear prior to enabling an interrupt.

REGISTER 2-5:	PIR1 REGISTER (ADDRESS 0Ch)									
	R/W-0	R/W-0	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-0		
	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF		
	bit 7							bit 0		
bit 7	 t 7 PSPIF: Parallel Slave Port Read/Write Interrupt Flag bit⁽¹⁾ 1 = A read or a write operation has taken place (must be cleared in software) 0 = No read or write has occurred Note 1: PSPIF is reserved on PIC16F873A/876A devices; always maintain this bit clear. 									
bit 6	ADIF: A/D Converter Interrupt Flag bit 1 = An A/D conversion completed 0 = The A/D conversion is not complete									
bit 5	RCIF : USA 1 = The US	RCIF : USART Receive Interrupt Flag bit 1 = The USART receive buffer is full 0 = The USART receive buffer is empty								
bit 4	TXIF : USAF 1 = The US	RT Transmit		ag bit empty						
bit 3 SSPIF: Synchronous Serial Port (SSP) Interrupt Flag bit 1 = The SSP interrupt condition has occurred and must be cleared in software before in from the Interrupt Service Routine. The conditions that will set this bit are: • SPI – A transmission/reception has taken place. • I ² C Slave – A transmission/reception has taken place. • I ² C Master • A transmission/reception has taken place. • The initiated Start condition was completed by the SSP module. • The initiated Restart condition was completed by the SSP module. • The initiated Restart condition was completed by the SSP module. • The initiated Restart condition was completed by the SSP module. • The initiated Acknowledge condition was completed by the SSP module. • A Start condition occurred while the SSP module was Idle (multi-master syster • A Start condition occurred while the SSP module was Idle (multi-master syster • A SSP interrupt rondition has occurred bit 2 CCP1IF: CCP1 Interrupt Flag bit <u>Capture mode:</u> 1 = A TMR1 register capture occurred (must be cleared in software) 0 = No TMR1 register compare match occurred (must be cleared in software) 0 = No TMR1 register compare match occurred <u>PWM mode:</u> Unused in this mode.							rstem).			
bit 1 TMR2IF : TMR2 to PR2 Match Interrupt Flag bit 1 = TMR2 to PR2 match occurred (must be cleared in software) 0 = No TMR2 to PR2 match occurred										
bit 0	bit 0 TMR1IF : TMR1 Overflow Interrupt Flag bit 1 = TMR1 register overflowed (must be cleared in software) 0 = TMR1 register did not overflow									
	Legend:									
	R = Readat			ritable bit		nplemented				
	- n = Value	at POR	'1' = Β	it is set	'0' = Bit i	s cleared	x = Bit is u	nknown		

2.2.2.6 PIE2 Register

The PIE2 register contains the individual enable bits for the CCP2 peripheral interrupt, the SSP bus collision interrupt, EEPROM write operation interrupt and the comparator interrupt.

Note:	Bit PEIE (INTCON<6>) must be set to
	enable any peripheral interrupt.

REGISTER 2-6:	PIE2 REG	ISTER (AD	DRESS 8	BDh)						
	U-0	R/W-0	U-0	R/W-0	R/W-0	U-0	U-0	R/W-0		
		CMIE	_	EEIE	BCLIE	—	—	CCP2IE		
	bit 7							bit 0		
bit 7	-	ented: Rea								
bit 6	CMIE: Con	nparator Inte	errupt Enab	le bit						
	 1 = Enables the comparator interrupt 0 = Disable the comparator interrupt 									
bit 5	Unimplem	ented: Rea	d as '0'							
bit 4	EEIE: EEP	ROM Write	Operation I	nterrupt Ena	able bit					
	 1 = Enable EEPROM write interrupt 0 = Disable EEPROM write interrupt 									
bit 3	BCLIE: Bus Collision Interrupt Enable bit									
	1 = Enable bus collision interrupt0 = Disable bus collision interrupt									
bit 2-1	Unimplemented: Read as '0'									
bit 0 CCP2IE: CCP2 Interrupt Enable bit 1 = Enables the CCP2 interrupt										
	0 = Disable	0 = Disables the CCP2 interrupt								
	Legend:]		
	B - Boodo	hla hit	M = M	Vritabla hit		omontod b	it road on	' O'		

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented	bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

2.2.2.7 PIR2 Register

The PIR2 register contains the flag bits for the CCP2 interrupt, the SSP bus collision interrupt, EEPROM write operation interrupt and the comparator interrupt.

Note: Interrupt flag bits are set when an interrupt condition occurs regardless of the state of its corresponding enable bit or the global enable bit, GIE (INTCON<7>). User software should ensure the appropriate interrupt flag bits are clear prior to enabling an interrupt.

REGISTER 2-7: PIR2 REGISTER (ADDRESS 0Dh)

U-0	R/W-0	U-0	R/W-0	R/W-0	U-0	U-0	R/W-0
—	CMIF		EEIF	BCLIF			CCP2IF
bit 7							bit 0

bit 7	Unimplemented: Read as '0'
bit 6	CMIF: Comparator Interrupt Flag bit
	 1 = The comparator input has changed (must be cleared in software) 0 = The comparator input has not changed
bit 5	Unimplemented: Read as '0'
bit 4	EEIF: EEPROM Write Operation Interrupt Flag bit
	 1 = The write operation completed (must be cleared in software) 0 = The write operation is not complete or has not been started
bit 3	BCLIF: Bus Collision Interrupt Flag bit
	 1 = A bus collision has occurred in the SSP when configured for I²C Master mode 0 = No bus collision has occurred
bit 2-1	Unimplemented: Read as '0'
bit 0	CCP2IF: CCP2 Interrupt Flag bit
	Capture mode:
	 1 = A TMR1 register capture occurred (must be cleared in software) 0 = No TMR1 register capture occurred
	Compare mode:
	 1 = A TMR1 register compare match occurred (must be cleared in software) 0 = No TMR1 register compare match occurred
	<u>PWM mode:</u> Unused.

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented	bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

2.2.2.8 PCON Register

bit 1

The Power Control (PCON) register contains flag bits to allow differentiation between a Power-on Reset (POR), a Brown-out Reset (BOR), a Watchdog Reset (WDT) and an external MCLR Reset.

Note: BOR is unknown on Power-on Reset. It must be set by the user and checked on subsequent Resets to see if BOR is clear, indicating a brown-out has occurred. The BOR status bit is a "don't care" and is not predictable if the brown-out circuit is disabled (by clearing the BODEN bit in the configuration word).

REGISTER 2-8: PCON REGISTER (ADDRESS 8Eh)

U-0	U-0	U-0	U-0	U-0	U-0	R/W-0	R/W-1
—	—	—	—	—	—	POR	BOR
bit 7							bit 0

- bit 7-2 Unimplemented: Read as '0'
 - POR: Power-on Reset Status bit
 - 1 = No Power-on Reset occurred

0 = A Power-on Reset occurred (must be set in software after a Power-on Reset occurs)

- bit 0 BOR: Brown-out Reset Status bit
 - 1 = No Brown-out Reset occurred

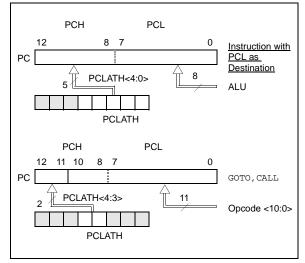
0 = A Brown-out Reset occurred (must be set in software after a Brown-out Reset occurs)

Legend:				
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'		
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown	

2.3 PCL and PCLATH

The Program Counter (PC) is 13 bits wide. The low byte comes from the PCL register which is a readable and writable register. The upper bits (PC<12:8>) are not readable, but are indirectly writable through the PCLATH register. On any Reset, the upper bits of the PC will be cleared. Figure 2-5 shows the two situations for the loading of the PC. The upper example in the figure shows how the PC is loaded on a write to PCL (PCLATH<4:0> \rightarrow PCH). The lower example in the figure shows how the PC is loaded during a CALL or GOTO instruction (PCLATH<4:3> \rightarrow PCH).

FIGURE 2-5: LOADING OF PC IN DIFFERENT SITUATIONS



2.3.1 COMPUTED GOTO

A computed GOTO is accomplished by adding an offset to the program counter (ADDWF PCL). When doing a table read using a computed GOTO method, care should be exercised if the table location crosses a PCL memory boundary (each 256-byte block). Refer to the application note, *AN556, "Implementing a Table Read"* (DS00556).

2.3.2 STACK

The PIC16F87XA family has an 8-level deep x 13-bit wide hardware stack. The stack space is not part of either program or data space and the stack pointer is not readable or writable. The PC is PUSHed onto the stack when a CALL instruction is executed, or an interrupt causes a branch. The stack is POP'ed in the event of a RETURN, RETLW or a RETFIE instruction execution. PCLATH is not affected by a PUSH or POP operation.

The stack operates as a circular buffer. This means that after the stack has been PUSHed eight times, the ninth push overwrites the value that was stored from the first push. The tenth push overwrites the second push (and so on).

- **Note 1:** There are no status bits to indicate stack overflow or stack underflow conditions.
 - 2: There are no instructions/mnemonics called PUSH or POP. These are actions that occur from the execution of the CALL, RETURN, RETLW and RETFIE instructions or the vectoring to an interrupt address.

2.4 Program Memory Paging

All PIC16F87XA devices are capable of addressing a continuous 8K word block of program memory. The CALL and GOTO instructions provide only 11 bits of address to allow branching within any 2K program memory page. When doing a CALL or GOTO instruction, the upper 2 bits of the address are provided by PCLATH<4:3>. When doing a CALL or GOTO instruction, the user must ensure that the page select bits are programmed so that the desired program memory page is addressed. If a return from a CALL instruction (or interrupt) is executed, the entire 13-bit PC is popped off the stack. Therefore, manipulation of the PCLATH<4:3> bits is not required for the RETURN instructions (which POPs the address from the stack).

Note:	The contents of the PCLATH register are				
	unchanged after a RETURN or RETFIE				
	instruction is executed. The user must				
	rewrite the contents of the PCLATH regis-				
	ter for any subsequent subroutine calls or				
	GOTO instructions.				

Example 2-1 shows the calling of a subroutine in page 1 of the program memory. This example assumes that PCLATH is saved and restored by the Interrupt Service Routine (if interrupts are used).

EXAMPLE 2-1: CALL OF A SUBROUTINE IN PAGE 1 FROM PAGE 0

	ORG 0x500	
	BCF PCLATH,4	
	BSF PCLATH,3	;Select page 1 ;(800h-FFFh)
	CALL SUB1_P1	;Call subroutine in
	:	;page 1 (800h-FFFh)
	:	
	ORG 0x900	;page 1 (800h-FFFh)
SUB1_P1		
	:	;called subroutine
		;page 1 (800h-FFFh)
	:	
	RETURN	;return to
		;Call subroutine
		;in page 0
		;(000h-7FFh)

2.5 Indirect Addressing, INDF and FSR Registers

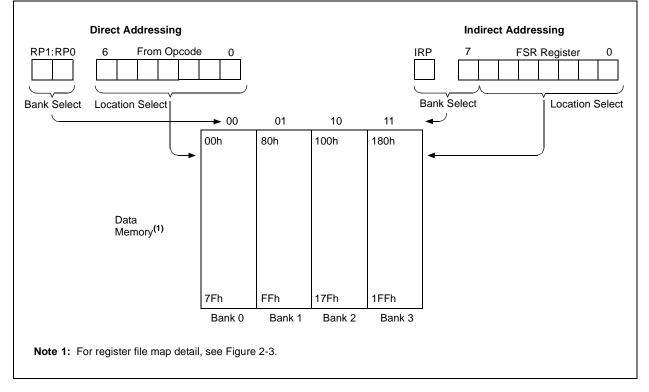
The INDF register is not a physical register. Addressing the INDF register will cause indirect addressing.

Indirect addressing is possible by using the INDF register. Any instruction using the INDF register actually accesses the register pointed to by the File Select Register, FSR. Reading the INDF register itself, indirectly (FSR = 0) will read 00h. Writing to the INDF register indirectly results in a no operation (although status bits may be affected). An effective 9-bit address is obtained by concatenating the 8-bit FSR register and the IRP bit (Status<7>) as shown in Figure 2-6. A simple program to clear RAM locations 20h-2Fh using indirect addressing is shown in Example 2-2.

EXAMPLE 2-2: INDIRECT ADDRESSING

	MOVLW	0x20	;initialize pointer
	MOVWF	FSR	;to RAM
NEXT	CLRF	INDF	clear INDF register;
	INCF	FSR,F	;inc pointer
	BTFSS	FSR,4	;all done?
	GOTO	NEXT	;no clear next
CONTINUE			
	:		;yes continue
			-





NOTES:

3.0 DATA EEPROM AND FLASH PROGRAM MEMORY

The data EEPROM and Flash program memory is readable and writable during normal operation (over the full VDD range). This memory is not directly mapped in the register file space. Instead, it is indirectly addressed through the Special Function Registers. There are six SFRs used to read and write this memory:

- EECON1
- EECON2
- EEDATA
- EEDATH
- EEADR
- EEADRH

When interfacing to the data memory block, EEDATA holds the 8-bit data for read/write and EEADR holds the address of the EEPROM location being accessed. These devices have 128 or 256 bytes of data EEPROM (depending on the device), with an address range from 00h to FFh. On devices with 128 bytes, addresses from 80h to FFh are unimplemented and will wraparound to the beginning of data EEPROM memory. When writing to unimplemented locations, the on-chip charge pump will be turned off.

When interfacing the program memory block, the EEDATA and EEDATH registers form a two-byte word that holds the 14-bit data for read/write and the EEADR and EEADRH registers form a two-byte word that holds the 13-bit address of the program memory location being accessed. These devices have 4 or 8K words of program Flash, with an address range from 0000h to 0FFFh for the PIC16F873A/874A and 0000h to 1FFFh for the PIC16F876A/877A. Addresses above the range of the respective device will wraparound to the beginning of program memory.

The EEPROM data memory allows single-byte read and write. The Flash program memory allows single-word reads and four-word block writes. Program memory write operations automatically perform an erase-before-write on blocks of four words. A byte write in data EEPROM memory automatically erases the location and writes the new data (erase-before-write).

The write time is controlled by an on-chip timer. The write/erase voltages are generated by an on-chip charge pump, rated to operate over the voltage range of the device for byte or word operations.

When the device is code-protected, the CPU may continue to read and write the data EEPROM memory. Depending on the settings of the write-protect bits, the device may or may not be able to write certain blocks of the program memory; however, reads of the program memory are allowed. When code-protected, the device programmer can no longer access data or program memory; this does NOT inhibit internal reads or writes.

3.1 EEADR and EEADRH

The EEADRH:EEADR register pair can address up to a maximum of 256 bytes of data EEPROM or up to a maximum of 8K words of program EEPROM. When selecting a data address value, only the LSByte of the address is written to the EEADR register. When selecting a program address value, the MSByte of the address is written to the EEADRH register and the LSByte is written to the EEADR register.

If the device contains less memory than the full address reach of the address register pair, the Most Significant bits of the registers are not implemented. For example, if the device has 128 bytes of data EEPROM, the Most Significant bit of EEADR is not implemented on access to data EEPROM.

3.2 EECON1 and EECON2 Registers

EECON1 is the control register for memory accesses.

Control bit, EEPGD, determines if the access will be a program or data memory access. When clear, as it is when reset, any subsequent operations will operate on the data memory. When set, any subsequent operations will operate on the program memory.

Control bits, RD and WR, initiate read and write or erase, respectively. These bits cannot be cleared, only set, in software. They are cleared in hardware at completion of the read or write operation. The inability to clear the WR bit in software prevents the accidental, premature termination of a write operation.

The WREN bit, when set, will allow a write or erase operation. On power-up, the WREN bit is clear. The WRERR bit is set when a write (or erase) operation is interrupted by a MCLR or a WDT Time-out Reset during normal operation. In these situations, following Reset, the user can check the WRERR bit and rewrite the location. The data and address will be unchanged in the EEDATA and EEADR registers.

Interrupt flag bit, EEIF in the PIR2 register, is set when the write is complete. It must be cleared in software.

EECON2 is not a physical register. Reading EECON2 will read all '0's. The EECON2 register is used exclusively in the EEPROM write sequence.

Note: The self-programming mechanism for Flash program memory has been changed. On previous PIC16F87X devices, Flash programming was done in single-word erase/ write cycles. The newer PIC18F87XA devices use a four-word erase/write cycle. See Section 3.6 "Writing to Flash Program Memory" for more information.

REGISTER 3-1:	EECON1 F	REGISTER	(ADDRES	S 18Ch)				
	R/W-x	U-0	U-0	U-0	R/W-x	R/W-0	R/S-0	R/S-0
	EEPGD		_	_	WRERR	WREN	WR	RD
	bit 7							bit 0
bit 7	EEPGD : Program/Data EEPROM Select bit 1 = Accesses program memory 0 = Accesses data memory Reads '0' after a POR; this bit cannot be changed while a write operation is in progress.							
bit 6-4	Unimplemented: Read as '0'							
bit 3	WRERR: E							
	 1 = A write operation is prematurely terminated (any MCLR or any WDT Reset during normal operation) 0 = The write operation completed 							
bit 2	WREN: EEPROM Write Enable bit							
	1 = Allows write cycles0 = Inhibits write to the EEPROM							
bit 1	WR: Write Control bit							
	 1 = Initiates a write cycle. The bit is cleared by hardware once write is complete. The WR bit can only be set (not cleared) in software. 0 = Write cycle to the EEPROM is complete 							
bit 0	 RD: Read Control bit 1 = Initiates an EEPROM read; RD is cleared in hardware. The RD bit can only be set (not cleared) in software. 0 = Does not initiate an EEPROM read 							
	Legend:							

Legend:					
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'			
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown		

3.3 Reading Data EEPROM Memory

To read a data memory location, the user must write the address to the EEADR register, clear the EEPGD control bit (EECON1<7>) and then set control bit RD (EECON1<0>). The data is available in the very next cycle in the EEDATA register; therefore, it can be read in the next instruction (see Example 3-1). EEDATA will hold this value until another read or until it is written to by the user (during a write operation).

The steps to reading the EEPROM data memory are:

- 1. Write the address to EEADR. Make sure that the address is not larger than the memory size of the device.
- 2. Clear the EEPGD bit to point to EEPROM data memory.
- 3. Set the RD bit to start the read operation.
- 4. Read the data from the EEDATA register.

EXAMPLE 3-1: DATA EEPROM READ

-			-
BSF	STATUS, RP1	;	
BCF	STATUS, RPO	;	Bank 2
MOVF	DATA_EE_ADDR,W	;	Data Memory
MOVWF	EEADR	;	Address to read
BSF	STATUS, RPO	;	Bank 3
BCF	EECON1, EEPGD	;	Point to Data
		;	memory
BSF	EECON1,RD	;	EE Read
BCF	STATUS, RPO	;	Bank 2
MOVF	EEDATA,W	;	W = EEDATA

3.4 Writing to Data EEPROM Memory

To write an EEPROM data location, the user must first write the address to the EEADR register and the data to the EEDATA register. Then the user must follow a specific write sequence to initiate the write for each byte.

The write will not initiate if the write sequence is not exactly followed (write 55h to EECON2, write AAh to EECON2, then set WR bit) for each byte. We strongly recommend that interrupts be disabled during this code segment (see Example 3-2).

Additionally, the WREN bit in EECON1 must be set to enable write. This mechanism prevents accidental writes to data EEPROM due to errant (unexpected) code execution (i.e., lost programs). The user should keep the WREN bit clear at all times, except when updating EEPROM. The WREN bit is not cleared by hardware

After a write sequence has been initiated, clearing the WREN bit will not affect this write cycle. The WR bit will be inhibited from being set unless the WREN bit is set. At the completion of the write cycle, the WR bit is cleared in hardware and the EE Write Complete Interrupt Flag bit (EEIF) is set. The user can either enable this interrupt or poll this bit. EEIF must be cleared by software.

The steps to write to EEPROM data memory are:

- 1. If step 10 is not implemented, check the WR bit to see if a write is in progress.
- 2. Write the address to EEADR. Make sure that the address is not larger than the memory size of the device.
- 3. Write the 8-bit data value to be programmed in the EEDATA register.
- 4. Clear the EEPGD bit to point to EEPROM data memory.
- 5. Set the WREN bit to enable program operations.
- 6. Disable interrupts (if enabled).
- 7. Execute the special five instruction sequence:
 - Write 55h to EECON2 in two steps (first to W, then to EECON2)
 - Write AAh to EECON2 in two steps (first to W, then to EECON2)
 - · Set the WR bit
- 8. Enable interrupts (if using interrupts).
- 9. Clear the WREN bit to disable program operations.
- At the completion of the write cycle, the WR bit is cleared and the EEIF interrupt flag bit is set. (EEIF must be cleared by firmware.) If step 1 is not implemented, then firmware should check for EEIF to be set, or WR to clear, to indicate the end of the program cycle.

EXAMPLE 3-2: DATA EEPROM WRITE

	BSF	STATUS, RP1	i
	BSF	STATUS, RPO	
	BTFSC	EECON1,WR	;Wait for write
	GOTO	\$-1	;to complete
	BCF	STATUS, RPO	;Bank 2
	MOVF	DATA_EE_ADDR,W	;Data Memory
	MOVWF	EEADR	;Address to write
	MOVF	DATA_EE_DATA,W	;Data Memory Value
	MOVWF	EEDATA	;to write
	BSF	STATUS, RPO	;Bank 3
	BCF	EECON1, EEPGD	;Point to DATA
			;memory
	BSF	EECON1,WREN	;Enable writes
	BCF	INTCON,GIE	;Disable INTs.
	MOVLW		;
g g	MOVWF	EECON2 AAh EECON2	;Write 55h
Required Sequence	MOVLW	AAh	;
Seq	MOVWF	EECON2	;Write AAh
ш ()	BSF	EECON1,WR	;Set WR bit to
			;begin write
	BSF	INTCON,GIE	;Enable INTs.
	BCF	EECON1,WREN	;Disable writes

ſ

3.5 Reading Flash Program Memory

To read a program memory location, the user must write two bytes of the address to the EEADR and EEADRH registers, set the EEPGD control bit (EECON1<7>) and then set control bit RD (EECON1<0>). Once the read control bit is set, the program memory Flash controller will use the next two instruction cycles to read the data. This causes these two instructions immediately following the "BSF EECON1, RD" instruction to be ignored. The data is available in the very next cycle in the EEDATA and EEDATH registers; therefore, it can be read as two bytes in the following instructions. EEDATA and EEDATH registers will hold this value until another read or until it is written to by the user (during a write operation).

EXAMPLE 3-3:	FLASH PROGRAM READ

	BSF	STATUS, RP1	;	
	BCF	STATUS, RPO	; Bank 2	
	MOVLW	MS PROG EE ADDR	;	
	MOVWF	EEADRH	; MS Byte of Program Address to read	
	MOVLW	LS PROG EE ADDR		
	MOVWF	EEADR — —	; LS Byte of Program Address to read	
	BSF	STATUS, RPO	; Bank 3	
	BSF	EECON1, EEPGD	; Point to PROGRAM memory	
	BSF	EECON1, RD	; EE Read	
e e	DOF	EECONI, RD	; EE REAU	
Required Sequence ;				
nbe	NOP			
Se Re	NOP		; Any instructions here are ignored as program	
			; memory is read in second cycle after BSF EECON1,RD	
;				
	BCF	STATUS, RPO	; Bank 2	
	MOVF	EEDATA, W	; W = LS Byte of Program EEDATA	
	MOVWF	DATAL		
	MOVF	EEDATH, W	; W = MS Byte of Program EEDATA	
	MOVWF	, DATAH	;	
	110 7 111			

3.6 Writing to Flash Program Memory

Flash program memory may only be written to if the destination address is in a segment of memory that is not write-protected, as defined in bits WRT1:WRT0 of the device configuration word (Register 14-1). Flash program memory must be written in four-word blocks. A block consists of four words with sequential addresses, with a lower boundary defined by an address, where EEADR<1:0> = 00. At the same time, all block writes to program memory are done as erase and write operations. The write operation is edge-aligned and cannot occur across boundaries.

To write program data, it must first be loaded into the buffer registers (see Figure 3-1). This is accomplished by first writing the destination address to EEADR and EEADRH and then writing the data to EEDATA and EEDATH. After the address and data have been set up, then the following sequence of events must be executed:

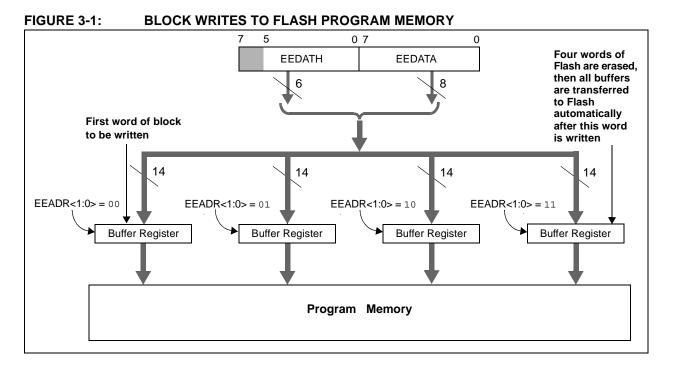
- 1. Set the EEPGD control bit (EECON1<7>).
- 2. Write 55h, then AAh, to EECON2 (Flash programming sequence).
- 3. Set the WR control bit (EECON1<1>).

All four buffer register locations **MUST** be written to with correct data. If only one, two or three words are being written to in the block of four words, then a read from the program memory location(s) not being written to must be performed. This takes the data from the program location(s) not being written and loads it into the EEDATA and EEDATH registers. Then the sequence of events to transfer data to the buffer registers must be executed. To transfer data from the buffer registers to the program memory, the EEADR and EEADRH must point to the last location in the four-word block (EEADR<1:0> = 11). Then the following sequence of events must be executed:

- 1. Set the EEPGD control bit (EECON1<7>).
- 2. Write 55h, then AAh, to EECON2 (Flash programming sequence).
- 3. Set control bit WR (EECON1<1>) to begin the write operation.

The user must follow the same specific sequence to initiate the write for each word in the program block, writing each program word in sequence (00, 01, 10, 11). When the write is performed on the last word (EEADR<1:0> = 11), the block of four words are automatically erased and the contents of the buffer registers are written into the program memory.

After the "BSF EECON1, WR" instruction, the processor requires two cycles to set up the erase/write operation. The user must place two NOP instructions after the WR bit is set. Since data is being written to buffer registers, the writing of the first three words of the block appears to occur immediately. The processor will halt internal operations for the typical 4 ms, only during the cycle in which the erase takes place (i.e., the last word of the four-word block). This is not Sleep mode as the clocks and peripherals will continue to run. After the write cycle, the processor will resume operation with the third instruction after the EECON1 write instruction. If the sequence is performed to any other location, the action is ignored.



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An example of the complete four-word write sequence is shown in Example 3-4. The initial address is loaded into the EEADRH:EEADR register pair; the four words of data are loaded using indirect addressing.

EXAMPLE 3-4: WRITING TO FLASH PROGRAM MEMORY

; This write routine assumes the following: ; 1. A valid starting address (the least significant bits = '00') is loaded in ADDRH: ADDRL ; 2. The 8 bytes of data are loaded, starting at the address in DATADDR ; 3. ADDRH, ADDRL and DATADDR are all located in shared data memory 0x70 - 0x7f ; STATUS, RP1 BSF ; ; Bank 2 BCF STATUS, RPO ADDRH,W ; Load initial address MOVF MOVWF EEADRH MOVF ADDRL,W ; MOVWF EEADR MOVF DATAADDR,W ; Load initial data address MOVWF FSR LOOP MOVF INDF,W ; Load first data byte into lower MOVWF EEDATA ; INCF FSR, F ; Next byte ; Load second data byte into upper MOVF INDF,W MOVWF EEDATH ; INCF FSR,F ; STATUS, RPO BSF ; Bank 3 EECON1, EEPGD ; Point to program memory BSF ; Enable writes BSF EECON1,WREN BCF INTCON,GIE ; Disable interrupts (if using) MOVLW 55h ; Start of required write sequence: MOVWF EECON2 ; Write 55h MOVLW AAh ; Required Sequence ; Write AAh MOVWF EECON2 BSF EECON1,WR ; Set WR bit to begin write ; Any instructions here are ignored as processor NOP ; halts to begin write sequence NOP ; processor will stop here and wait for write complete ; after write processor continues with 3rd instruction BCF EECON1,WREN ; Disable writes BSF INTCON,GIE ; Enable interrupts (if using) STATUS, RPO BCF ; Bank 2 INCF EEADR, F ; Increment address ; Check if lower two bits of address are `00' MOVF EEADR,W ANDLW 0x03 ; Indicates when four words have been programmed XORLW 0x03 ; BTFSC STATUS,Z ; Exit if more than four words, GOTO LOOP ; Continue if less than four words

3.7 Protection Against Spurious Write

There are conditions when the device should not write to the data EEPROM or Flash program memory. To protect against spurious writes, various mechanisms have been built-in. On power-up, WREN is cleared. Also, the Power-up Timer (72 ms duration) prevents an EEPROM write.

The write initiate sequence and the WREN bit together help prevent an accidental write during brown-out, power glitch or software malfunction.

3.8 Operation During Code-Protect

When the data EEPROM is code-protected, the microcontroller can read and write to the EEPROM normally. However, all external access to the EEPROM is disabled. External write access to the program memory is also disabled.

When program memory is code-protected, the microcontroller can read and write to program memory normally, as well as execute instructions. Writes by the device may be selectively inhibited to regions of the memory depending on the setting of bits WR1:WR0 of the configuration word (see **Section 14.1 "Configuration Bits"** for additional information). External access to the memory is also disabled.

TABLE 3-1:REGISTERS/BITS ASSOCIATED WITH DATA EEPROM AND
FLASH PROGRAM MEMORIES

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on Power-on Reset	Value on all other Resets
10Ch	EEDATA	EEPRON	1/Flash D		XXXX XXXX	uuuu uuuu					
10Dh	EEADR	EEPRON	1/Flash A	ddress R	egister L	ow Byte				xxxx xxxx	uuuu uuuu
10Eh	EEDATH	_		EEPRO	M/Flash [Data Regist	er High By	te		xxxx xxxx	0 q000
10Fh	EEADRH	_		-	EEPRO	M/Flash Ad	dress Regi	ister High B	syte	xxxx xxxx	
18Ch	EECON1	EEPGD		_	_	WRERR	WREN	WR	RD	x x000	0 q000
18Dh	EECON2	EEPRON	1 Control	Register	2 (not a j	physical reg	gister)				
0Dh	PIR2	_	CMIF	—	EEIF	BCLIF	_	—	CCP2IF	-0-0 00	-0-0 00
8Dh	PIE2	_	CMIE	-	EEIE	BCLIE	_	_	CCP2IE	-0-0 00	-0-0 00

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0', q = value depends upon condition. Shaded cells are not used by data EEPROM or Flash program memory.

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NOTES:

4.0 I/O PORTS

Some pins for these I/O ports are multiplexed with an alternate function for the peripheral features on the device. In general, when a peripheral is enabled, that pin may not be used as a general purpose I/O pin.

Additional information on I/O ports may be found in the PICmicro[™] Mid-Range Reference Manual (DS33023).

4.1 PORTA and the TRISA Register

PORTA is a 6-bit wide, bidirectional port. The corresponding data direction register is TRISA. Setting a TRISA bit (= 1) will make the corresponding PORTA pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISA bit (= 0) will make the corresponding PORTA pin an output (i.e., put the contents of the output latch on the selected pin).

Reading the PORTA register reads the status of the pins, whereas writing to it will write to the port latch. All write operations are read-modify-write operations. Therefore, a write to a port implies that the port pins are read, the value is modified and then written to the port data latch.

Pin RA4 is multiplexed with the Timer0 module clock input to become the RA4/T0CKI pin. The RA4/T0CKI pin is a Schmitt Trigger input and an open-drain output. All other PORTA pins have TTL input levels and full CMOS output drivers.

Other PORTA pins are multiplexed with analog inputs and the analog VREF input for both the A/D converters and the comparators. The operation of each pin is selected by clearing/setting the appropriate control bits in the ADCON1 and/or CMCON registers.

Note:	On a Power-on Reset, these pins are con-									
	figured as analog inputs and read as '0'.									
	The comparators are in the off (digital)									
	state.									

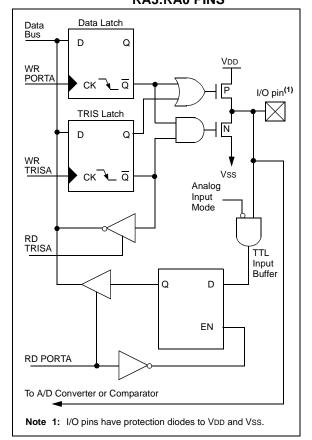
The TRISA register controls the direction of the port pins even when they are being used as analog inputs. The user must ensure the bits in the TRISA register are maintained set when using them as analog inputs.

EXAMPLE 4-1: INITIALIZING PORTA

BCF	STATUS,	RP0	;	
BCF	STATUS,	RP1	;	Bank0
CLRF	PORTA		;	Initialize PORTA by
			;	clearing output
			;	data latches
BSF	STATUS,	RP0	;	Select Bank 1
MOVLW	0x06		;	Configure all pins
MOVWF	ADCON1		;	as digital inputs
MOVLW	0xCF		;	Value used to
			;	initialize data
			;	direction
MOVWF	TRISA		;	Set RA<3:0> as inputs
			;	RA<5:4> as outputs
			;	TRISA<7:6>are always
			;	read as '0'.



BLOCK DIAGRAM OF RA3:RA0 PINS



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FIGURE 4-2: BLOCK DIAGRAM OF RA4/T0CKI PIN

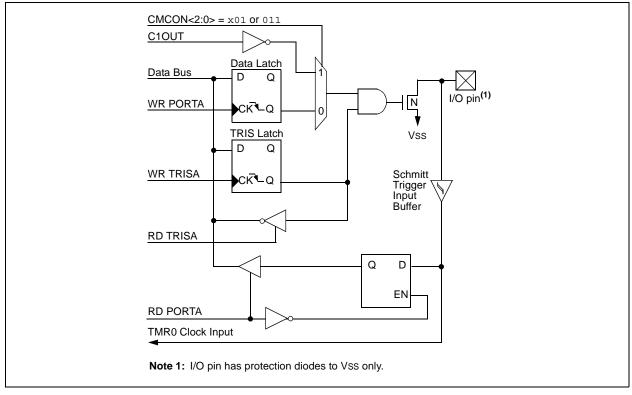
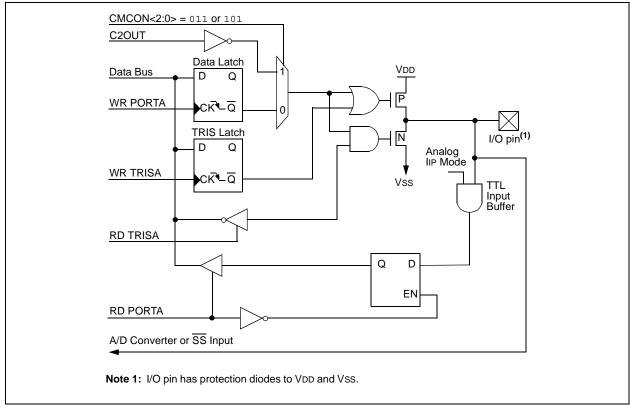


FIGURE 4-3: BLOCK DIAGRAM OF RA5 PIN



Name	Bit#	Buffer	Function					
RA0/AN0	bit 0	TTL	Input/output or analog input.					
RA1/AN1	bit 1	TTL	Input/output or analog input.					
RA2/AN2/VREF-/CVREF	bit 2	TTL	Input/output or analog input or VREF- or CVREF.					
RA3/AN3/VREF+	bit 3	TTL	Input/output or analog input or VREF+.					
RA4/T0CKI/C1OUT	bit 4	ST	Input/output or external clock input for Timer0 or comparator output. Output is open-drain type.					
RA5/AN4/SS/C2OUT	bit 5	TTL	Input/output or analog input or slave select input for synchronous serial port or comparator output.					

TABLE 4-1: PORTA FUNCTIONS

Legend: TTL = TTL input, ST = Schmitt Trigger input

TABLE 4-2: SUMMARY OF REGISTERS ASSOCIATED WITH PORTA

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets	
05h	PORTA	_	—	RA5	RA4	RA3	RA2	RA1	RA0	0x 0000	0u 0000	
85h	TRISA	_	_	PORTA D	ata Direct	ion Regist	er			11 1111	11 1111	
9Ch	CMCON	C2OUT	C1OUT	C2INV	C1INV	CIS	CM2	CM1	CM0	0000 0111	0000 0111	
9Dh	CVRCON	CVREN	CVROE	CVRR	_	CVR3	CVR2	CVR1	CVR0	000- 0000	000- 0000	
9Fh	ADCON1	ADFM	ADCS2	_	_	PCFG3	PCFG2	PCFG1	PCFG0	00 0000	00 0000	

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by PORTA.

Note: When using the SSP module in SPI Slave mode and \overline{SS} enabled, the A/D converter must be set to one of the following modes, where PCFG3:PCFG0 = 0100, 0101, 011x, 1101, 1110, 1111.

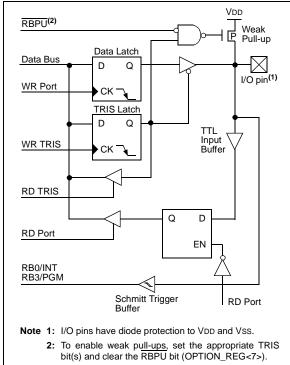
4.2 PORTB and the TRISB Register

PORTB is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISB. Setting a TRISB bit (= 1) will make the corresponding PORTB pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISB bit (= 0) will make the corresponding PORTB pin an output (i.e., put the contents of the output latch on the selected pin).

Three pins of PORTB are multiplexed with the In-Circuit Debugger and Low-Voltage Programming function: RB3/PGM, RB6/PGC and RB7/PGD. The alternate functions of these pins are described in **Section 14.0 "Special Features of the CPU"**.

Each of the PORTB pins has a weak internal pull-up. A single control bit can turn on all the pull-ups. This is performed by clearing bit RBPU (OPTION_REG<7>). The weak pull-up is automatically turned off when the port pin is configured as an output. The pull-ups are disabled on a Power-on Reset.





Four of the PORTB pins, RB7:RB4, have an interrupton-change feature. Only pins configured as inputs can cause this interrupt to occur (i.e., any RB7:RB4 pin configured as an output is excluded from the interrupton-change comparison). The input pins (of RB7:RB4) are compared with the old value latched on the last read of PORTB. The "mismatch" outputs of RB7:RB4 are OR'ed together to generate the RB port change interrupt with flag bit RBIF (INTCON<0>). This interrupt can wake the device from Sleep. The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- a) Any read or write of PORTB. This will end the mismatch condition.
- b) Clear flag bit RBIF.

A mismatch condition will continue to set flag bit RBIF. Reading PORTB will end the mismatch condition and allow flag bit RBIF to be cleared.

The interrupt-on-change feature is recommended for wake-up on key depression operation and operations where PORTB is only used for the interrupt-on-change feature. Polling of PORTB is not recommended while using the interrupt-on-change feature.

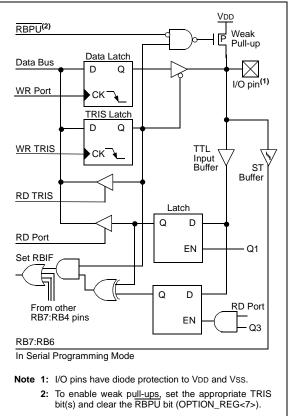
This interrupt-on-mismatch feature, together with software configurable pull-ups on these four pins, allow easy interface to a keypad and make it possible for wake-up on key depression. Refer to the application note, *AN552, "Implementing Wake-up on Key Stroke"* (DS00552).

RB0/INT is an external interrupt input pin and is configured using the INTEDG bit (OPTION_REG<6>).

RB0/INT is discussed in detail in **Section 14.11.1 "INT Interrupt**".



BLOCK DIAGRAM OF RB7:RB4 PINS



Name	Bit#	Buffer	Function
RB0/INT	bit 0	TTL/ST ⁽¹⁾	Input/output pin or external interrupt input. Internal software programmable weak pull-up.
RB1	bit 1	TTL	Input/output pin. Internal software programmable weak pull-up.
RB2	bit 2	TTL	Input/output pin. Internal software programmable weak pull-up.
RB3/PGM ⁽³⁾	bit 3	TTL	Input/output pin or programming pin in LVP mode. Internal software programmable weak pull-up.
RB4	bit 4	TTL	Input/output pin (with interrupt-on-change). Internal software programmable weak pull-up.
RB5	bit 5	TTL	Input/output pin (with interrupt-on-change). Internal software programmable weak pull-up.
RB6/PGC	bit 6	TTL/ST ⁽²⁾	Input/output pin (with interrupt-on-change) or in-circuit debugger pin. Internal software programmable weak pull-up. Serial programming clock.
RB7/PGD	bit 7	TTL/ST ⁽²⁾	Input/output pin (with interrupt-on-change) or in-circuit debugger pin. Internal software programmable weak pull-up. Serial programming data.

TABLE 4-3: PORTB FUNCTIONS

Legend: TTL = TTL input, ST = Schmitt Trigger input

Note 1: This buffer is a Schmitt Trigger input when configured as the external interrupt.

2: This buffer is a Schmitt Trigger input when used in Serial Programming mode or in-circuit debugger.

3: Low-Voltage ICSP Programming (LVP) is enabled by default which disables the RB3 I/O function. LVP must be disabled to enable RB3 as an I/O pin and allow maximum compatibility to the other 28-pin and 40-pin mid-range devices.

TABLE 4-4: SU	MMARY OF REGISTERS ASSOCIATED WITH PORTB
---------------	--

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	all o	Value on all other Resets	
06h, 106h	PORTB	RB7	RB6	RB5	RB4	RB3	RB2	RB1	RB0	xxxx xxxx	uuuu	uuuu	
86h, 186h	TRISB	PORTB	Data Direc	1111 1111	1111	1111							
81h, 181h	OPTION_REG	RBPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111	1111	

Legend: x = unknown, u = unchanged. Shaded cells are not used by PORTB.

4.3 PORTC and the TRISC Register

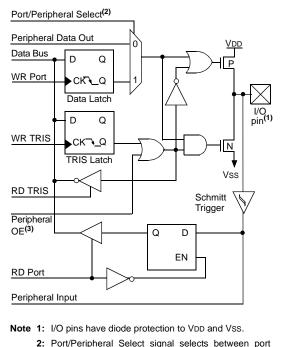
PORTC is an 8-bit wide, bidirectional port. The corresponding data direction register is TRISC. Setting a TRISC bit (= 1) will make the corresponding PORTC pin an input (i.e., put the corresponding output driver in a High-Impedance mode). Clearing a TRISC bit (= 0) will make the corresponding PORTC pin an output (i.e., put the contents of the output latch on the selected pin).

PORTC is multiplexed with several peripheral functions (Table 4-5). PORTC pins have Schmitt Trigger input buffers.

When the I^2C module is enabled, the PORTC<4:3> pins can be configured with normal I^2C levels, or with SMBus levels, by using the CKE bit (SSPSTAT<6>).

When enabling peripheral functions, care should be taken in defining TRIS bits for each PORTC pin. Some peripherals override the TRIS bit to make a pin an output, while other peripherals override the TRIS bit to make a pin an input. Since the TRIS bit override is in effect while the peripheral is enabled, read-modifywrite instructions (BSF, BCF, XORWF) with TRISC as the destination, should be avoided. The user should refer to the corresponding peripheral section for the correct TRIS bit settings.

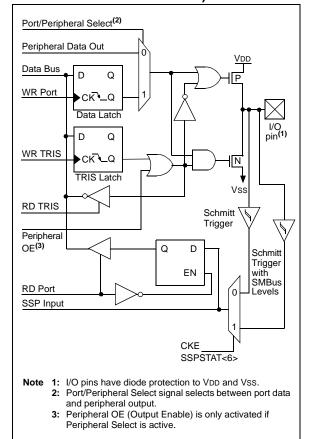
FIGURE 4-6: PORTC BLOCK DIAGRAM (PERIPHERAL OUTPUT OVERRIDE) RC<2:0>, RC<7:5>



- 2: Port/Peripheral Select signal selects between port data and peripheral output.
- 3: Peripheral OE (Output Enable) is only activated if Peripheral Select is active.

FIGURE 4-7:

PORTC BLOCK DIAGRAM (PERIPHERAL OUTPUT OVERRIDE) RC<4:3>



Name	Bit#	Buffer Type	Function
RC0/T1OSO/T1CKI	bit 0	ST	Input/output port pin or Timer1 oscillator output/Timer1 clock input.
RC1/T1OSI/CCP2	bit 1	ST	Input/output port pin or Timer1 oscillator input or Capture2 input/ Compare2 output/PWM2 output.
RC2/CCP1	bit 2	ST	Input/output port pin or Capture1 input/Compare1 output/ PWM1 output.
RC3/SCK/SCL	bit 3	ST	RC3 can also be the synchronous serial clock for both SPI and I^2C modes.
RC4/SDI/SDA	bit 4	ST	RC4 can also be the SPI data in (SPI mode) or data I/O (I ² C mode).
RC5/SDO	bit 5	ST	Input/output port pin or Synchronous Serial Port data output.
RC6/TX/CK	bit 6	ST	Input/output port pin or USART asynchronous transmit or synchronous clock.
RC7/RX/DT	bit 7	ST	Input/output port pin or USART asynchronous receive or synchronous data.

TABLE 4-5: PORTC FUNCTIONS

Legend: ST = Schmitt Trigger input

TABLE 4-0: SUMMART OF REGISTERS ASSOCIATED WITH PORT	TABLE 4-6:	SUMMARY OF REGISTERS ASSOCIATED WITH PORTC
--	------------	--

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
07h	PORTC	RC7	RC6	RC5	RC4	RC3	RC2	RC1	RC0	xxxx xxxx	uuuu uuuu
87h	TRISC	PORTC	PORTC Data Direction Register							1111 1111	1111 1111

Legend: x = unknown, u = unchanged

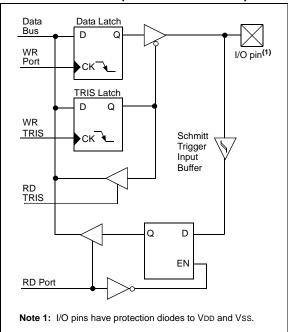
4.4 **PORTD and TRISD Registers**

Note:	PORTD and TRISD are not implemented
	on the 28-pin devices.

PORTD is an 8-bit port with Schmitt Trigger input buffers. Each pin is individually configurable as an input or output.

PORTD can be configured as an 8-bit wide microprocessor port (Parallel Slave Port) by setting control bit, PSPMODE (TRISE<4>). In this mode, the input buffers are TTL.

FIGURE 4-8: PORTD BLOCK DIAGRAM (IN I/O PORT MODE)



Name	Bit#	Buffer Type	Function
RD0/PSP0	bit 0	ST/TTL ⁽¹⁾	Input/output port pin or Parallel Slave Port bit 0.
RD1/PSP1	bit 1	ST/TTL ⁽¹⁾	Input/output port pin or Parallel Slave Port bit 1.
RD2/PSP2	bit2	ST/TTL ⁽¹⁾	Input/output port pin or Parallel Slave Port bit 2.
RD3/PSP3	bit 3	ST/TTL ⁽¹⁾	Input/output port pin or Parallel Slave Port bit 3.
RD4/PSP4	bit 4	ST/TTL ⁽¹⁾	Input/output port pin or Parallel Slave Port bit 4.
RD5/PSP5	bit 5	ST/TTL ⁽¹⁾	Input/output port pin or Parallel Slave Port bit 5.
RD6/PSP6	bit 6	ST/TTL ⁽¹⁾	Input/output port pin or Parallel Slave Port bit 6.
RD7/PSP7	bit 7	ST/TTL ⁽¹⁾	Input/output port pin or Parallel Slave Port bit 7.

TABLE 4-7:PORTD FUNCTIONS

Legend: ST = Schmitt Trigger input, TTL = TTL input

Note 1: Input buffers are Schmitt Triggers when in I/O mode and TTL buffers when in Parallel Slave Port mode.

TABLE 4-8:	SUMMARY OF REGISTERS ASSOCIATED WITH PORTD
------------	--

Address	Name	Bit 7	Bit 6	Bit 5 Bit 4 Bit 3 Bit 2 Bit 1 Bit 0					Value POR,	e on: BOR	Valu all o Res	ther	
08h	PORTD	RD7	RD7 RD6 RD5 RD4 RD3 RD2 RD1 RD0 :						xxxx	xxxx	uuuu	uuuu	
88h	TRISD	PORTI	PORTD Data Direction Register							1111	1111	1111	1111
89h	TRISE	IBF	IBF OBF IBOV PSPMODE — PORTE Data Direction Bits							0000	-111	0000	-111

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by PORTD.

4.5 PORTE and TRISE Register

Note:	PORTE and TRISE are not implemented
	on the 28-pin devices.

PORTE has three pins (RE0/RD/AN5, RE1/WR/AN6 and RE2/CS/AN7) which are individually configurable as inputs or outputs. These pins have Schmitt Trigger input buffers.

The PORTE pins become the I/O control inputs for the microprocessor port when bit PSPMODE (TRISE<4>) is set. In this mode, the user must make certain that the TRISE<2:0> bits are set and that the pins are configured as digital inputs. Also, ensure that ADCON1 is configured for digital I/O. In this mode, the input buffers are TTL.

Register 4-1 shows the TRISE register which also controls the Parallel Slave Port operation.

PORTE pins are multiplexed with analog inputs. When selected for analog input, these pins will read as '0's.

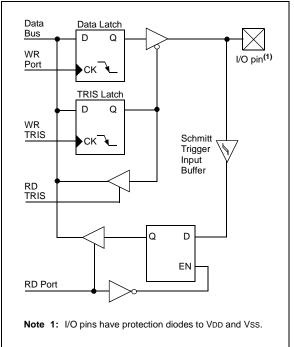
TRISE controls the direction of the RE pins, even when they are being used as analog inputs. The user must make sure to keep the pins configured as inputs when using them as analog inputs.

Note: On a Power-on Reset, these pins are configured as analog inputs and read as '0'.

TABLE 4-9: PORTE FUNCTIONS

FIGURE 4-9:

PORTE BLOCK DIAGRAM (IN I/O PORT MODE)



Name	Bit#	Buffer Type	Function
RE0/RD/AN5	bit 0	ST/TTL ⁽¹⁾	 I/O port pin or read control input in Parallel Slave Port mode or analog input: RD 1 = Idle 0 = Read operation. Contents of PORTD register are output to PORTD I/O pins (if chip selected).
RE1/WR/AN6	bit 1	ST/TTL ⁽¹⁾	 I/O port pin or write control input in Parallel Slave Port mode or analog input: WR 1 = Idle 0 = Write operation. Value of PORTD I/O pins is latched into PORTD register (if chip selected).
RE2/CS/AN7	bit 2	ST/TTL ⁽¹⁾	 I/O port pin or chip select control input in Parallel Slave Port mode or analog input: CS 1 = Device is not selected 0 = Device is selected

Legend: ST = Schmitt Trigger input, TTL = TTL input

Note 1: Input buffers are Schmitt Triggers when in I/O mode and TTL buffers when in Parallel Slave Port mode.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
09h	PORTE	—	—	_	_	_	RE2	RE1	RE0	xxx	uuu
89h	TRISE	IBF	OBF	IBOV	PSPMODE	_	PORTE D	ata Direct	tion bits	0000 -111	0000 -111
9Fh	ADCON1	ADFM	ADCS2	—	—	PCFG3	PCFG2	PCFG1	PCFG0	00 0000	00 0000

TABLE 4-10: SUMMARY OF REGISTERS ASSOCIATED WITH PORTE

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by PORTE.

REGISTER 4-1: TRISE REGISTER (ADDRESS 89h)

R-0	R-0	R/W-0	R/W-0	U-0	R/W-1	R/W-1	R/W-1
IBF	OBF	IBOV	PSPMODE	—	Bit 2	Bit 1	Bit 0
bit 7							bit 0

Parallel Slave Port Status/Control Bits:

bit 7	IBF: Input Buffer Full Status bit
	 1 = A word has been received and is waiting to be read by the CPU 0 = No word has been received
bit 6	OBF : Output Buffer Full Status bit
	 1 = The output buffer still holds a previously written word 0 = The output buffer has been read
bit 5	IBOV: Input Buffer Overflow Detect bit (in Microprocessor mode)
	 1 = A write occurred when a previously input word has not been read (must be cleared in software) 0 = No overflow occurred
bit 4	PSPMODE: Parallel Slave Port Mode Select bit
	1 = PORTD functions in Parallel Slave Port mode
	0 = PORTD functions in general purpose I/O mode
bit 3	Unimplemented: Read as '0'
	PORTE Data Direction Bits:
bit 2	Bit 2: Direction Control bit for pin RE2/CS/AN7
	1 = Input 0 = Output
bit 1	Bit 1: Direction Control bit for pin RE1/WR/AN6
	1 = Input 0 = Output
bit 0	Bit 0: Direction Control bit for pin RE0/RD/AN5
	1 = Input 0 = Output
	Legend:
	R = Readable bit $W = Writable bit$ $U = Unimplemented bit, read as '0'$
	- n = Value at POR '1' = Bit is set '0' = Bit is cleared x = Bit is unknown

4.6 Parallel Slave Port

The Parallel Slave Port (PSP) is not implemented on the PIC16F873A or PIC16F876A.

PORTD operates as an 8-bit wide Parallel Slave Port, or microprocessor port, when control bit PSPMODE (TRISE<4>) is set. In Slave mode, it is asynchronously readable and writable by the external world through RD control input pin, RE0/RD/AN5, and WR control input pin, RE1/WR/AN6.

The PSP can directly interface to an 8-bit microprocessor data bus. The external microprocessor can read or write the PORTD latch as an 8-bit latch. Setting bit PSPMODE enables port pin RE0/RD/AN5 to be the RD input, RE1/WR/AN6 to be the WR input and RE2/CS/AN7 to be the CS (Chip Select) input. For this functionality, the corresponding data direction bits of the TRISE register (TRISE<2:0>) must be configured as inputs (set). The A/D port configuration bits, PCFG3:PCFG0 (ADCON1<3:0>), must be set to configure pins RE2:RE0 as digital I/O.

There are actually two 8-bit latches: one for data output and one for data input. The user writes 8-bit data to the PORTD data latch and reads data from the port pin latch (note that they have the same address). In this mode, the TRISD register is ignored since the external device is controlling the direction of data flow.

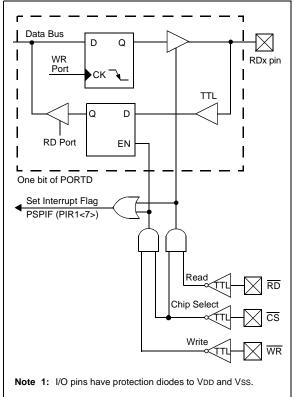
A write to the PSP occurs when both the \overline{CS} and \overline{WR} lines are first detected low. When either the \overline{CS} or \overline{WR} lines become high (level triggered), the Input Buffer Full (IBF) status flag bit (TRISE<7>) is set on the Q4 clock cycle, following the next Q2 cycle, to signal the write is complete (Figure 4-11). The interrupt flag bit, PSPIF (PIR1<7>), is also set on the same Q4 clock cycle. IBF can only be cleared by reading the PORTD input latch. The Input Buffer Overflow (IBOV) status flag bit (TRISE<5>) is set if a second write to the PSP is attempted when the previous byte has not been read out of the buffer.

A read from the PSP occurs when both the \overline{CS} and \overline{RD} lines are first detected low. The Output Buffer Full (OBF) status flag bit (TRISE<6>) is cleared immediately (Figure 4-12), indicating that the PORTD latch is waiting to be read by the external bus. When either the \overline{CS} or \overline{RD} pin becomes high (level triggered), the interrupt flag bit PSPIF is set on the Q4 clock cycle, following the next Q2 cycle, indicating that the read is complete. OBF remains low until data is written to PORTD by the user firmware.

When not in PSP mode, the IBF and OBF bits are held clear. However, if flag bit IBOV was previously set, it must be cleared in firmware.

An interrupt is generated and latched into flag bit PSPIF when a read or write operation is completed. PSPIF must be cleared by the user in firmware and the interrupt can be disabled by clearing the interrupt enable bit PSPIE (PIE1<7>).

FIGURE 4-10: PORTD AND PORTE BLOCK DIAGRAM (PARALLEL SLAVE PORT)



PIC16F87XA

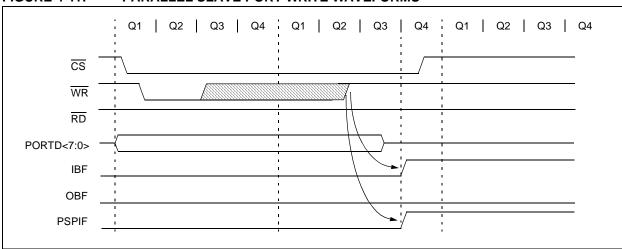


FIGURE 4-11: PARALLEL SLAVE PORT WRITE WAVEFORMS



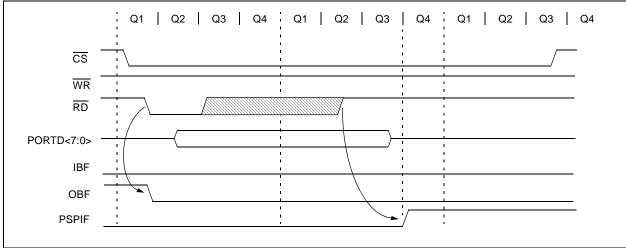


TABLE 4-11: REGISTERS ASSOCIATED WITH PARALLEL SLAVE PORT

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value POR,		all c	ie on other sets
08h	PORTD	Port Data	Latch wh	en writte	en; Port pins w	hen read				xxxx	xxxx	uuuu	uuuu
09h	PORTE	—	_	—	_	_	RE2	RE1	RE0		-xxx		-uuu
89h	TRISE	IBF	OBF	IBOV	PSPMODE	_	PORTE D	ata Directi	on bits	0000	-111	0000	-111
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000	0000	0000	0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000	0000	0000	0000
9Fh	ADCON1	ADFM	ADCS2	_		PCFG3	PCFG2	PCFG1	PCFG0	00	0000	00	0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the Parallel Slave Port. **Note 1:** Bits PSPIE and PSPIF are reserved on the PIC16F873A/876A; always maintain these bits clear.

5.0 TIMER0 MODULE

The Timer0 module timer/counter has the following features:

- 8-bit timer/counter
- Readable and writable
- 8-bit software programmable prescaler
- Internal or external clock select
- Interrupt on overflow from FFh to 00h
- Edge select for external clock

Figure 5-1 is a block diagram of the Timer0 module and the prescaler shared with the WDT.

Additional information on the Timer0 module is available in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

Timer mode is selected by clearing bit TOCS (OPTION_REG<5>). In Timer mode, the Timer0 module will increment every instruction cycle (without prescaler). If the TMR0 register is written, the increment is inhibited for the following two instruction cycles. The user can work around this by writing an adjusted value to the TMR0 register.

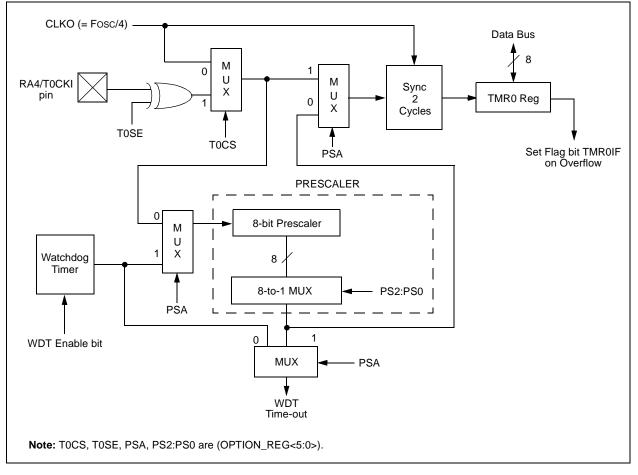
Counter mode is selected by setting bit TOCS (OPTION_REG<5>). In Counter mode, Timer0 will increment either on every rising or falling edge of pin RA4/T0CKI. The incrementing edge is determined by the Timer0 Source Edge Select bit, T0SE (OPTION_REG<4>). Clearing bit T0SE selects the rising edge. Restrictions on the external clock input are discussed in detail in Section 5.2 "Using Timer0 with an External Clock".

The prescaler is mutually exclusively shared between the Timer0 module and the Watchdog Timer. The prescaler is not readable or writable. **Section 5.3** "**Prescaler**" details the operation of the prescaler.

5.1 Timer0 Interrupt

The TMR0 interrupt is generated when the TMR0 register overflows from FFh to 00h. This overflow sets bit TMR0IF (INTCON<2>). The interrupt can be masked by clearing bit TMR0IE (INTCON<5>). Bit TMR0IF must be cleared in software by the Timer0 module Interrupt Service Routine before re-enabling this interrupt. The TMR0 interrupt cannot awaken the processor from Sleep since the timer is shut-off during Sleep.

FIGURE 5-1: BLOCK DIAGRAM OF THE TIMER0/WDT PRESCALER



5.2 Using Timer0 with an External Clock

When no prescaler is used, the external clock input is the same as the prescaler output. The synchronization of T0CKI with the internal phase clocks is accomplished by sampling the prescaler output on the Q2 and Q4 cycles of the internal phase clocks. Therefore, it is necessary for T0CKI to be high for at least 2 Tosc (and a small RC delay of 20 ns) and low for at least 2 Tosc (and a small RC delay of 20 ns). Refer to the electrical specification of the desired device.

5.3 Prescaler

REGISTER 5-1:

There is only one prescaler available which is mutually exclusively shared between the Timer0 module and the Watchdog Timer. A prescaler assignment for the

OPTION REG REGISTER

Timer0 module means that there is no prescaler for the Watchdog Timer and vice versa. This prescaler is not readable or writable (see Figure 5-1).

The PSA and PS2:PS0 bits (OPTION_REG<3:0>) determine the prescaler assignment and prescale ratio.

When assigned to the Timer0 module, all instructions writing to the TMR0 register (e.g., CLRF1, MOVWF1, BSF1, x....etc.) will clear the prescaler. When assigned to WDT, a CLRWDT instruction will clear the prescaler along with the Watchdog Timer. The prescaler is not readable or writable.

Note: Writing to TMR0 when the prescaler is assigned to Timer0 will clear the prescaler count, but will not change the prescaler assignment.

i. Or								
	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1	R/W-1
	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0
bi	t 7							bit 0
		2						
R	BPU 👱							
IN	TEDG							
т	DCS: TMF	R0 Clock Sou	urce Select	t bit				
		ion on TOCK	•	(CLKO)				
		R0 Source E	•	. ,				
1	= Increm	ent on high-t ent on low-to	o-low trans	sition on TO				
P	SA: Pres	caler Assigni	ment bit					
		ler is assigne ler is assigne			le			
P	S2:PS0:	Prescaler Ra	te Select b	oits				
E	Bit Value	TMR0 Rate	WDT Rate	•				
	000	1:2	1:1	-				
	001	1:4	1:2					
	010 011	1:8 1:16	1:4 1:8					
	100	1:32	1:16					
	101	1:64	1:32					
	110	1:128	1:64					
	111	1 : 256	1 : 128					
	egend:							
R	= Reada	ble bit	VV = V	Vritable bit	U = Unimp	lemented b	it, read as '	0'
- I	n = Value	at POR	'1' = E	Bit is set	'0' = Bit is (cleared	x = Bit is u	nknown
I	Note:	PICmicro [®] N	/lid-Range	MCU Fami	Reset, the inst ly Reference M ler assignment	/lanual (DS	33023) mu	st be exe-

sequence must be followed even if the WDT is disabled.

TABLE 3-1. REGISTERS ASSOCIATED WITH HIVERV	TABLE 5-1:	REGISTERS ASSOCIATED WITH TIMER0
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Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
01h,101h	TMR0	Timer0 N	imer0 Module Register							xxxx xxxx	uuuu uuuu
0Bh,8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 000x	0000 000u
81h,181h	OPTION_REG	RBPU	INTEDG	TOCS	T0SE	PSA	PS2	PS1	PS0	1111 1111	1111 1111

Legend: x = unknown, u = unchanged, - = unimplemented locations read as '0'. Shaded cells are not used by Timer0.

PIC16F87XA

NOTES:

6.0 TIMER1 MODULE

The Timer1 module is a 16-bit timer/counter consisting of two 8-bit registers (TMR1H and TMR1L) which are readable and writable. The TMR1 register pair (TMR1H:TMR1L) increments from 0000h to FFFFh and rolls over to 0000h. The TMR1 interrupt, if enabled, is generated on overflow which is latched in interrupt flag bit, TMR1IF (PIR1<0>). This interrupt can be enabled/disabled by setting/clearing TMR1 interrupt enable bit, TMR1IE (PIE1<0>).

Timer1 can operate in one of two modes:

- As a Timer
- As a Counter

The operating mode is determined by the clock select bit, TMR1CS (T1CON<1>).

In Timer mode, Timer1 increments every instruction cycle. In Counter mode, it increments on every rising edge of the external clock input.

Timer1 can be enabled/disabled by setting/clearing control bit, TMR1ON (T1CON<0>).

Timer1 also has an internal "Reset input". This Reset can be generated by either of the two CCP modules (Section 8.0 "Capture/Compare/PWM Modules"). Register 6-1 shows the Timer1 Control register.

When the Timer1 oscillator is enabled (T1OSCEN is set), the RC1/T1OSI/CCP2 and RC0/T1OSO/T1CKI pins become inputs. That is, the TRISC<1:0> value is ignored and these pins read as '0'.

Additional information on timer modules is available in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

STER 6-1:	T1CON: TI	MER1 CO	ONTROL R		ADDRESS 1	0h)				
	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0		
	_	—	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR10N		
	bit 7							bit 0		
bit 7-6	Unimpleme	ented: Rea	ad as '0'							
bit 5-4	T1CKPS1:1	T1CKPS0:	Timer1 Inpu	ut Clock Pres	cale Select bit	S				
	11 = 1:8 pre									
		10 = 1:4 prescale value								
		01 = 1:2 prescale value								
h it 0	-	00 = 1:1 prescale value								
bit 3		T1OSCEN: Timer1 Oscillator Enable Control bit								
		 1 = Oscillator is enabled 0 = Oscillator is shut-off (the oscillator inverter is turned off to eliminate power drain) 								
h it 0										
bit 2		T1SYNC: Timer1 External Clock Input Synchronization Control bit								
		<u>When TMR1CS = 1:</u> 1 = Do not synchronize external clock input								
			nal clock in							
	When TMR			Jul						
			ner1 uses th	e internal clo	ock when TMR	1CS = 0.				
bit 1	TMR1CS: T	imer1 Clo	ck Source S	elect bit						
	1 = Externa	I clock from	m pin RC0/T	10SO/T1Ck	(I (on the rising	edge)				
	0 = Internal	clock (Fo	sc/4)							
bit 0	TMR1ON: 7	Fimer1 On	bit							
	1 = Enables	1 = Enables Timer1								
	0 = Stops T	imer1								
	Legend:									
	R = Readat	ole bit	W = V	Vritable bit	U = Unimpl	emented b	it, read as '	0'		
	- n = Value	at POR	'1' = E	Bit is set	'0' = Bit is c	leared	x = Bit is u	nknown		

REGISTER 6-1: T1CON: TIMER1 CONTROL REGISTER (ADDRESS 10h)

6.1 Timer1 Operation in Timer Mode

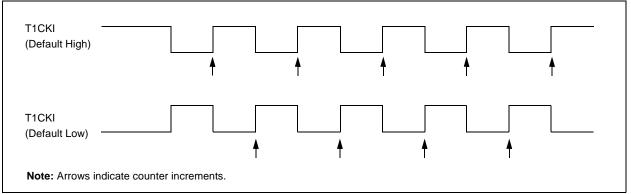
Timer mode is selected by clearing the TMR1CS (T1CON<1>) bit. In this mode, the input clock to the timer is FOSC/4. The synchronize control bit, T1SYNC (T1CON<2>), has no effect since the internal clock is always in sync.

6.2 Timer1 Counter Operation

Timer1 may operate in either a Synchronous, or an Asynchronous mode, depending on the setting of the TMR1CS bit.

When Timer1 is being incremented via an external source, increments occur on a rising edge. After Timer1 is enabled in Counter mode, the module must first have a falling edge before the counter begins to increment.

FIGURE 6-1: TIMER1 INCREMENTING EDGE



6.3 Timer1 Operation in Synchronized Counter Mode

Counter mode is selected by setting bit TMR1CS. In this mode, the timer increments on every rising edge of clock input on pin RC1/T1OSI/CCP2 when bit T1OSCEN is set, or on pin RC0/T1OSO/T1CKI when bit T1OSCEN is cleared.

If $\overline{\text{T1SYNC}}$ is cleared, then the external clock input is synchronized with internal phase clocks. The synchronization is done after the prescaler stage. The prescaler stage is an asynchronous ripple counter.

In this configuration, during Sleep mode, Timer1 will not increment even if the external clock is present since the synchronization circuit is shut-off. The prescaler, however, will continue to increment.

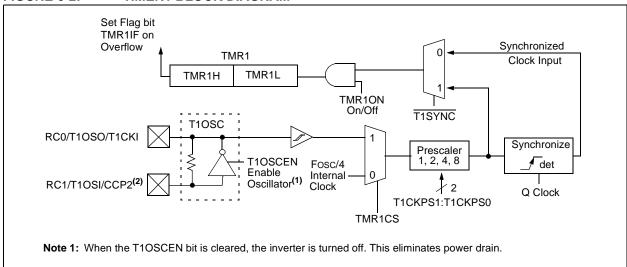


FIGURE 6-2: TIMER1 BLOCK DIAGRAM

6.4 Timer1 Operation in Asynchronous Counter Mode

If control bit $\overline{T1SYNC}$ (T1CON<2>) is set, the external clock input is not synchronized. The timer continues to increment asynchronous to the internal phase clocks. The timer will continue to run during Sleep and can generate an interrupt-on-overflow which will wake-up the processor. However, special precautions in software are needed to read/write the timer.

In Asynchronous Counter mode, Timer1 cannot be used as a time base for capture or compare operations.

6.4.1 READING AND WRITING TIMER1 IN ASYNCHRONOUS COUNTER MODE

Reading TMR1H or TMR1L while the timer is running from an external asynchronous clock will ensure a valid read (taken care of in hardware). However, the user should keep in mind that reading the 16-bit timer in two 8-bit values itself, poses certain problems, since the timer may overflow between the reads.

For writes, it is recommended that the user simply stop the timer and write the desired values. A write contention may occur by writing to the timer registers while the register is incrementing. This may produce an unpredictable value in the timer register.

Reading the 16-bit value requires some care. Examples 12-2 and 12-3 in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023) show how to read and write Timer1 when it is running in Asynchronous mode.

6.5 Timer1 Oscillator

A crystal oscillator circuit is built-in between pins T1OSI (input) and T1OSO (amplifier output). It is enabled by setting control bit, T1OSCEN (T1CON<3>). The oscillator is a low-power oscillator, rated up to 200 kHz. It will continue to run during Sleep. It is primarily intended for use with a 32 kHz crystal. Table 6-1 shows the capacitor selection for the Timer1 oscillator.

The Timer1 oscillator is identical to the LP oscillator. The user must provide a software time delay to ensure proper oscillator start-up.

TABLE 6-1:CAPACITOR SELECTION FOR
THE TIMER1 OSCILLATOR

Osc Type	Freq.	C1	C2					
LP	32 kHz	33 pF	33 pF					
	100 kHz	15 pF	15 pF					
200 kHz 15 pF 15 pF								
These values are for design guidance only.								
	Crystals	Tested:						
32.768 kHz	Epson C-00	1R32.768K-A	± 20 PPM					
100 kHz	Epson C-2 100.00 KC-P ± 20 PPM							
200 kHz	STD XTL 200.000 kHz ± 20 PPM							
Note 1. Hi	abor conocito	nco incroasos	the stability					

Note 1: Higher capacitance increases the stability of oscillator but also increases the start-up time.

2: Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.

6.6 Resetting Timer1 Using a CCP Trigger Output

If the CCP1 or CCP2 module is configured in Compare mode to generate a "special event trigger" (CCP1M3:CCP1M0 = 1011), this signal will reset Timer1.

Note:	The special event triggers from the CCP1
	and CCP2 modules will not set interrupt
	flag bit, TMR1IF (PIR1<0>).

Timer1 must be configured for either Timer or Synchronized Counter mode to take advantage of this feature. If Timer1 is running in Asynchronous Counter mode, this Reset operation may not work.

In the event that a write to Timer1 coincides with a special event trigger from CCP1 or CCP2, the write will take precedence.

In this mode of operation, the CCPRxH:CCPRxL register pair effectively becomes the period register for Timer1.

6.7 Resetting of Timer1 Register Pair (TMR1H, TMR1L)

TMR1H and TMR1L registers are not reset to 00h on a POR, or any other Reset, except by the CCP1 and CCP2 special event triggers.

T1CON register is reset to 00h on a Power-on Reset, or a Brown-out Reset, which shuts off the timer and leaves a 1:1 prescale. In all other Resets, the register is unaffected.

6.8 Timer1 Prescaler

The prescaler counter is cleared on writes to the TMR1H or TMR1L registers.

TABLE 6-2: REGISTERS ASSOCIATED WITH TIMER1 AS A TIMER/COUNTER

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value POR,		all o	e on other sets
0Bh,8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE INTE RBIE TMR0IF INTF RBIF						0000	000x	0000	000u
0Ch	PIR1	PSPIF ⁽¹⁾	PSPIF ⁽¹⁾ ADIF RCIF TXIF SSPIF CCP1IF TMR2IF TMR1IF						0000	0000	0000	0000	
8Ch	PIE1	PSPIE ⁽¹⁾	(1) ADIE RCIE TXIE SSPIE CCP1IE TMR2IE TMR1IE						0000	0000	0000	0000	
0Eh	TMR1L	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register						r	xxxx	xxxx	uuuu	uuuu	
0Fh	TMR1H	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register								xxxx	xxxx	uuuu	uuuu
10h	T1CON	_	_	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR1ON	00	0000	uu	uuuu

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the Timer1 module.

Note 1: Bits PSPIE and PSPIF are reserved on the 28-pin devices; always maintain these bits clear.

7.0 **TIMER2 MODULE**

Timer2 is an 8-bit timer with a prescaler and a postscaler. It can be used as the PWM time base for the PWM mode of the CCP module(s). The TMR2 register is readable and writable and is cleared on any device Reset.

The input clock (Fosc/4) has a prescale option of 1:1, 1:4 or 1:16, selected by control bits T2CKPS1:T2CKPS0 (T2CON<1:0>).

The Timer2 module has an 8-bit period register, PR2. Timer2 increments from 00h until it matches PR2 and then resets to 00h on the next increment cycle. PR2 is a readable and writable register. The PR2 register is initialized to FFh upon Reset.

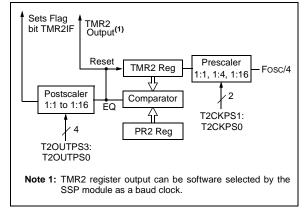
The match output of TMR2 goes through a 4-bit postscaler (which gives a 1:1 to 1:16 scaling inclusive) to generate a TMR2 interrupt (latched in flag bit, TMR2IF (PIR1<1>)).

Timer2 can be shut-off by clearing control bit, TMR2ON (T2CON<2>), to minimize power consumption.

Register 7-1 shows the Timer2 Control register.

Additional information on timer modules is available in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

FIGURE 7-1: **TIMER2 BLOCK DIAGRAM**



REGISTER

R 7-1:	T2CON: 1	IMER2 CO		EGISTER		12h)		
	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
		TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0
	bit 7							bit 0
bit 7	Unimplen	nented: Read	d as '0'					
bit 6-3	TOUTPS3	:TOUTPS0:	Timer2 Out	out Postscal	e Select bits			
		1 postscale						
		2 postscale						
	0010 = 1:	3 postscale						
	•							
	•							
	1111 = 1 :	16 postscale						
bit 2	TMR2ON:	Timer2 On b	oit					
	1 = Timer2	2 is on						
	0 = Timer2	2 is off						
bit 1-0	T2CKPS1	:T2CKPS0:	Timer2 Cloc	k Prescale S	Select bits			
	00 = Pres	caler is 1						
	01 = Pres							
	1x = Pres	caler is 16						
	Legend:							
	R = Reada	able bit	W = W	/ritable bit	U = Unim	plemented	bit, read as	'0'
	- n = Value	e at POR	'1' = B	it is set	'0' = Bit is	s cleared	x = Bit is u	nknown

7.1 Timer2 Prescaler and Postscaler

The prescaler and postscaler counters are cleared when any of the following occurs:

- a write to the TMR2 register
- a write to the T2CON register
- any device Reset (POR, MCLR Reset, WDT Reset or BOR)

TMR2 is not cleared when T2CON is written.

7.2 Output of TMR2

The output of TMR2 (before the postscaler) is fed to the SSP module, which optionally uses it to generate the shift clock.

TABLE 7-1:	REGISTERS ASSOCIATED WITH TIMER2 AS A TIMER/COUNTER

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value POR,	-		e on other sets
0Bh, 8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000	000x	0000	000u
0Ch	PIR1	PSPIF ⁽¹⁾ ADIF RCIF TXIF SSPIF CCP1IF TMR2IF TMR1IF							0000	0000	0000	0000	
8Ch	PIE1	PSPIE ⁽¹⁾	PSPIE ⁽¹⁾ ADIE RCIE TXIE SSPIE CCP1IE TMR2IE TMR1IE							0000	0000	0000	0000
11h	TMR2	Timer2 M	Timer2 Module's Register							0000	0000	0000	0000
12h	T2CON	- TOUTPS3 TOUTPS2 TOUTPS1 TOUTPS0 TMR2ON T2CKPS1 T2CKPS0							-000	0000	-000	0000	
92h	PR2	Timer2 P	eriod Regis	ter						1111	1111	1111	1111

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by the Timer2 module.

Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices; always maintain these bits clear.

8.0 CAPTURE/COMPARE/PWM MODULES

Each Capture/Compare/PWM (CCP) module contains a 16-bit register which can operate as a:

- 16-bit Capture register
- 16-bit Compare register
- PWM Master/Slave Duty Cycle register

Both the CCP1 and CCP2 modules are identical in operation, with the exception being the operation of the special event trigger. Table 8-1 and Table 8-2 show the resources and interactions of the CCP module(s). In the following sections, the operation of a CCP module is described with respect to CCP1. CCP2 operates the same as CCP1 except where noted.

CCP1 Module:

Capture/Compare/PWM Register 1 (CCPR1) is comprised of two 8-bit registers: CCPR1L (low byte) and CCPR1H (high byte). The CCP1CON register controls the operation of CCP1. The special event trigger is generated by a compare match and will reset Timer1.

CCP2 Module:

Capture/Compare/PWM Register 2 (CCPR2) is comprised of two 8-bit registers: CCPR2L (low byte) and CCPR2H (high byte). The CCP2CON register controls the operation of CCP2. The special event trigger is generated by a compare match and will reset Timer1 and start an A/D conversion (if the A/D module is enabled).

Additional information on CCP modules is available in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023) and in application note *AN594*, *"Using the CCP Module(s)"* (DS00594).

TABLE 8-1: CCP MODE – TIMER RESOURCES REQUIRED

CCP Mode	Timer Resource
Capture	Timer1
Compare	Timer1
PWM	Timer2

TABLE 8-2:INTERACTION OF TWO CCP MODULES

CCPx Mode	CCPy Mode	Interaction
Capture	Capture	Same TMR1 time base
Capture	Compare	The compare should be configured for the special event trigger which clears TMR1
Compare	Compare	The compare(s) should be configured for the special event trigger which clears TMR1
PWM	PWM	The PWMs will have the same frequency and update rate (TMR2 interrupt)
PWM	Capture	None
PWM	Compare	None

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REGISTER 8-1:	CCP1CON		ER/CCP2C	ON REGIS	STER (ADD	RESS 17h	/1Dh)					
	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0				
		—	CCPxX	CCPxY	CCPxM3	CCPxM2	CCPxM1	CCPxM0				
	bit 7							bit 0				
bit 7-6	Unimplem	ented: Rea	d as '0'									
bit 5-4	CCPxX:CC	CPxY: PWM	Least Sign	ificant bits								
	<u>Capture m</u> Unused.	<u>ode</u> :										
	<u>Compare n</u> Unused.	<u>node:</u>										
	<u>PWM mode</u> These bits		LSbs of the	e PWM duty	cycle. The eig	ght MSbs ar	e found in (CCPRxL.				
bit 3-0	CCPxM3:CCPxM0: CCPx Mode Select bits											
				· ·	ets CCPx mod	dule)						
		pture mode										
	0101 = Capture mode, every rising edge											
	0110 = Capture mode, every 4th rising edge 0111 = Capture mode, every 16th rising edge											
	1000 = Compare mode, set output on match (CCPxIF bit is set)											
	1001 = Compare mode, set output on match (CCPxIF bit is set)											
	1010 = Compare mode, generate software interrupt on match (CCPxIF bit is set, CCPx pin is unaffected)											
	1011 = Compare mode, trigger special event (CCPxIF bit is set, CCPx pin is unaffected); CCP1 resets TMR1; CCP2 resets TMR1 and starts an A/D conversion (if A/D module is enabled)											
	11xx = PW	VM mode										
	Legend:											

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented	bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

8.1 Capture Mode

In Capture mode, CCPR1H:CCPR1L captures the 16-bit value of the TMR1 register when an event occurs on pin RC2/CCP1. An event is defined as one of the following:

- Every falling edge
- Every rising edge
- Every 4th rising edge
- Every 16th rising edge

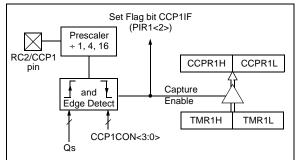
The type of event is configured by control bits, CCP1M3:CCP1M0 (CCPxCON<3:0>). When a capture is made, the interrupt request flag bit, CCP1IF (PIR1<2>), is set. The interrupt flag must be cleared in software. If another capture occurs before the value in register CCPR1 is read, the old captured value is overwritten by the new value.

8.1.1 CCP PIN CONFIGURATION

In Capture mode, the RC2/CCP1 pin should be configured as an input by setting the TRISC<2> bit.

Note:	If the RC2/CCP1 pin is configured as an
	output, a write to the port can cause a
	Capture condition.

FIGURE 8-1: CAPTURE MODE OPERATION BLOCK DIAGRAM



8.1.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode, or Synchronized Counter mode, for the CCP module to use the capture feature. In Asynchronous Counter mode, the capture operation may not work.

8.1.3 SOFTWARE INTERRUPT

When the Capture mode is changed, a false capture interrupt may be generated. The user should keep bit CCP1IE (PIE1<2>) clear to avoid false interrupts and should clear the flag bit, CCP1IF, following any such change in operating mode.

8.1.4 CCP PRESCALER

There are four prescaler settings, specified by bits CCP1M3:CCP1M0. Whenever the CCP module is turned off, or the CCP module is not in Capture mode, the prescaler counter is cleared. Any Reset will clear the prescaler counter.

Switching from one capture prescaler to another may generate an interrupt. Also, the prescaler counter will not be cleared, therefore, the first capture may be from a non-zero prescaler. Example 8-1 shows the recommended method for switching between capture prescalers. This example also clears the prescaler counter and will not generate the "false" interrupt.

EXAMPLE 8-1: CHANGING BETWEEN CAPTURE PRESCALERS

-			Turn CCP module off
MOVLW	NEW_CAPT_PS	;	Load the W reg with
		;	the new prescaler
		;	move value and CCP ON
MOVWF	CCP1CON	;	Load CCP1CON with this
		;	value

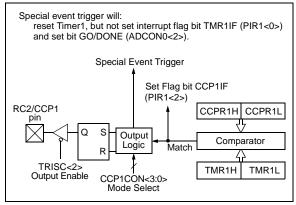
8.2 Compare Mode

In Compare mode, the 16-bit CCPR1 register value is constantly compared against the TMR1 register pair value. When a match occurs, the RC2/CCP1 pin is:

- Driven high
- Driven low
- Remains unchanged

The action on the pin is based on the value of control bits, CCP1M3:CCP1M0 (CCP1CON<3:0>). At the same time, interrupt flag bit CCP1IF is set.

FIGURE 8-2: COMPARE MODE OPERATION BLOCK DIAGRAM



8.2.1 CCP PIN CONFIGURATION

The user must configure the RC2/CCP1 pin as an output by clearing the TRISC<2> bit.

Note:	Clearing the CCP1CON register will force
	the RC2/CCP1 compare output latch to
	the default low level. This is not the
	PORTC I/O data latch.

8.2.2 TIMER1 MODE SELECTION

Timer1 must be running in Timer mode, or Synchronized Counter mode, if the CCP module is using the compare feature. In Asynchronous Counter mode, the compare operation may not work.

8.2.3 SOFTWARE INTERRUPT MODE

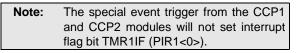
When Generate Software Interrupt mode is chosen, the CCP1 pin is not affected. The CCPIF bit is set, causing a CCP interrupt (if enabled).

8.2.4 SPECIAL EVENT TRIGGER

In this mode, an internal hardware trigger is generated which may be used to initiate an action.

The special event trigger output of CCP1 resets the TMR1 register pair. This allows the CCPR1 register to effectively be a 16-bit programmable period register for Timer1.

The special event trigger output of CCP2 resets the TMR1 register pair and starts an A/D conversion (if the A/D module is enabled).



8.3 PWM Mode (PWM)

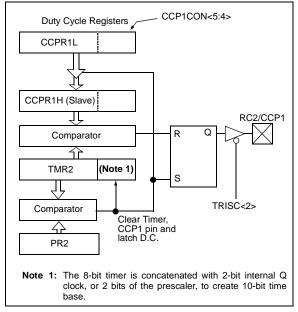
In Pulse Width Modulation mode, the CCPx pin produces up to a 10-bit resolution PWM output. Since the CCP1 pin is multiplexed with the PORTC data latch, the TRISC<2> bit must be cleared to make the CCP1 pin an output.

Note:	Clearing the CCP1CON register will force
	the CCP1 PWM output latch to the default
	low level. This is not the PORTC I/O data
	latch.

Figure 8-3 shows a simplified block diagram of the CCP module in PWM mode.

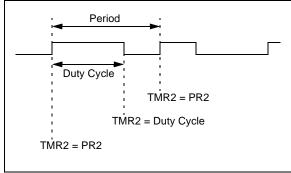
For a step-by-step procedure on how to set up the CCP module for PWM operation, see **Section 8.3.3 "Setup for PWM Operation**".





A PWM output (Figure 8-4) has a time base (period) and a time that the output stays high (duty cycle). The frequency of the PWM is the inverse of the period (1/period).





8.3.1 PWM PERIOD

The PWM period is specified by writing to the PR2 register. The PWM period can be calculated using the following formula:

 $PWM Period = [(PR2) + 1] \bullet 4 \bullet TOSC \bullet$ (TMR2 Prescale Value)

PWM frequency is defined as 1/[PWM period].

When TMR2 is equal to PR2, the following three events occur on the next increment cycle:

- TMR2 is cleared
- The CCP1 pin is set (exception: if PWM duty cycle = 0%, the CCP1 pin will not be set)
- The PWM duty cycle is latched from CCPR1L into CCPR1H

Note:	The Timer2 postscaler (see Section 7.1
	"Timer2 Prescaler and Postscaler") is
	not used in the determination of the PWM
	frequency. The postscaler could be used
	to have a servo update rate at a different
	frequency than the PWM output.

8.3.2 PWM DUTY CYCLE

The PWM duty cycle is specified by writing to the CCPR1L register and to the CCP1CON<5:4> bits. Up to 10-bit resolution is available. The CCPR1L contains the eight MSbs and the CCP1CON<5:4> contains the two LSbs. This 10-bit value is represented by CCPR1L:CCP1CON<5:4>. The following equation is used to calculate the PWM duty cycle in time:

PWM Duty Cycle =(CCPR1L:CCP1CON<5:4>) • Tosc • (TMR2 Prescale Value)

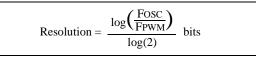
CCPR1L and CCP1CON<5:4> can be written to at any time, but the duty cycle value is not latched into CCPR1H until after a match between PR2 and TMR2 occurs (i.e., the period is complete). In PWM mode, CCPR1H is a read-only register.

The CCPR1H register and a 2-bit internal latch are used to double-buffer the PWM duty cycle. This double-buffering is essential for glitch-free PWM operation.

When the CCPR1H and 2-bit latch match TMR2, concatenated with an internal 2-bit Q clock or 2 bits of the TMR2 prescaler, the CCP1 pin is cleared.

The maximum PWM resolution (bits) for a given PWM frequency is given by the following formula.

EQUATION 8-1:



Note: If the PWM duty cycle value is longer than the PWM period, the CCP1 pin will not be cleared.

8.3.3 SETUP FOR PWM OPERATION

The following steps should be taken when configuring the CCP module for PWM operation:

- 1. Set the PWM period by writing to the PR2 register.
- Set the PWM duty cycle by writing to the CCPR1L register and CCP1CON<5:4> bits.
- Make the CCP1 pin an output by clearing the TRISC<2> bit.
- 4. Set the TMR2 prescale value and enable Timer2 by writing to T2CON.
- 5. Configure the CCP1 module for PWM operation.

PWM Frequency	1.22 kHz	4.88 kHz	19.53 kHz	78.12kHz	156.3 kHz	208.3 kHz
Timer Prescaler (1, 4, 16)	16	4	1	1	1	1
PR2 Value	0xFFh	0xFFh	0xFFh	0x3Fh	0x1Fh	0x17h
Maximum Resolution (bits)	10	10	10	8	7	5.5

TABLE 8-3: EXAMPLE PWM FREQUENCIES AND RESOLUTIONS AT 20 MHz

TABLE 8-4: REGISTERS ASSOCIATED WITH CAPTURE, COMPARE AND TIMER1

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value or POR, BC		Valu all o Res	ther
0Bh,8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 00	0x	0000	000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 00	00	0000	0000
0Dh	PIR2	—		_	—	—	—	_	CCP2IF		- 0		0
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 00	00	0000	0000
8Dh	PIE2	—	_	_	_	_	—	_	CCP2IE		- 0		0
87h	TRISC	PORTC Data Direction Register								1111 11	11	1111	1111
0Eh	TMR1L	Holding R	Holding Register for the Least Significant Byte of the 16-bit TMR1 Register							xxxx xx	xx	uuuu	uuuu
0Fh	TMR1H	Holding R	Holding Register for the Most Significant Byte of the 16-bit TMR1 Register							xxxx xx	xx	uuuu	uuuu
10h	T1CON	—	—	T1CKPS1	T1CKPS0	T1OSCEN	T1SYNC	TMR1CS	TMR1ON	00 00	00	uu	uuuu
15h	CCPR1L	Capture/C	Capture/Compare/PWM Register 1 (LSB)							xxxx xx	xx	uuuu	uuuu
16h	CCPR1H	Capture/C	Capture/Compare/PWM Register 1 (MSB)							xxxx xx	xx	uuuu	uuuu
17h	CCP1CON	—	—	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00 00	00	00	0000
1Bh	CCPR2L	Capture/C	compare/F	PWM Regis	ter 2 (LSB)					xxxx xx	xx	uuuu	uuuu
1Ch	CCPR2H	Capture/C	compare/F	PWM Regis	ter 2 (MSB)					xxxx xx	xx	uuuu	uuuu
1Dh	CCP2CON	—		CCP2X	CCP2Y	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 00	00	00	0000

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used by Capture and Timer1.

Note 1: The PSP is not implemented on 28-pin devices; always maintain these bits clear.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value POR, I		all o	e on other sets
0Bh,8Bh, 10Bh, 18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 (000x	0000	000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000	0000	0000	0000
0Dh	PIR2	—	_	_	_	_	_	_	CCP2IF		0		0
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000	0000	0000	0000
8Dh	PIE2	—	_	_	_	_	_	_	CCP2IE		0		0
87h	TRISC	PORTC D	PORTC Data Direction Register								1111	1111	1111
11h	TMR2	Timer2 M	odule's Reg	ister						0000	0000	0000	0000
92h	PR2	Timer2 M	odule's Peri	od Register	•					1111 :	1111	1111	1111
12h	T2CON	_	TOUTPS3	TOUTPS2	TOUTPS1	TOUTPS0	TMR2ON	T2CKPS1	T2CKPS0	-000 (0000	-000	0000
15h	CCPR1L	Capture/C	Capture/Compare/PWM Register 1 (LSB)								xxxx	uuuu	uuuu
16h	CCPR1H	Capture/C	Capture/Compare/PWM Register 1 (MSB)								xxxx	uuuu	uuuu
17h	CCP1CON	_	—	CCP1X	CCP1Y	CCP1M3	CCP1M2	CCP1M1	CCP1M0	00 (0000	00	0000
1Bh	CCPR2L	Capture/C	Capture/Compare/PWM Register 2 (LSB)								xxxx	uuuu	นนนบ
1Ch	CCPR2H	Capture/C	Compare/PV	VM Registe	r 2 (MSB)					XXXX X	xxxx	uuuu	uuuu
1Dh	CCP2CON	_	_	CCP2X	CCP2Y	CCP2M3	CCP2M2	CCP2M1	CCP2M0	00 (0000	00	0000

	TABLE 8-5:	REGISTERS ASSOCIATED WITH PWM AND TIMER2
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 $\label{eq:logend: Legend: x = unknown, u = unchanged, - = unimplemented, read as `0'. Shaded cells are not used by PWM and Timer2.$

Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices; always maintain these bits clear.

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NOTES:

9.0 MASTER SYNCHRONOUS SERIAL PORT (MSSP) MODULE

9.1 Master SSP (MSSP) Module Overview

The Master Synchronous Serial Port (MSSP) module is a serial interface, useful for communicating with other peripheral or microcontroller devices. These peripheral devices may be serial EEPROMs, shift registers, display drivers, A/D converters, etc. The MSSP module can operate in one of two modes:

- Serial Peripheral Interface (SPI)
- Inter-Integrated Circuit (I²C)
 - Full Master mode
 - Slave mode (with general address call)

The I^2C interface supports the following modes in hardware:

- Master mode
- Multi-Master mode
- Slave mode

9.2 Control Registers

The MSSP module has three associated registers. These include a status register (SSPSTAT) and two control registers (SSPCON and SSPCON2). The use of these registers and their individual configuration bits differ significantly, depending on whether the MSSP module is operated in SPI or I^2C mode.

Additional details are provided under the individual sections.

9.3 SPI Mode

The SPI mode allows 8 bits of data to be synchronously transmitted and received simultaneously. All four modes of SPI are supported. To accomplish communication, typically three pins are used:

- Serial Data Out (SDO) RC5/SDO
- Serial Data In (SDI) RC4/SDI/SDA
- Serial Clock (SCK) RC3/SCK/SCL

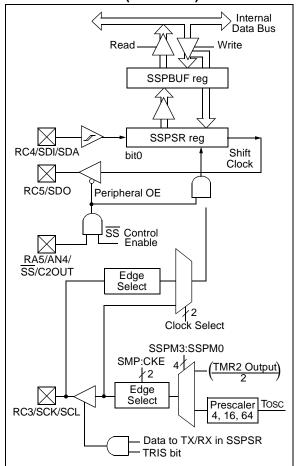
Additionally, a fourth pin may be used when in a Slave mode of operation:

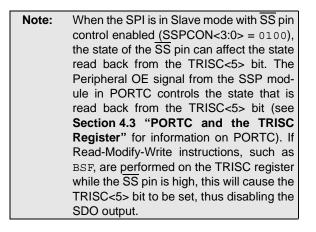
Slave Select (SS) – RA5/AN4/SS/C2OUT

Figure 9-1 shows the block diagram of the MSSP module when operating in SPI mode.

FIGURE 9-1:

MSSP BLOCK DIAGRAM (SPI MODE)





9.3.1 REGISTERS

The MSSP module has four registers for SPI mode operation. These are:

- MSSP Control Register (SSPCON)
- MSSP Status Register (SSPSTAT)
- Serial Receive/Transmit Buffer Register (SSPBUF)
- MSSP Shift Register (SSPSR) Not directly accessible

SSPCON and SSPSTAT are the control and status registers in SPI mode operation. The SSPCON register is readable and writable. The lower six bits of the SSPSTAT are read-only. The upper two bits of the SSPSTAT are read/write. SSPSR is the shift register used for shifting data in or out. SSPBUF is the buffer register to which data bytes are written to or read from.

In receive operations, SSPSR and SSPBUF together create a double-buffered receiver. When SSPSR receives a complete byte, it is transferred to SSPBUF and the SSPIF interrupt is set.

During transmission, the SSPBUF is not doublebuffered. A write to SSPBUF will write to both SSPBUF and SSPSR.

REGISTER 9-1: SSPSTAT: MSSP STATUS REGISTER (SPI MODE) (ADDRESS 94h)

	R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0
	SMP	CKE	D/A	Р	S	R/W	UA	BF
	bit 7							bit 0
bit 7	SMP: Sam	ple bit						
	SPI Master							
		ata sampled ata sampled						
	SPI Slave	•	at midule 0	i uala oulpu	t unie			
		be cleared v	vhen SPI is	used in Slav	/e mode.			
bit 6		Clock Select						
	1 = Transm	nit occurs on	transition fr	om active to	ldle clock s	state		
	0 = Transm	nit occurs on	transition fr	om Idle to a	ctive clock s	state		
	Note:	Polarity of c	clock state is	s set by the	CKP bit (SS	PCON1<4>).	
bit 5	D/A: Data/	Address bit						
	Used in I ² C	c mode only.						
bit 4	P: Stop bit							
	Used in I ² C	mode only.	This bit is cle	ared when t	he MSSP m	odule is disa	bled, SSPE	N is cleared.
bit 3	S: Start bit							
	Used in I ² C	c mode only.						
bit 2	R/W: Read	Write bit inf	ormation					
	Used in I ² C	c mode only.						
bit 1	UA: Update	e Address bi	it					
	Used in I ² C	c mode only.						
bit 0	BF: Buffer	Full Status b	oit (Receive	mode only)				
		e complete,						
	0 = Receiv	e not comple	ete, SSPBU	F is empty				
	Legend:							
	R = Reada	ble bit	W = W	ritable bit	U = Unin	nplemented	bit, read as	'0'
	- n = Value	at POR	'1' = B	it is set	'0' = Bit i	s cleared	x = Bit is ι	Inknown

	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0
	bit 7						1	bit 0
bit 7	WCOL: W	rite Collision	Detect bit (1	ransmit mo	de only)			
		SPBUF regi d in software Ilision		en while it is	s still transm	nitting the p	revious wor	d. (Must be
bit 6	SSPOV: R	eceive Over	flow Indicato	or bit				
	SPI Slave	mode:						
	of ove must r	byte is rece rflow, the da ead the SSF d in software erflow	ita in SSPSF PBUF, even i	R is lost. Ov	erflow can o	only occur ir	n Slave mod	e. The user
	Note:		mode, the on) is initiated					eption (and
bit 5	SSPEN: S	ynchronous	Serial Port E	nable bit				
		es serial port es serial port					ial port pins	
	Note:	When enab	oled, these p	ins must be	properly co	nfigured as	input or outp	out.
bit 4	CKP: Cloc	k Polarity Se	elect bit					
		ate for clock ate for clock	•					
bit 3-0	SSPM3:SS	SPM0: Syncl	nronous Seri	ial Port Mod	le Select bits	3		
	SSPM3:SSPM0: Synchronous Serial Port Mode Select bits 0101 = SPI Slave mode, clock = SCK pin. <u>SS</u> pin control disabled. <u>SS</u> can be used as I/O pin 0100 = SPI Slave mode, clock = SCK pin. <u>SS</u> pin control enabled. 0011 = SPI Master mode, clock = TMR2 output/2 0010 = SPI Master mode, clock = Fosc/64 0001 = SPI Master mode, clock = Fosc/16 0000 = SPI Master mode, clock = Fosc/4							
	Note:	Bit combina I ² C mode c	ations not sp only.	ecifically lis	ted here are	e either rese	erved or imp	lemented in

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented	bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

9.3.2 OPERATION

When initializing the SPI, several options need to be specified. This is done by programming the appropriate control bits (SSPCON<5:0> and SSPSTAT<7:6>). These control bits allow the following to be specified:

- Master mode (SCK is the clock output)
- Slave mode (SCK is the clock input)
- Clock Polarity (Idle state of SCK)
- Data Input Sample Phase (middle or end of data output time)
- Clock Edge (output data on rising/falling edge of SCK)
- Clock Rate (Master mode only)
- · Slave Select mode (Slave mode only)

The MSSP consists of a transmit/receive shift register (SSPSR) and a buffer register (SSPBUF). The SSPSR shifts the data in and out of the device, MSb first. The SSPBUF holds the data that was written to the SSPSR until the received data is ready. Once the eight bits of data have been received, that byte is moved to the SSPBUF register. Then, the Buffer Full detect bit, BF (SSPSTAT<0>), and the interrupt flag bit, SSPIF, are set. This double-buffering of the received data (SSPBUF) allows the next byte to start reception before reading the data that was just received. Any write to the

SSPBUF register during transmission/reception of data will be ignored and the write collision detect bit, WCOL (SSPCON<7>), will be set. User software must clear the WCOL bit so that it can be determined if the following write(s) to the SSPBUF register completed successfully.

When the application software is expecting to receive valid data, the SSPBUF should be read before the next byte of data to transfer is written to the SSPBUF. Buffer Full bit, BF (SSPSTAT<0>), indicates when SSPBUF has been loaded with the received data (transmission is complete). When the SSPBUF is read, the BF bit is cleared. This data may be irrelevant if the SPI is only a transmitter. Generally, the MSSP interrupt is used to determine when the transmission/reception has completed. The SSPBUF must be read and/or written. If the interrupt method is not going to be used, then software polling can be done to ensure that a write collision does not occur. Example 9-1 shows the loading of the SSPBUF (SSPSR) for data transmission.

The SSPSR is not directly readable or writable and can only be accessed by addressing the SSPBUF register. Additionally, the MSSP Status register (SSPSTAT) indicates the various status conditions.

EXAMPLE 9-1: LOADING THE SSPBUF (SSPSR) REGISTER

LOOP	BTFSS	SSPSTAT, BF	;Has data been received(transmit complete)?
	BRA	LOOP	; No
	MOVF	SSPBUF, W	;WREG reg = contents of SSPBUF
	MOVWF	RXDATA	;Save in user RAM, if data is meaningful
	MOVF	TXDATA, W	;W reg = contents of TXDATA
	MOVWF	SSPBUF	;New data to xmit

9.3.3 ENABLING SPI I/O

To enable the serial port, SSP Enable bit, SSPEN (SSPCON<5>), must be set. To reset or reconfigure SPI mode, clear the SSPEN bit, re-initialize the SSPCON registers and then set the SSPEN bit. This configures the SDI, SDO, SCK and SS pins as serial port pins. For the pins to behave as the serial port function, some must have their data direction bits (in the TRIS register) appropriately programmed. That is:

- SDI is automatically controlled by the SPI module
- SDO must have TRISC<5> bit cleared
- SCK (Master mode) must have TRISC<3> bit cleared
- SCK (Slave mode) must have TRISC<3> bit set
- SS must have TRISC<4> bit set

Any serial port function that is not desired may be overridden by programming the corresponding data direction (TRIS) register to the opposite value.

9.3.4 TYPICAL CONNECTION

Figure 9-2 shows a typical connection between two microcontrollers. The master controller (Processor 1) initiates the data transfer by sending the SCK signal. Data is shifted out of both shift registers on their programmed clock edge and latched on the opposite edge of the clock. Both processors should be programmed to the same Clock Polarity (CKP), then both controllers would send and receive data at the same time. Whether the data is meaningful (or dummy data) depends on the application software. This leads to three scenarios for data transmission:

- Master sends data Slave sends dummy data
- Master sends data Slave sends data
- Master sends dummy data Slave sends data

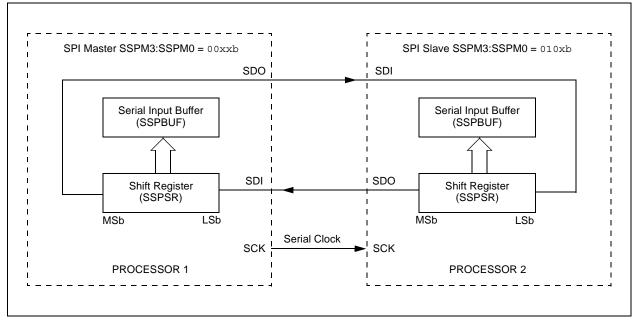


FIGURE 9-2: SPI MASTER/SLAVE CONNECTION

9.3.5 MASTER MODE

The master can initiate the data transfer at any time because it controls the SCK. The master determines when the slave (Processor 2, Figure 9-2) is to broadcast data by the software protocol.

In Master mode, the data is transmitted/received as soon as the SSPBUF register is written to. If the SPI is only going to receive, the SDO output could be disabled (programmed as an input). The SSPSR register will continue to shift in the signal present on the SDI pin at the programmed clock rate. As each byte is received, it will be loaded into the SSPBUF register as if a normal received byte (interrupts and status bits appropriately set). This could be useful in receiver applications as a "Line Activity Monitor" mode.

The clock polarity is selected by appropriately programming the CKP bit (SSPCON<4>). This then, would give waveforms for SPI communication as shown in Figure 9-3, Figure 9-5 and Figure 9-6, where the MSB is transmitted first. In Master mode, the SPI clock rate (bit rate) is user programmable to be one of the following:

- Fosc/4 (or Tcy)
- Fosc/16 (or 4 Tcy)
- Fosc/64 (or 16 Tcy)
- Timer2 output/2

This allows a maximum data rate (at 40 MHz) of 10.00 Mbps.

Figure 9-3 shows the waveforms for Master mode. When the CKE bit is set, the SDO data is valid before there is a clock edge on SCK. The change of the input sample is shown based on the state of the SMP bit. The time when the SSPBUF is loaded with the received data is shown.

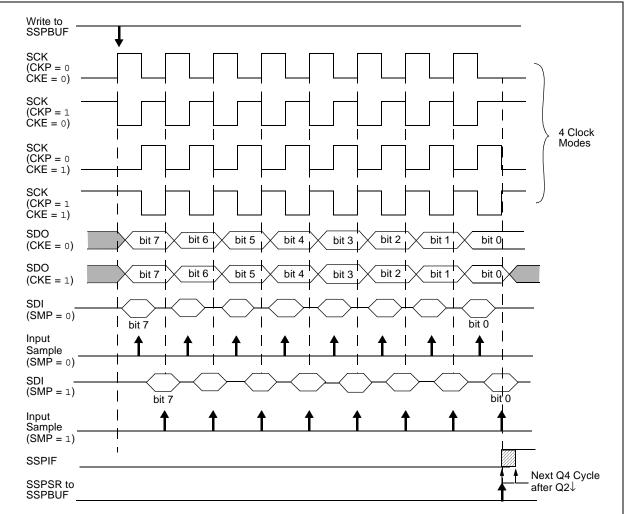


FIGURE 9-3: SPI MODE WAVEFORM (MASTER MODE)

9.3.6 SLAVE MODE

In Slave mode, the data is transmitted and received as the external clock pulses appear on SCK. When the last bit is latched, the SSPIF interrupt flag bit is set.

While in Slave mode, the external clock is supplied by the external clock source on the SCK pin. This external clock must meet the minimum high and low times as specified in the electrical specifications.

While in Sleep mode, the slave can transmit/receive data. When a byte is received, the device will wake-up from Sleep.

9.3.7 SLAVE SELECT SYNCHRONIZATION

The \overline{SS} pin allows a Synchronous Slave mode. The SPI must be in Slave mode with \overline{SS} pin control enabled (SSPCON<3:0> = 04h). The pin must not be driven low for the \overline{SS} pin to function as an input. The data latch must be high. When the \overline{SS} pin is low, transmission and reception are enabled and the SDO pin is driven. When

the \overline{SS} pin goes high, the SDO pin is no longer driven even if in the middle of a transmitted byte and becomes a floating output. External pull-up/pull-down resistors may be desirable, depending on the application.

Note 1:	When the SPI is in Slave mode with \overline{SS} pin
	control enabled (SSPCON<3: $\underline{0}$ = 0100),
	the SPI module will reset if the \overline{SS} pin is set
	to VDD.

 If the SPI is used in Slave Mode with CKE set, then the SS pin control must be enabled.

When the SPI module resets, the bit counter is forced to '0'. This can be done by either forcing the \overline{SS} pin to a high level or clearing the SSPEN bit.

To emulate two-wire communication, the SDO pin can be connected to the SDI pin. When the SPI needs to operate as a receiver, the SDO pin can be configured as an input. This disables transmissions from the SDO. The SDI can always be left as an input (SDI function) since it cannot create a bus conflict.

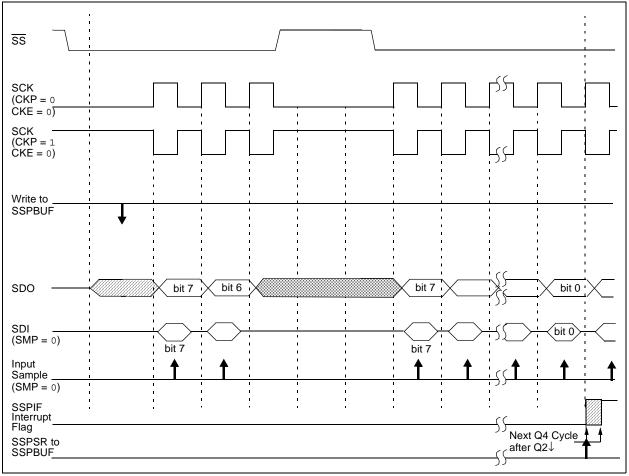


FIGURE 9-4: SLAVE SYNCHRONIZATION WAVEFORM



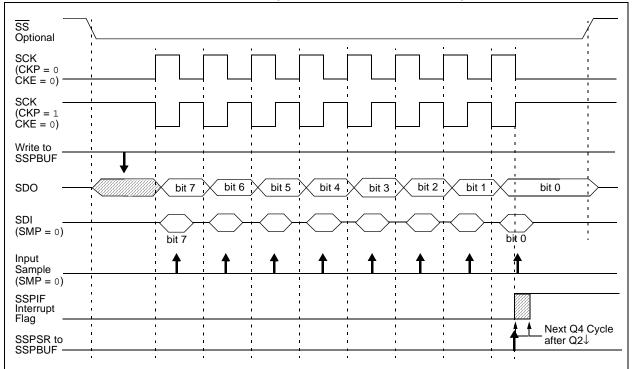
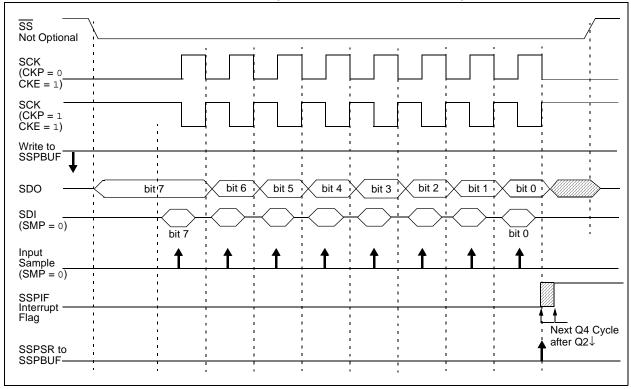


FIGURE 9-6: SPI MODE WAVEFORM (SLAVE MODE WITH CKE = 1)



9.3.8 SLEEP OPERATION

In Master mode, all module clocks are halted and the transmission/reception will remain in that state until the device wakes from Sleep. After the device returns to normal mode, the module will continue to transmit/ receive data.

In Slave mode, the SPI Transmit/Receive Shift register operates asynchronously to the device. This allows the device to be placed in Sleep mode and data to be shifted into the SPI Transmit/Receive Shift register. When all 8 bits have been received, the MSSP interrupt flag bit will be set and if enabled, will wake the device from Sleep.

9.3.9 EFFECTS OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

9.3.10 BUS MODE COMPATIBILITY

Table 9-1 shows the compatibility between the standard SPI modes and the states of the CKP and CKE control bits.

TABLE 9-1: SPI BUS MODES

Standard SPI Mode	Control Bits State				
Terminology	СКР	CKE			
0, 0	0	1			
0, 1	0	0			
1, 0	1	1			
1, 1	1	0			

There is also a SMP bit which controls when the data is sampled.

Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Valu POR,		Valu all o Res	
INTCON	GIE/ GIEH	PEIE/ GIEL	TMR0IE	INTOIE	RBIE	TMR0IF	INTOIF	RBIF	0000	000x	0000	000u
PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000	0000	0000	0000
PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000	0000	0000	0000
TRISC	PORTC D	Data Direc	tion Regis	ter					1111	1111	1111	1111
SSPBUF	Synchron	ous Seria	I Port Rec	eive Buffe	er/Transmit	Register			xxxx	xxxx	uuuu	uuuu
SSPCON	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0	0000	0000	0000	0000
TRISA	—	PORTA Data Direction Register						11	1111	11	1111	
SSPSTAT	SMP	CKE	D/Ā	Р	S	R/W	UA	BF	0000	0000	0000	0000

TABLE 9-2: REGISTERS ASSOCIATED WITH SPI OPERATION

 $\label{eq:Legend: x = unknown, u = unchanged, - = unimplemented, read as `0'. Shaded cells are not used by the MSSP in SPI mode.$

Note 1: The PSPIF, PSPIE and PSPIP bits are reserved on 28-pin devices; always maintain these bits clear.

9.4 I²C Mode

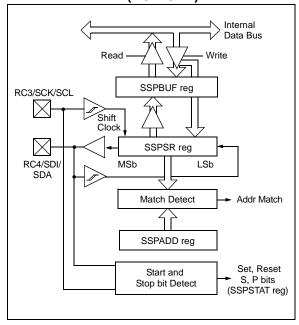
The MSSP module in I^2C mode fully implements all master and slave functions (including general call support) and provides interrupts on Start and Stop bits in hardware to determine a free bus (multi-master function). The MSSP module implements the standard mode specifications, as well as 7-bit and 10-bit addressing.

Two pins are used for data transfer:

- Serial clock (SCL) RC3/SCK/SCL
- Serial data (SDA) RC4/SDI/SDA

The user must configure these pins as inputs or outputs through the TRISC<4:3> bits.

FIGURE 9-7: MSSP BLOCK DIAGRAM (I²C MODE)



9.4.1 REGISTERS

The MSSP module has six registers for $\mathsf{I}^2\mathsf{C}$ operation. These are:

- MSSP Control Register (SSPCON)
- MSSP Control Register 2 (SSPCON2)
- MSSP Status Register (SSPSTAT)
- Serial Receive/Transmit Buffer Register (SSPBUF)
- MSSP Shift Register (SSPSR) Not directly accessible
- MSSP Address Register (SSPADD)

SSPCON, SSPCON2 and SSPSTAT are the control and status registers in I^2C mode operation. The SSPCON and SSPCON2 registers are readable and writable. The lower six bits of the SSPSTAT are read-only. The upper two bits of the SSPSTAT are read/write.

SSPSR is the shift register used for shifting data in or out. SSPBUF is the buffer register to which data bytes are written to or read from.

SSPADD register holds the slave device address when the SSP is configured in I^2C Slave mode. When the SSP is configured in Master mode, the lower seven bits of SSPADD act as the baud rate generator reload value.

In receive operations, SSPSR and SSPBUF together create a double-buffered receiver. When SSPSR receives a complete byte, it is transferred to SSPBUF and the SSPIF interrupt is set.

During transmission, the SSPBUF is not doublebuffered. A write to SSPBUF will write to both SSPBUF and SSPSR.

	R/W-0	R/W-0	R-0	R-0	R-0	R-0	R-0	R-0	
	SMP	CKE	D/A	Р	S	R/W	UA	BF	
	bit 7							bit 0	
bit 7	SMP: Slew	Rate Contro	ol bit						
		or Slave mod							
		ate control dis ate control er					1 MHZ)		
bit 6		sus Select bit		gn opeed n		12)			
		or Slave mod							
		SMBus spece SMBus spe							
bit 5	D/A: Data/	Address bit							
	In Master r	<u>node:</u>							
	Reserved.	odo							
	In Slave me 1 = Indicate	<u>ode:</u> es that the la	st byte rece	ived or tran	smitted was	data			
		es that the la							
bit 4	P: Stop bit								
		es that a Stop		en detectec	l last				
	•	t was not det			000551				
hit O	Note:	This bit is cl	eared on R	eset and wr	nen SSPEN	is cleared.			
bit 3	S: Start bit								
	 1 = Indicates that a Start bit has been detected last 0 = Start bit was not detected last 								
	Note:	This bit is cl		eset and wh	nen SSPEN	is cleared.			
bit 2	R/W: Read	I/Write bit info	ormation (I ²	C mode onl	ly)				
	In Slave me	ode:							
	1 = Read								
	0 = Write Note:	This hit hold	he the $P\overline{M}$	hit informat	ion following	g the last add	droce match	This hit is	
	Note.					Start bit, Stop			
	In Master r	-							
		nit is in progre							
		nit is not in pr	-				indiaata if th		
	Note:	in Idle mode		, KOEN, FE	IN, ROEN O	r ACKEN will		10 101337 15	
bit 1	-	e Address (1							
		es that the us is does not n			e address in	the SSPADE	D register		
bit 0	BF: Buffer Full Status bit								
	<u>In Transmit mode:</u> 1 = Receive complete, SSPBUF is full 0 = Receive not complete, SSPBUF is empty								
		-	te, 55PBU	- is empty					
	In Receive mode: 1 = Data Transmit in progress (does not include the \overline{ACK} and Stop bits), SSPBUF is full 0 = Data Transmit complete (does not include the \overline{ACK} and Stop bits), SSPBUF is empty								
	Legend:	hla hit	10/ 10/	ritabla kit	-:ما ا	nnlomontod	hit road as '	o'	
	R = Reada		vv = vv	ritable bit	U = Uhin	nplemented	bit, read as	U	

'1' = Bit is set

'0' = Bit is cleared

- n = Value at POR

x = Bit is unknown

REGISTER 9-4:	SSPCON	I: MSSP CO		EGISTER	1 (I ² C MO	DE) (ADD	RESS 14h)	
	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
	WCOL	SSPOV	SSPEN	CKP	SSPM3	SSPM2	SSPM1	SSPM0
	bit 7							bit 0
bit 7	WCOL: W	rite Collision	Detect bit					
	1 = A writ a tran 0 = No co <u>In Slave T</u> 1 = The S	ransmit mode SPBUF regis	BUF registe e started. (N <u>e:</u> ster is writte	Aust be clea	ared in softw	are.)		
	0 = No co <u>In Receive</u>	ed in software Ilision <u>e mode (Maste</u> don't care" bit	ér or Slave	modes):				
bit 6	SSPOV: R	eceive Overf	low Indicato	r bit				
	-	e is received d in software erflow		SPBUF reg	ister is still h	nolding the p	previous byt	e. (Must be
	This is a "o	don't care" bit	in Transmit	mode.				
bit 5	1 = Enable	ynchronous S es the serial p es the serial p	ort and con	figures the			ne serial por	pins
	Note:	•		•	ns must be p		gured as inp	ut or output.
bit 4	CKP: SCK	Release Co	ntrol bit					
	<u>In Slave m</u> 1 = Releas 0 = Holds In Master	i <u>ode:</u> se clock clock low (clo		(Used to er	nsure data se	etup time.)		
bit 3-0	$1111 = I^{2}(1)$ $1110 = I^{2}(1)$ $1011 = I^{2}(1)$ $1000 = I^{2}(1)$ $0111 = I^{2}(1)$	SPM0: Synch Slave mode Slave mode Firmware C Master mode Slave mode Slave mode	e, 10-bit add e, 7-bit addre ontrolled Ma le, clock = F e, 10-bit add e, 7-bit addre	ress with Sta ass with Sta aster mode FOSC/(4 * (S ress ess	tart and Stop Irt and Stop I (Slave Idle) SPADD + 1)	bit interrup bit interrupts)	enabled	emented in
	NOLE.	SPI mode o	-	someany lis	sted here are			
	Legend:							
	R = Reada	able bit	W = W	ritable bit	U = Unim	plemented	bit, read as	'0'

'1' = Bit is set

'0' = Bit is cleared

- n = Value at POR

x = Bit is unknown

REGISTER 9-5:	SSPCON	2: MSSP CC		EGISTER 2	(I ² C MOD	E) (ADDF	RESS 91h))
	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0
	GCEN	ACKSTAT	ACKDT	ACKEN	RCEN	PEN	RSEN	SEN
	bit 7							bit 0
bit 7	GCEN: G	eneral Call En	able bit (Sla	ve mode only	()			
		e interrupt whe	•	call address	(0000h) is	received in	the SSPSF	2
bit 6	ACKSTAT	F: Acknowledg	e Status bit	(Master Tran	smit mode o	only)		
		wledge was n wledge was n						
bit 5	ACKDT: A	Acknowledge I	Data bit (Ma	ster Receive	mode only)			
	1 = Not A 0 = Ackno	cknowledge wledge						
	Note:	Value that w the end of a		itted when th	e user initia	tes an Ackı	nowledge s	equence at
bit 4	1 = Initiat	Acknowledge e Acknowledg natically clear	ge sequence	e on SDA ar				T data bit.
		owledge sequ						
bit 3	RCEN: Re	eceive Enable	bit (Master	mode only)				
	1 = Enabl 0 = Recei	es Receive m ve Idle	ode for I ² C					
bit 2	PEN: Stop	Condition Er	able bit (Ma	ster mode or	nly)			
		e Stop conditio condition Idle	on on SDA a	nd SCL pins.	Automatica	ally cleared	by hardwa	re.
bit 1	RSEN: Re	epeated Start	Condition Er	nabled bit (Ma	aster mode	only)		
		e Repeated Sta ated Start con		on SDA and	SCL pins. A	utomatically	y cleared by	hardware.
bit 0	SEN: Star	t Condition Er	abled/Streto	ch Enabled bi	t			
	In Master							
		e Start conditio condition Idle	on on SDA a	nd SCL pins.	Automatica	ally cleared	by hardwa	re.
	In Slave mode: 1 = Clock stretching is enabled for both slave transmit and slave receive (stretch enabled) 0 = Clock stretching is enabled for slave transmit only (PIC16F87X compatibility)							
	Legend:							
	R = Read	able bit	W = WI	ritable bit	U = Unimp	olemented	bit, read as	'0'
	- n = Value	e at POR	'1' = Bit	is set	'0' = Bit is	cleared	x = Bit is u	unknown

Note: For bits ACKEN, RCEN, PEN, RSEN, SEN: If the I²C module is not in the Idle mode, this bit may not be set (no spooling) and the SSPBUF may not be written (or writes to the SSPBUF are disabled).

9.4.2 OPERATION

The MSSP module functions are enabled by setting MSSP Enable bit, SSPEN (SSPCON<5>).

The SSPCON register allows control of the I^2C operation. Four mode selection bits (SSPCON<3:0>) allow one of the following I^2C modes to be selected:

- I²C Master mode, clock = OSC/4 (SSPADD + 1)
- I²C Slave mode (7-bit address)
- I²C Slave mode (10-bit address)
- I²C Slave mode (7-bit address) with Start and Stop bit interrupts enabled
- I²C Slave mode (10-bit address) with Start and Stop bit interrupts enabled
- I²C Firmware Controlled Master mode, slave is Idle

Selection of any I²C mode, with the SSPEN bit set, forces the SCL and SDA pins to be open-drain, provided these pins are programmed to inputs by setting the appropriate TRISC bits. To ensure proper operation of the module, pull-up resistors must be provided externally to the SCL and SDA pins.

9.4.3 SLAVE MODE

In Slave mode, the SCL and SDA pins must be configured as inputs (TRISC<4:3> set). The MSSP module will override the input state with the output data when required (slave-transmitter).

The I²C Slave mode hardware will always generate an interrupt on an address match. Through the mode select bits, the user can also choose to interrupt on Start and Stop bits

When an address is matched, or the data transfer after an address match is received, the hardware automatically will generate the Acknowledge (\overline{ACK}) pulse and load the SSPBUF register with the received value currently in the SSPSR register.

Any combination of the following conditions will cause the MSSP module not to give this ACK pulse:

- The buffer full bit, BF (SSPSTAT<0>), was set before the transfer was received.
- The overflow bit, SSPOV (SSPCON<6>), was set before the transfer was received.

In this case, the SSPSR register value is not loaded into the SSPBUF, but bit SSPIF (PIR1<3>) is set. The BF bit is cleared by reading the SSPBUF register, while bit SSPOV is cleared through software.

The SCL clock input must have a minimum high and low for proper operation. The high and low times of the I^2C specification, as well as the requirement of the MSSP module, are shown in timing parameter #100 and parameter #101.

9.4.3.1 Addressing

Once the MSSP module has been enabled, it waits for a Start condition to occur. Following the Start condition, the 8 bits are shifted into the SSPSR register. All incoming bits are sampled with the rising edge of the clock (SCL) line. The value of register SSPSR<7:1> is compared to the value of the SSPADD register. The address is compared on the falling edge of the eighth clock (SCL) pulse. If the addresses match, and the BF and SSPOV bits are clear, the following events occur:

- 1. The SSPSR register value is loaded into the SSPBUF register.
- 2. The Buffer Full bit, BF, is set.
- 3. An ACK pulse is generated.
- MSSP Interrupt Flag bit, SSPIF (PIR1<3>), is set (interrupt is generated if enabled) on the falling edge of the ninth SCL pulse.

In 10-bit Address mode, two address bytes need to be received by the slave. The five Most Significant bits (MSbs) of the first address byte specify if this is a 10-bit address. Bit R/W (SSPSTAT<2>) must specify a write so the slave device will receive the second address byte. For a 10-bit address, the first byte would equal '11110 A9 A8 0', where 'A9' and 'A8' are the two MSbs of the address. The sequence of events for 10-bit address is as follows, with steps 7 through 9 for the slave-transmitter:

- 1. Receive first (high) byte of address (bits SSPIF, BF and bit UA (SSPSTAT<1>) are set).
- 2. Update the SSPADD register with second (low) byte of address (clears bit UA and releases the SCL line).
- 3. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.
- 4. Receive second (low) byte of address (bits SSPIF, BF and UA are set).
- Update the SSPADD register with the first (high) byte of address. If match releases SCL line, this will clear bit UA.
- 6. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.
- 7. Receive Repeated Start condition.
- 8. Receive first (high) byte of address (bits SSPIF and BF are set).
- 9. Read the SSPBUF register (clears bit BF) and clear flag bit SSPIF.

9.4.3.2 Reception

When the R/\overline{W} bit of the address byte is clear and an address match occurs, the R/\overline{W} bit of the SSPSTAT register is cleared. The received address is loaded into the SSPBUF register and the SDA line is held low (ACK).

When the address byte overflow condition exists, then the No Acknowledge (ACK) pulse is given. An overflow condition is defined as either bit BF (SSPSTAT<0>) is set or bit SSPOV (SSPCON<6>) is set.

An MSSP interrupt is generated for each data transfer byte. Flag bit SSPIF (PIR1<3>) must be cleared in software. The SSPSTAT register is used to determine the status of the byte.

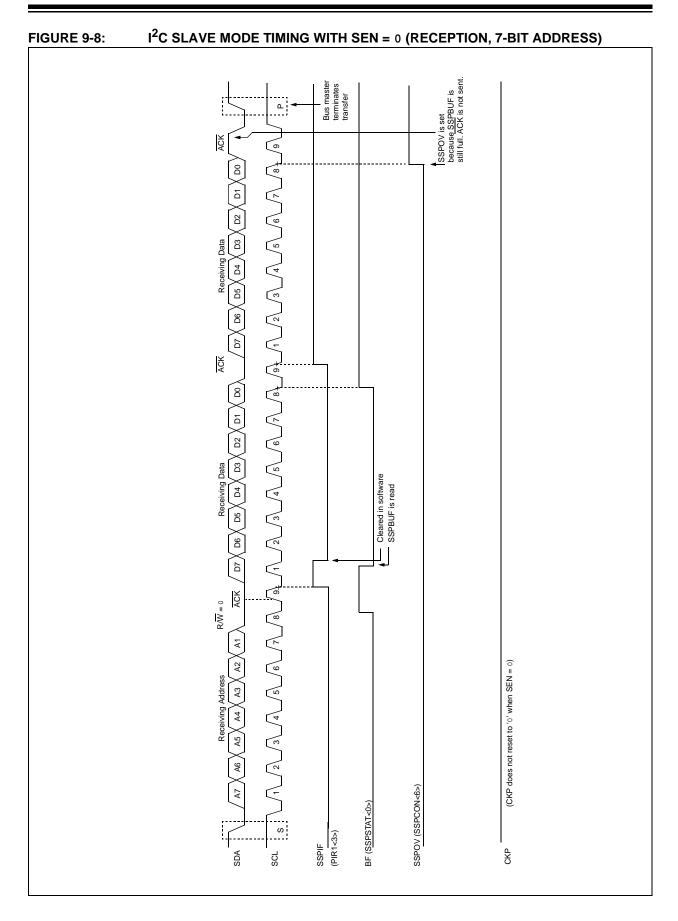
If SEN is enabled (SSPCON<0> = 1), RC3/SCK/SCL will be held low (clock stretch) following each data transfer. The clock must be released by setting bit CKP (SSPCON<4>). See **Section 9.4.4** "**Clock Stretching**" for more detail.

9.4.3.3 Transmission

When the $R\overline{W}$ bit of the incoming address byte is set and an address match occurs, the $R\overline{W}$ bit of the SSPSTAT register is set. The received address is loaded into the SSPBUF register. The ACK pulse will be sent on the ninth bit and pin RC3/SCK/SCL is held low regardless of SEN (see **Section 9.4.4 "Clock Stretching"** for more detail). By stretching the clock, the master will be unable to assert another clock pulse until the slave is done preparing the transmit data. The transmit data must be loaded into the SSPBUF register, which also loads the SSPSR register. Then pin RC3/SCK/SCL should be enabled by setting bit CKP (SSPCON<4>). The eight data bits are shifted out on the falling edge of the SCL input. This ensures that the SDA signal is valid during the SCL high time (Figure 9-9).

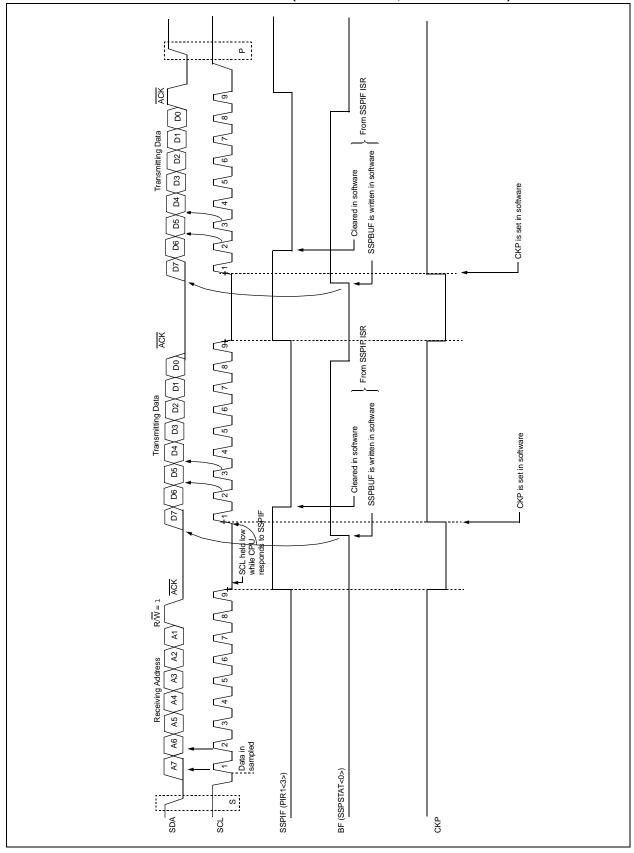
The ACK pulse from the master-receiver is latched on the rising edge of the ninth SCL input pulse. If the SDA line is high (not ACK), then the data transfer is complete. In this case, when the ACK is latched by the slave, the slave logic is reset (resets SSPSTAT register) and the slave monitors for another occurrence of the Start bit. If the SDA line was low (ACK), the next transmit data must be loaded into the SSPBUF register. Again, pin RC3/SCK/SCL must be enabled by setting bit CKP.

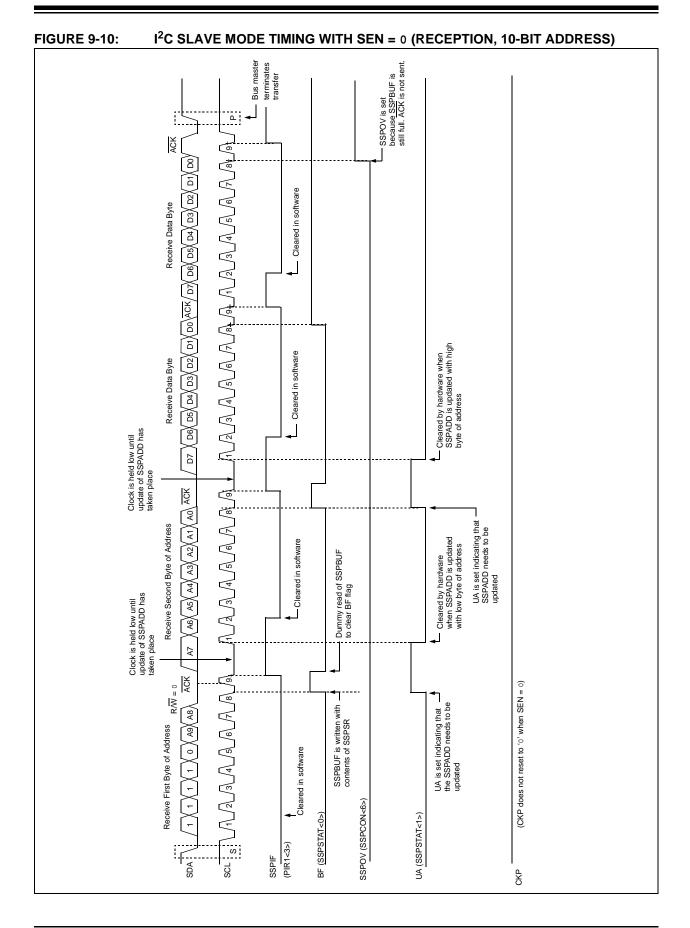
An MSSP interrupt is generated for each data transfer byte. The SSPIF bit must be cleared in software and the SSPSTAT register is used to determine the status of the byte. The SSPIF bit is set on the falling edge of the ninth clock pulse.

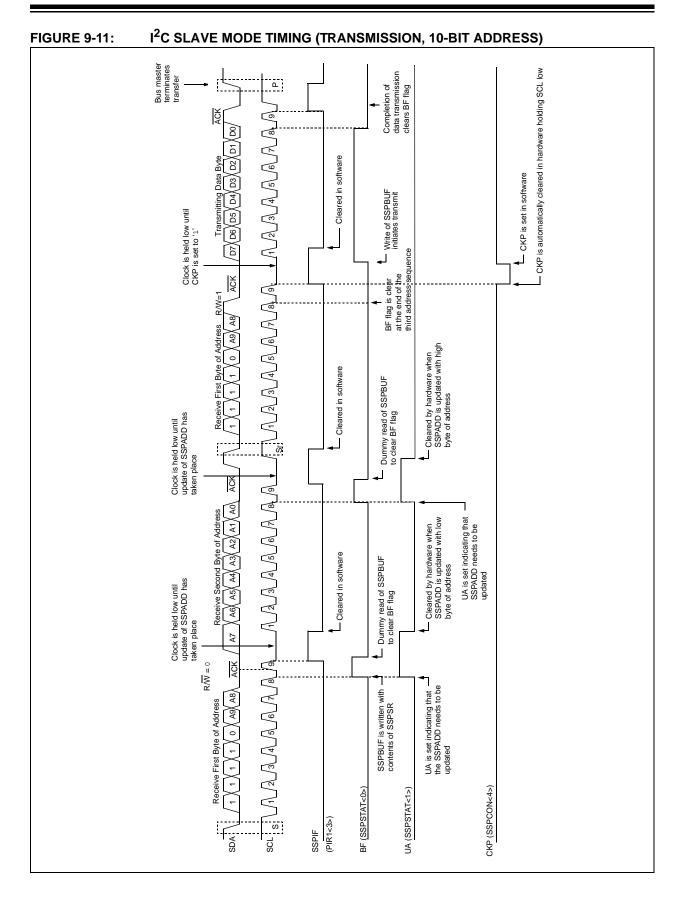




I²C SLAVE MODE TIMING (TRANSMISSION, 7-BIT ADDRESS)







9.4.4 CLOCK STRETCHING

Both 7 and 10-bit Slave modes implement automatic clock stretching during a transmit sequence.

The SEN bit (SSPCON2<0>) allows clock stretching to be enabled during receives. Setting SEN will cause the SCL pin to be held low at the end of each data receive sequence.

9.4.4.1 Clock Stretching for 7-bit Slave Receive Mode (SEN = 1)

In 7-bit Slave Receive mode, <u>on the falling edge of the</u> ninth clock at the end of the ACK sequence, if the BF bit is set, the CKP bit in the SSPCON register is automatically cleared, forcing the SCL output to be held low. The CKP bit being cleared to '0' will assert the SCL line low. The CKP bit must be set in the user's ISR before reception is allowed to continue. By holding the SCL line low, the user has time to service the ISR and read the contents of the SSPBUF before the master device can initiate another receive sequence. This will prevent buffer overruns from occurring (see Figure 9-13).

- Note 1: If the user reads the contents of the SSPBUF before the falling edge of the ninth clock, thus clearing the BF bit, the CKP bit will not be cleared and clock stretching will not occur.
 - 2: The CKP bit can be set in software regardless of the state of the BF bit. The user should be careful to clear the BF bit in the ISR before the next receive sequence in order to prevent an overflow condition.

9.4.4.2 Clock Stretching for 10-bit Slave Receive Mode (SEN = 1)

In 10-bit Slave Receive mode, during the address sequence, clock stretching automatically takes place but CKP is not cleared. During this time, if the UA bit is set after the ninth clock, clock stretching is initiated. The UA bit is set after receiving the upper byte of the 10-bit address and following the receive of the second byte of the 10-bit address, with the R/W bit cleared to '0'. The release of the clock line occurs upon updating SSPADD. Clock stretching will occur on each data receive sequence as described in 7-bit mode.

Note: If the user polls the UA bit and clears it by updating the SSPADD register before the falling edge of the ninth clock occurs and if the user hasn't cleared the BF bit by reading the SSPBUF register before that time, then the CKP bit will still NOT be asserted low. Clock stretching, on the basis of the state of the BF bit, only occurs during a data sequence, not an address sequence.

9.4.4.3 Clock Stretching for 7-bit Slave Transmit Mode

7-bit Slave Transmit mode implements clock stretching by clearing the CKP bit after the falling edge of the ninth clock, if the BF bit is clear. This occurs regardless of the state of the SEN bit.

The user's ISR must set the CKP bit before transmission is allowed to continue. By holding the SCL line low, the user has time to service the ISR and load the contents of the SSPBUF before the master device can initiate another transmit sequence (see Figure 9-9).

Note 1:	If the user loads the contents of SSPBUF, setting the BF bit before the falling edge of the ninth clock, the CKP bit will not be cleared and clock stretching will not occur.
2:	The CKP bit can be set in software regardless of the state of the BF bit.

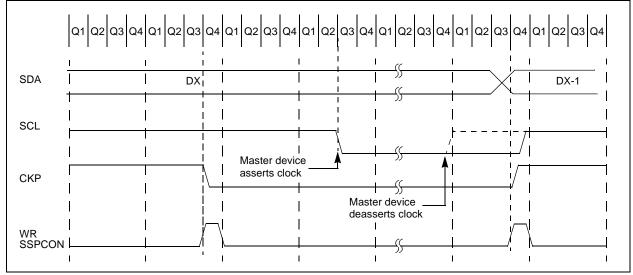
9.4.4.4 Clock Stretching for 10-bit Slave Transmit Mode

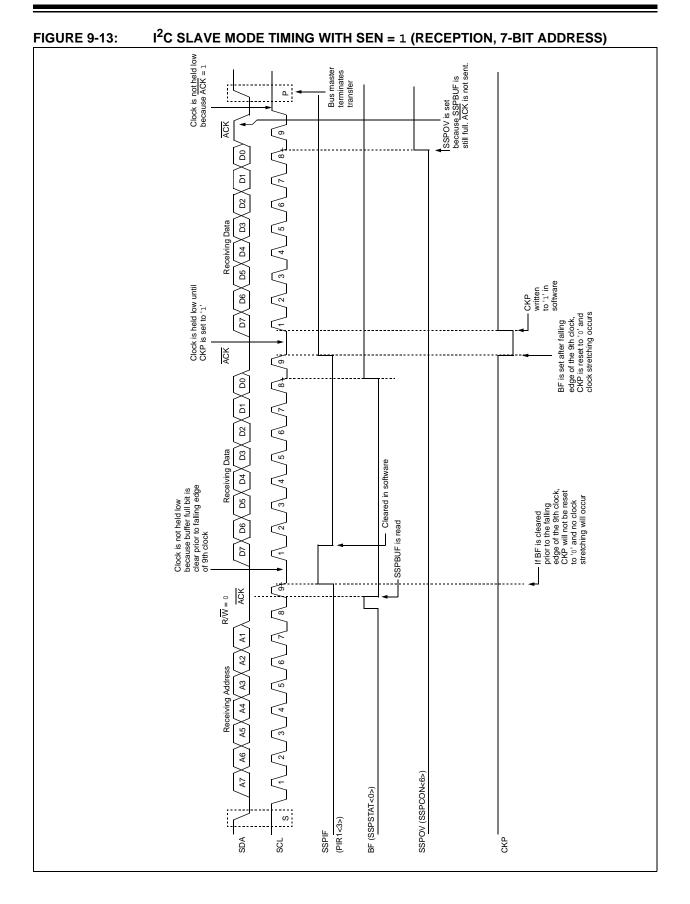
In 10-bit Slave Transmit mode, clock stretching is controlled during the first two address sequences by the state of the UA bit, just as it is in 10-bit Slave Receive mode. The first two addresses are followed by a third address sequence, which contains the high order bits of the 10-bit address and the R/W bit set to '1'. After the third address sequence is performed, the UA bit is not set, the module is now configured in Transmit mode and clock stretching is controlled by the BF flag as in 7-bit Slave Transmit mode (see Figure 9-11).

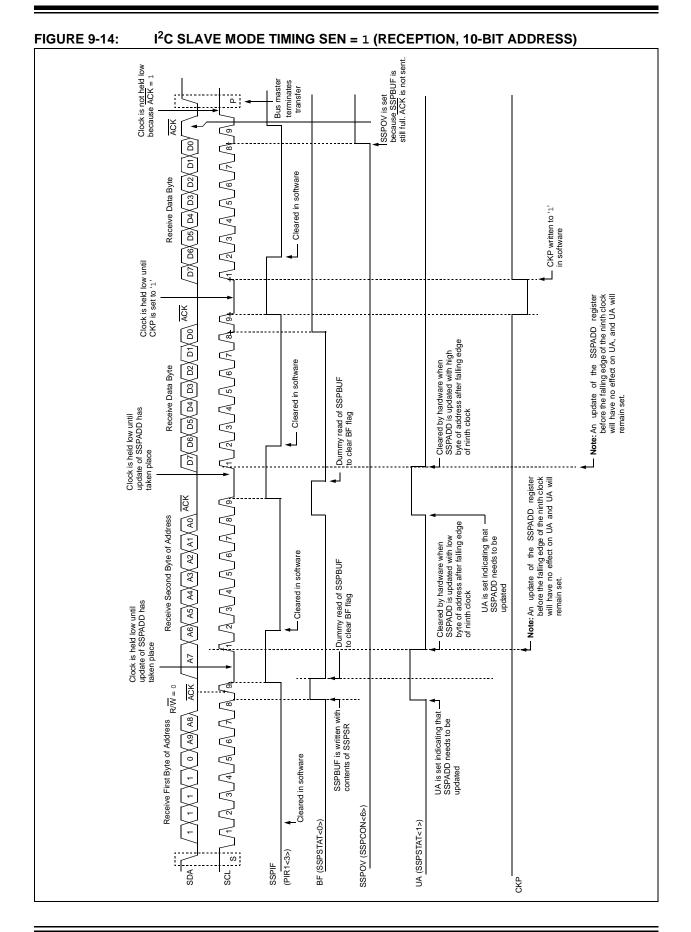
9.4.4.5 Clock Synchronization and the CKP Bit

When the CKP bit is cleared, the SCL output is forced to '0'; however, setting the CKP bit will not assert the SCL output low until the SCL output is already sampled low. Therefore, the CKP bit will not assert the SCL line until an external I^2C master device has already asserted the SCL line. The SCL output will remain low until the CKP bit is set and all other devices on the I^2C bus have deasserted SCL. This ensures that a write to the CKP bit will not violate the minimum high time requirement for SCL (see Figure 9-12).









9.4.5 GENERAL CALL ADDRESS SUPPORT

The addressing procedure for the I²C bus is such that the first byte after the Start condition usually determines which device will be the slave addressed by the master. The exception is the general call address which can address all devices. When this address is used, all devices should, in theory, respond with an Acknowledge.

The general call address is one of eight addresses reserved for specific purposes by the I^2C protocol. It consists of all '0's with R/W = 0.

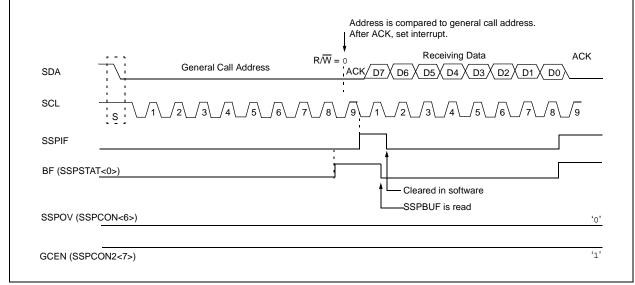
The general call address is recognized when the General Call Enable bit (GCEN) is enabled (SSPCON2<7> set). Following a Start bit detect, 8 bits are shifted into the SSPSR and the address is compared against the SSPADD. It is also compared to the general call address and fixed in hardware.

If the general call address matches, the SSPSR is transferred to the SSPBUF, the BF flag bit is set (eighth bit) and on the falling edge of the ninth bit (ACK bit), the SSPIF interrupt flag bit is set.

When the interrupt is serviced, the source for the interrupt can be checked by reading the contents of the SSPBUF. The value can be used to determine if the address was device specific or a general call address.

In 10-bit mode, the SSPADD is required to be updated for the second half of the address to match and the UA bit is set (SSPSTAT<1>). If the general call address is sampled when the GCEN bit is set, while the slave is configured in 10-bit Address mode, then the second half of the address is not necessary, the UA bit will not be set and the slave will begin receiving data after the Acknowledge (Figure 9-15).





9.4.6 MASTER MODE

Master mode is enabled by setting and clearing the appropriate SSPM bits in SSPCON and by setting the SSPEN bit. In Master mode, the SCL and SDA lines are manipulated by the MSSP hardware.

Master mode of operation is supported by interrupt generation on the detection of the Start and Stop conditions. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I^2C bus may be taken when the P bit is set or the bus is Idle, with both the S and P bits clear.

In Firmware Controlled Master mode, user code conducts all I^2C bus operations based on Start and Stop bit conditions.

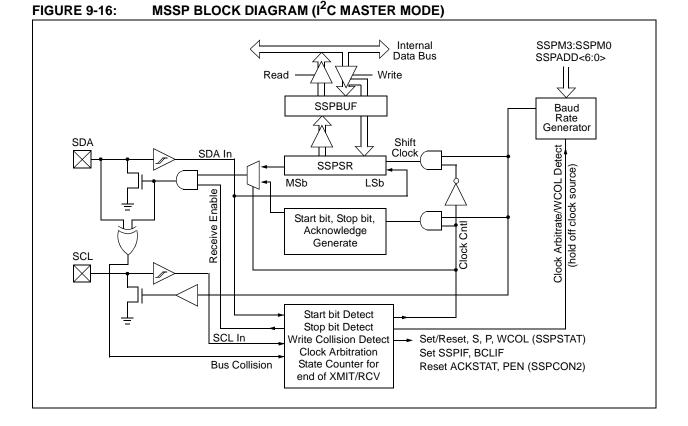
Once Master mode is enabled, the user has six options.

- 1. Assert a Start condition on SDA and SCL.
- 2. Assert a Repeated Start condition on SDA and SCL.
- 3. Write to the SSPBUF register, initiating transmission of data/address.
- 4. Configure the I²C port to receive data.
- 5. Generate an Acknowledge condition at the end of a received byte of data.
- 6. Generate a Stop condition on SDA and SCL.

Note: The MSSP module, when configured in I²C Master mode, does not allow queueing of events. For instance, the user is not allowed to initiate a Start condition and immediately write the SSPBUF register to initiate transmission before the Start condition is complete. In this case, the SSPBUF will not be written to and the WCOL bit will be set, indicating that a write to the SSPBUF did not occur.

The following events will cause SSP Interrupt Flag bit, SSPIF, to be set (SSP interrupt if enabled):

- · Start condition
- Stop condition
- Data transfer byte transmitted/received
- Acknowledge transmit
- Repeated Start



9.4.6.1 I²C Master Mode Operation

The master device generates all of the serial clock pulses and the Start and Stop conditions. A transfer is ended with a Stop condition or with a Repeated Start condition. Since the Repeated Start condition is also the beginning of the next serial transfer, the I²C bus will not be released.

In Master Transmitter mode, serial data is output through SDA while SCL outputs the serial clock. The first byte transmitted contains the slave address of the receiving device (7 bits) and the Read/Write (R/W) bit. In this case, the R/W bit will be logic '0'. Serial data is transmitted 8 bits at a time. After each byte is transmitted, an Acknowledge bit is received. Start and Stop conditions are output to indicate the beginning and the end of a serial transfer.

In Master Receive mode, the first byte transmitted contains the slave address of the transmitting device (7 bits) and the R/W bit. In this case, the R/W bit will be logic '1'. Thus, the first byte transmitted is a 7-bit slave address followed by a '1' to indicate the receive bit. Serial data is received via SDA while SCL outputs the serial clock. Serial data is received 8 bits at a time. After each byte is received, an Acknowledge bit is transmitted. Start and Stop conditions indicate the beginning and end of transmission.

The baud rate generator used for the SPI mode operation is used to set the SCL clock frequency for either 100 kHz, 400 kHz or 1 MHz I^2C operation. See **Section 9.4.7 "Baud Rate Generator"** for more detail. A typical transmit sequence would go as follows:

- 1. The user generates a Start condition by setting the Start Enable bit, SEN (SSPCON2<0>).
- SSPIF is set. The MSSP module will wait the required Start time before any other operation takes place.
- 3. The user loads the SSPBUF with the slave address to transmit.
- 4. Address is shifted out the SDA pin until all 8 bits are transmitted.
- 5. The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPCON2 register (SSPCON2<6>).
- 6. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- 7. The user loads the SSPBUF with eight bits of data.
- 8. Data is shifted out the SDA pin until all 8 bits are transmitted.
- The MSSP module shifts in the ACK bit from the slave device and writes its value into the SSPCON2 register (SSPCON2<6>).
- 10. The MSSP module generates an interrupt at the end of the ninth clock cycle by setting the SSPIF bit.
- 11. The user generates a Stop condition by setting the Stop Enable bit, PEN (SSPCON2<2>).
- 12. Interrupt is generated once the Stop condition is complete.

9.4.7 BAUD RATE GENERATOR

In I²C Master mode, the Baud Rate Generator (BRG) reload value is placed in the lower 7 bits of the SSPADD register (Figure 9-17). When a write occurs to SSPBUF, the Baud Rate Generator will automatically begin counting. The BRG counts down to 0 and stops until another reload has taken place. The BRG count is decremented twice per instruction cycle (TcY) on the Q2 and Q4 clocks. In I²C Master mode, the BRG is reloaded automatically.

Once the given operation is complete (i.e., transmission of the last data bit is followed by \overline{ACK}), the internal clock will automatically stop counting and the SCL pin will remain in its last state.

Table 9-3 demonstrates clock rates based on instruction cycles and the BRG value loaded into SSPADD.

FIGURE 9-17: BAUD RATE GENERATOR BLOCK DIAGRAM

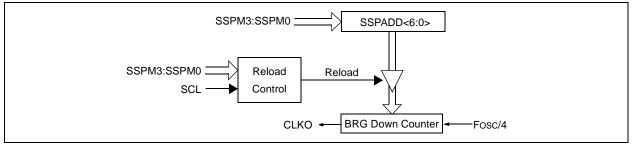


TABLE 9-3: I²C CLOCK RATE W/BRG

Fcy	Fcy*2	BRG Value	FscL (2 Rollovers of BRG)
10 MHz	20 MHz	19h	400 kHz ⁽¹⁾
10 MHz	20 MHz	20h	312.5 kHz
10 MHz	20 MHz	3Fh	100 kHz
4 MHz	8 MHz	0Ah	400 kHz ⁽¹⁾
4 MHz	8 MHz	0Dh	308 kHz
4 MHz	8 MHz	28h	100 kHz
1 MHz	2 MHz	03h	333 kHz ⁽¹⁾
1 MHz	2 MHz	0Ah	100 kHz
1 MHz	2 MHz	00h	1 MHz ⁽¹⁾

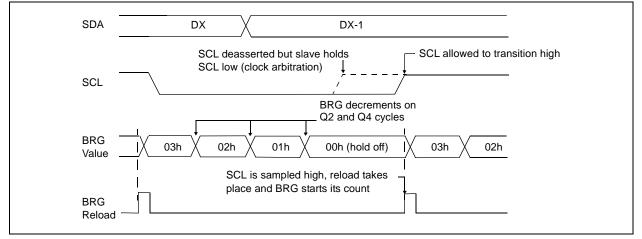
Note 1: The I²C interface does not conform to the 400 kHz I²C specification (which applies to rates greater than 100 kHz) in all details, but may be used with care where higher rates are required by the application.

9.4.7.1 Clock Arbitration

Clock arbitration occurs when the master, during any receive, transmit or Repeated Start/Stop condition, deasserts the SCL pin (SCL allowed to float high). When the SCL pin is allowed to float high, the Baud Rate Generator (BRG) is suspended from counting until the SCL pin is actually sampled high. When the

SCL pin is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. This ensures that the SCL high time will always be at least one BRG rollover count, in the event that the clock is held low by an external device (Figure 9-17).





9.4.8 I²C MASTER MODE START CONDITION TIMING

To initiate a Start condition, the user sets the Start condition enable bit, SEN (SSPCON2<0>). If the SDA and SCL pins are sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and starts its count. If SCL and SDA are both sampled high when the Baud Rate Generator times out (TBRG), the SDA pin is driven low. The action of the SDA being driven low, while SCL is high, is the Start condition and causes the S bit (SSPSTAT<3>) to be set. Following this, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and resumes its count. When the Baud Rate Generator times out (TBRG), the SEN bit (SSPCON2<0>) will be automatically cleared by hardware, the Baud Rate Generator is suspended, leaving the SDA line held low and the Start condition is complete.

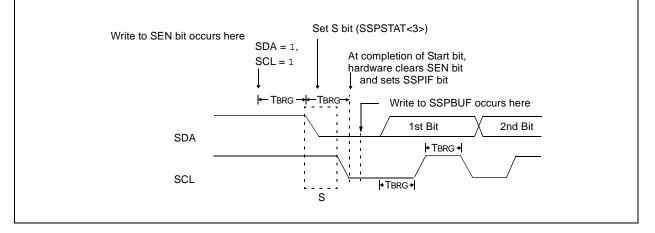
Note: If at the beginning of the Start condition, the SDA and SCL pins are already sampled low, or if during the Start condition, the SCL line is sampled low before the SDA line is driven low, a bus collision occurs, the Bus Collision Interrupt Flag (BCLIF) is set, the Start condition is aborted and the I²C module is reset into its Idle state.

9.4.8.1 WCOL Status Flag

If the user writes the SSPBUF when a Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing to the lower 5 bits of SSPCON2 is disabled until the Start condition is complete.

FIGURE 9-19: FIRST START BIT TIMING



9.4.9 I²C MASTER MODE REPEATED START CONDITION TIMING

A Repeated Start condition occurs when the RSEN bit (SSPCON2<1>) is programmed high and the I²C logic module is in the Idle state. When the RSEN bit is set, the SCL pin is asserted low. When the SCL pin is sampled low, the Baud Rate Generator is loaded with the contents of SSPADD<5:0> and begins counting. The SDA pin is released (brought high) for one Baud Rate Generator count (TBRG). When the Baud Rate Generator times out, if SDA is sampled high, the SCL pin will be deasserted (brought high). When SCL is sampled high, the Baud Rate Generator is reloaded with the contents of SSPADD<6:0> and begins counting. SDA and SCL must be sampled high for one TBRG. This action is then followed by assertion of the SDA pin (SDA = 0) for one TBRG while SCL is high. Following this, the RSEN bit (SSPCON2<1>) will be automatically cleared and the Baud Rate Generator will not be reloaded, leaving the SDA pin held low. As soon as a Start condition is detected on the SDA and SCL pins, the S bit (SSPSTAT<3>) will be set. The SSPIF bit will not be set until the Baud Rate Generator has timed out.

- Note 1: If RSEN is programmed while any other event is in progress, it will not take effect.
 - 2: A bus collision during the Repeated Start condition occurs if:
 - SDA is sampled low when SCL goes from low to high.
 - SCL goes low before SDA is asserted low. This may indicate that another master is attempting to transmit a data '1'.

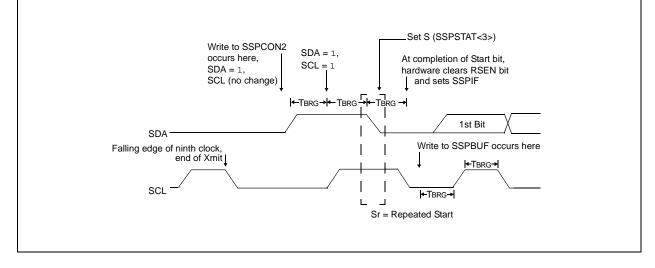
Immediately following the SSPIF bit getting set, the user may write the SSPBUF with the 7-bit address in 7-bit mode or the default first address in 10-bit mode. After the first eight bits are transmitted and an ACK is received, the user may then transmit an additional eight bits of address (10-bit mode) or eight bits of data (7-bit mode).

9.4.9.1 WCOL Status Flag

If the user writes the SSPBUF when a Repeated Start sequence is in progress, the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

Note: Because queueing of events is not allowed, writing of the lower 5 bits of SSPCON2 is disabled until the Repeated Start condition is complete.

FIGURE 9-20: REPEAT START CONDITION WAVEFORM



9.4.10 I²C MASTER MODE TRANSMISSION

Transmission of a data byte, a 7-bit address or the other half of a 10-bit address is accomplished by simply writing a value to the SSPBUF register. This action will set the Buffer Full flag bit, BF, and allow the Baud Rate Generator to begin counting and start the next transmission. Each bit of address/data will be shifted out onto the SDA pin after the falling edge of SCL is asserted (see data hold time specification, parameter #106). SCL is held low for one Baud Rate Generator rollover count (TBRG). Data should be valid before SCL is released high (see data setup time specification, parameter #107). When the SCL pin is released high, it is held that way for TBRG. The data on the SDA pin must remain stable for that duration and some hold time after the next falling edge of SCL. After the eighth bit is shifted out (the falling edge of the eighth clock), the BF flag is cleared and the master releases SDA. This allows the slave device being addressed to respond with an \overline{ACK} bit during the ninth bit time, if an address match occurred or if data was received properly. The status of \overline{ACK} is written into the ACKDT bit on the falling edge of the ninth clock. If the master receives an Acknowledge, the Acknowledge Status bit, ACKSTAT, is cleared. If not, the bit is set. After the ninth clock, the SSPIF bit is set and the master clock (Baud Rate Generator) is suspended until the next data byte is loaded into the SSPBUF, leaving SCL low and SDA unchanged (Figure 9-21).

After the write to the SSPBUF, each bit of address will be shifted out on the falling edge of SCL, until all seven address bits and the R/W bit are completed. On the falling edge of the eighth clock, the master will deassert the SDA pin, allowing the slave to respond with an Acknowledge. On the falling edge of the ninth clock, the master will sample the SDA pin to see if the address was recognized by a slave. The status of the ACK bit is loaded into the ACKSTAT status bit (SSPCON2<6>). Following the falling edge of the ninth clock transmission of the address, the SSPIF is set, the BF flag is cleared and the Baud Rate Generator is turned off until another write to the SSPBUF takes place, holding SCL low and allowing SDA to float.

9.4.10.1 BF Status Flag

In Transmit mode, the BF bit (SSPSTAT<0>) is set when the CPU writes to SSPBUF and is cleared when all eight bits are shifted out.

9.4.10.2 WCOL Status Flag

If the user writes the SSPBUF when a transmit is already in progress (i.e., SSPSR is still shifting out a data byte), the WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

WCOL must be cleared in software.

9.4.10.3 ACKSTAT Status Flag

In Transmit mode, the ACKSTAT bit (SSPCON2<6>) is cleared when the slave has sent an Acknowledge $(\overline{ACK} = 0)$ and is set when the slave does Not Acknowledge $(\overline{ACK} = 1)$. A slave sends an Acknowledge when it has recognized its address (including a general call) or when the slave has properly received its data.

9.4.11 I²C MASTER MODE RECEPTION

Master mode reception is enabled by programming the Receive Enable bit, RCEN (SSPCON2<3>).

Note: The MSSP module must be in an Idle state before the RCEN bit is set or the RCEN bit will be disregarded.

The Baud Rate Generator begins counting and on each rollover, the state of the SCL pin changes (high to low/ low to high) and data is shifted into the SSPSR. After the falling edge of the eighth clock, the receive enable flag is automatically cleared, the contents of the SSPSR are loaded into the SSPBUF, the BF flag bit is set, the SSPIF flag bit is set and the Baud Rate Generator is suspended from counting, holding SCL low. The MSSP is now in Idle state, awaiting the next command. When the buffer is read by the CPU, the BF flag bit is automatically cleared. The user can then send an Acknowledge bit at the end of reception by setting the Acknowledge Sequence Enable bit, ACKEN (SSPCON2<4>).

9.4.11.1 BF Status Flag

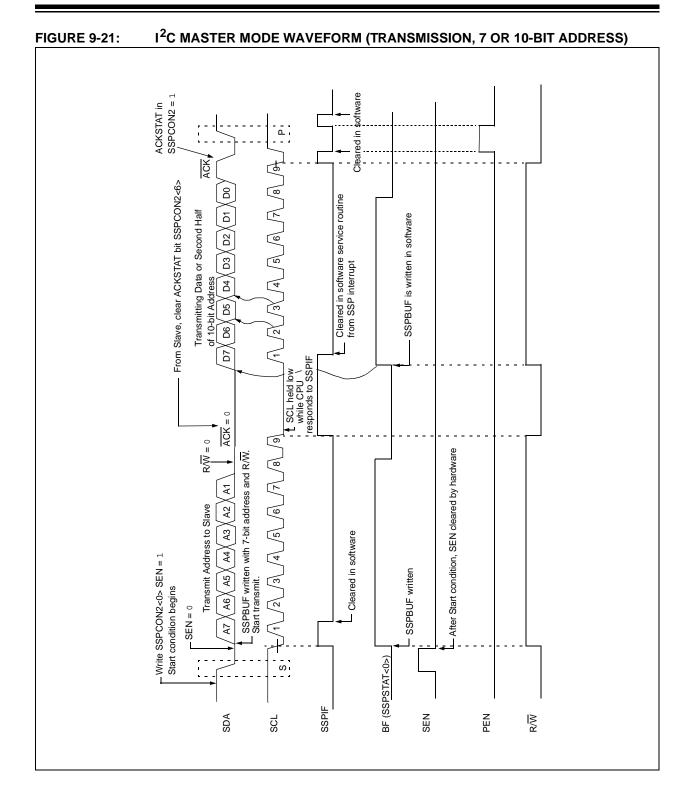
In receive operation, the BF bit is set when an address or data byte is loaded into SSPBUF from SSPSR. It is cleared when the SSPBUF register is read.

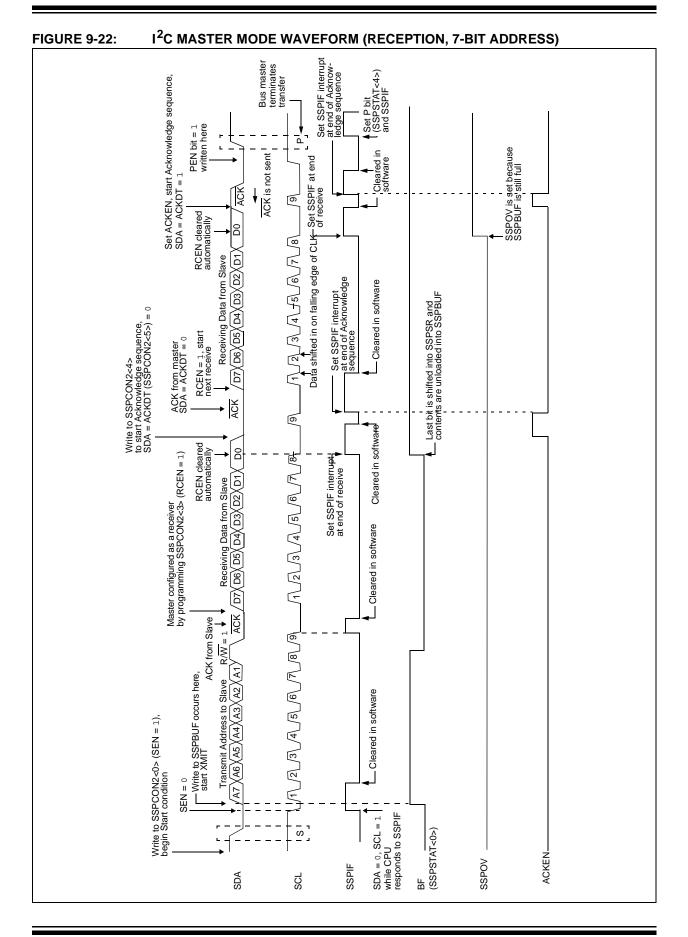
9.4.11.2 SSPOV Status Flag

In receive operation, the SSPOV bit is set when 8 bits are received into the SSPSR and the BF flag bit is already set from a previous reception.

9.4.11.3 WCOL Status Flag

If the user writes the SSPBUF when a receive is already in progress (i.e., SSPSR is still shifting in a data byte), the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).





9.4.12 ACKNOWLEDGE SEQUENCE TIMING

An Acknowledge sequence is enabled by setting the Acknowledge Sequence Enable bit. ACKEN (SSPCON2<4>). When this bit is set, the SCL pin is pulled low and the contents of the Acknowledge data bit are presented on the SDA pin. If the user wishes to generate an Acknowledge, then the ACKDT bit should be cleared. If not, the user should set the ACKDT bit before starting an Acknowledge sequence. The Baud Rate Generator then counts for one rollover period (TBRG) and the SCL pin is deasserted (pulled high). When the SCL pin is sampled high (clock arbitration), the Baud Rate Generator counts for TBRG. The SCL pin is then pulled low. Following this, the ACKEN bit is automatically cleared, the baud rate generator is turned off and the MSSP module then goes into Idle mode (Figure 9-23).

9.4.12.1 WCOL Status Flag

If the user writes the SSPBUF when an Acknowledge sequence is in progress, then WCOL is set and the contents of the buffer are unchanged (the write doesn't occur).

9.4.13 STOP CONDITION TIMING

A Stop bit is asserted on the SDA pin at the end of a receive/transmit by setting the Stop Sequence Enable bit, PEN (SSPCON2<2>). At the end of a receive/ transmit, the SCL line is held low after the falling edge of the ninth clock. When the PEN bit is set, the master will assert the SDA line low. When the SDA line is sampled low, the Baud Rate Generator is reloaded and counts down to 0. When the Baud Rate Generator times out, the SCL pin will be brought high and one TBRG (Baud Rate Generator rollover count) later, the SDA pin will be deasserted. When the SDA pin is sampled high while SCL is high, the P bit (SSPSTAT<4>) is set. A TBRG later, the PEN bit is cleared and the SSPIF bit is set (Figure 9-24).

9.4.13.1 WCOL Status Flag

If the user writes the SSPBUF when a Stop sequence is in progress, then the WCOL bit is set and the contents of the buffer are unchanged (the write doesn't occur).

FIGURE 9-23: ACKNOWLEDGE SEQUENCE WAVEFORM

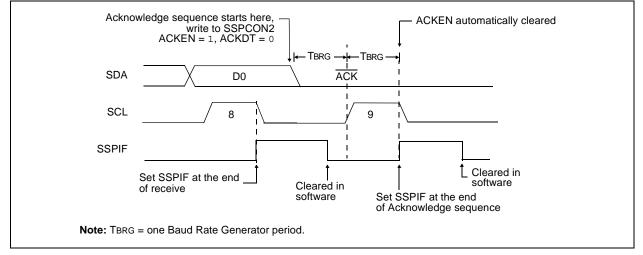
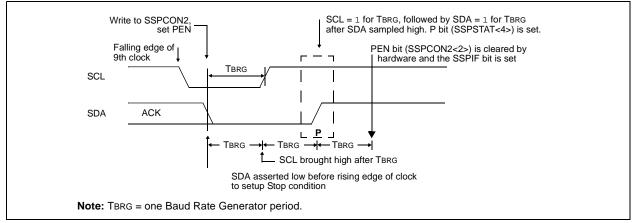


FIGURE 9-24: STOP CONDITION RECEIVE OR TRANSMIT MODE



9.4.14 SLEEP OPERATION

While in Sleep mode, the I²C module can receive addresses or data and when an address match or complete byte transfer occurs, wake the processor from Sleep (if the MSSP interrupt is enabled).

9.4.15 EFFECT OF A RESET

A Reset disables the MSSP module and terminates the current transfer.

9.4.16 MULTI-MASTER MODE

In Multi-Master mode, the interrupt generation on the detection of the Start and Stop conditions allows the determination of when the bus is free. The Stop (P) and Start (S) bits are cleared from a Reset or when the MSSP module is disabled. Control of the I^2C bus may be taken when the P bit (SSPSTAT<4>) is set, or the bus is Idle, with both the S and P bits clear. When the bus is busy, enabling the SSP interrupt will generate the interrupt when the Stop condition occurs.

In multi-master operation, the SDA line must be monitored for arbitration to see if the signal level is at the expected output level. This check is performed in hardware with the result placed in the BCLIF bit.

The states where arbitration can be lost are:

- Address Transfer
- Data Transfer
- A Start Condition
- A Repeated Start Condition
- An Acknowledge Condition

9.4.17 MULTI -MASTER COMMUNICATION, BUS COLLISION AND BUS ARBITRATION

Multi-Master mode support is achieved by bus arbitration. When the master outputs address/data bits onto the SDA pin, arbitration takes place when the master outputs a '1' on SDA by letting SDA float high and another master asserts a '0'. When the SCL pin floats high, data should be stable. If the expected data on SDA is a '1' and the data sampled on the SDA pin = 0, then a bus collision has taken place. The master will set the Bus Collision Interrupt Flag, BCLIF, and reset the I^2C port to its Idle state (Figure 9-25).

If a transmit was in progress when the bus collision occurred, the transmission is halted, the BF flag is cleared, the SDA and SCL lines are deasserted and the SSPBUF can be written to. When the user services the bus collision Interrupt Service Routine and if the I^2C bus is free, the user can resume communication by asserting a Start condition.

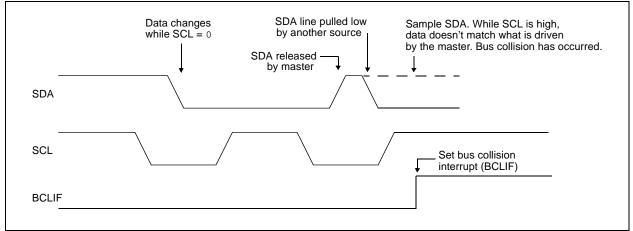
If a Start, Repeated Start, Stop or Acknowledge condition was in progress when the bus collision occurred, the condition is aborted, the SDA and SCL lines are deasserted and the respective control bits in the SSPCON2 register are cleared. When the user services the bus collision Interrupt Service Routine and if the I²C bus is free, the user can resume communication by asserting a Start condition.

The Master will continue to monitor the SDA and SCL pins. If a Stop condition occurs, the SSPIF bit will be set.

A write to the SSPBUF will start the transmission of data at the first data bit regardless of where the transmitter left off when the bus collision occurred.

In Multi-Master mode, the interrupt generation on the detection of Start and Stop conditions allows the determination of when the bus is free. Control of the I^2C bus can be taken when the P bit is set in the SSPSTAT register or the bus is Idle and the S and P bits are cleared.

FIGURE 9-25: BUS COLLISION TIMING FOR TRANSMIT AND ACKNOWLEDGE



9.4.17.1 Bus Collision During a Start Condition

During a Start condition, a bus collision occurs if:

- a) SDA or SCL are sampled low at the beginning of the Start condition (Figure 9-26).
- b) SCL is sampled low before SDA is asserted low (Figure 9-27).

During a Start condition, both the SDA and the SCL pins are monitored.

If the SDA pin is already low, or the SCL pin is already low, then all of the following occur:

- the Start condition is aborted,
- the BCLIF flag is set and
- the MSSP module is reset to its Idle state (Figure 9-26).

The Start condition begins with the SDA and SCL pins deasserted. When the SDA pin is sampled high, the Baud Rate Generator is loaded from SSPADD<6:0> and counts down to 0. If the SCL pin is sampled low while SDA is high, a bus collision occurs because it is assumed that another master is attempting to drive a data '1' during the Start condition.

If the SDA pin is sampled low during this count, the BRG is reset and the SDA line is asserted early (Figure 9-28). If, however, a '1' is sampled on the SDA pin, the SDA pin is asserted low at the end of the BRG count. The Baud Rate Generator is then reloaded and counts down to 0 and during this time, if the SCL pin is sampled as '0', a bus collision does not occur. At the end of the BRG count, the SCL pin is asserted low.

Note: The reason that bus collision is not a factor during a Start condition is that no two bus masters can assert a Start condition at the exact same time. Therefore, one master will always assert SDA before the other. This condition does not cause a bus collision because the two masters must be allowed to arbitrate the first address following the Start condition. If the address is the same, arbitration must be allowed to continue into the data portion, Repeated Start or Stop conditions.

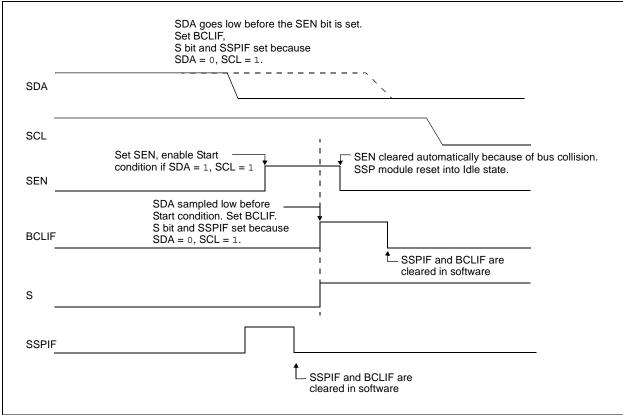
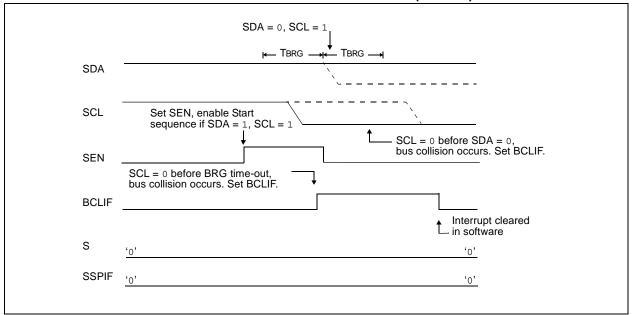
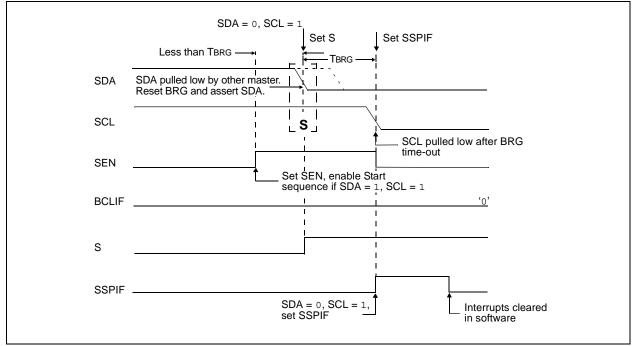


FIGURE 9-26: BUS COLLISION DURING START CONDITION (SDA ONLY)









9.4.17.2 Bus Collision During a Repeated Start Condition

During a Repeated Start condition, a bus collision occurs if:

- a) A low level is sampled on SDA when SCL goes from low level to high level.
- b) SCL goes low before SDA is asserted low, indicating that another master is attempting to transmit a data '1'.

When the user deasserts SDA and the pin is allowed to float high, the BRG is loaded with SSPADD<6:0> and counts down to 0. The SCL pin is then deasserted and when sampled high, the SDA pin is sampled.

If SDA is low, a bus collision has occurred (i.e., another master is attempting to transmit a data '0', see Figure 9-29). If SDA is sampled high, the BRG is

reloaded and begins counting. If SDA goes from high to low before the BRG times out, no bus collision occurs because no two masters can assert SDA at exactly the same time.

If SCL goes from high to low before the BRG times out and SDA has not already been asserted, a bus collision occurs. In this case, another master is attempting to transmit a data '1' during the Repeated Start condition (Figure 9-30).

If at the end of the BRG time-out, both SCL and SDA are still high, the SDA pin is driven low and the BRG is reloaded and begins counting. At the end of the count, regardless of the status of the SCL pin, the SCL pin is driven low and the Repeated Start condition is complete.

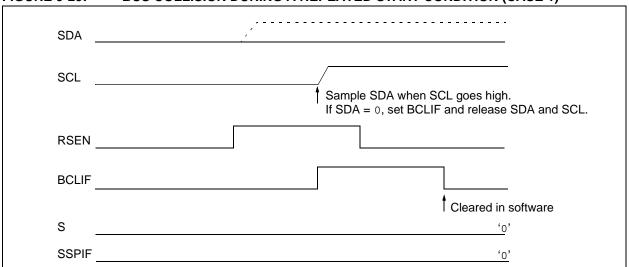
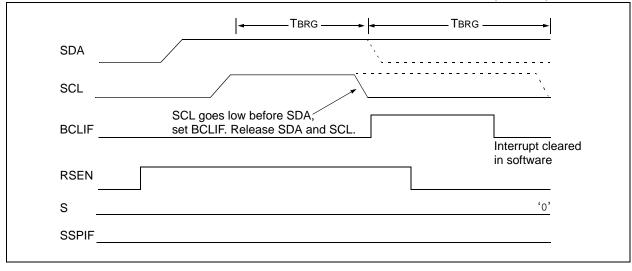


FIGURE 9-29: BUS COLLISION DURING A REPEATED START CONDITION (CASE 1)

FIGURE 9-30: BUS COLLISION DURING REPEATED START CONDITION (CASE 2)



9.4.17.3 Bus Collision During a Stop Condition

Bus collision occurs during a Stop condition if:

- a) After the SDA pin has been deasserted and allowed to float high, SDA is sampled low after the BRG has timed out.
- b) After the SCL pin is deasserted, SCL is sampled low before SDA goes high.

The Stop condition begins with SDA asserted low. When SDA is sampled low, the SCL pin is allowed to float. When the pin is sampled high (clock arbitration), the Baud Rate Generator is loaded with SSPADD<6:0> and counts down to 0. After the BRG times out, SDA is sampled. If SDA is sampled low, a bus collision has occurred. This is due to another master attempting to drive a data '0' (Figure 9-31). If the SCL pin is sampled low before SDA is allowed to float high, a bus collision occurs. This is another case of another master attempting to drive a data '0' (Figure 9-32).

FIGURE 9-31: BUS COLLISION DURING A STOP CONDITION (CASE 1)

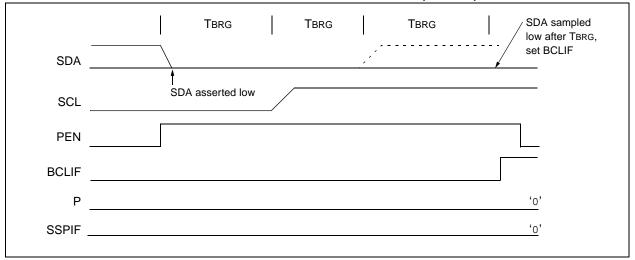
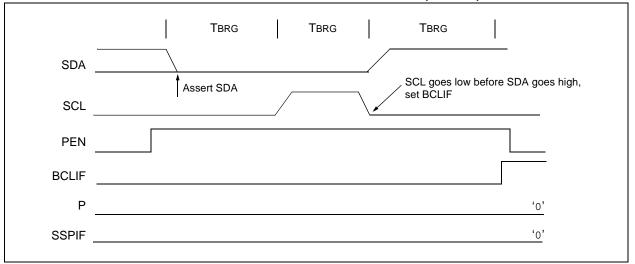


FIGURE 9-32: BUS COLLISION DURING A STOP CONDITION (CASE 2)



NOTES:

10.0 ADDRESSABLE UNIVERSAL SYNCHRONOUS ASYNCHRONOUS RECEIVER TRANSMITTER (USART)

The Universal Synchronous Asynchronous Receiver Transmitter (USART) module is one of the two serial I/O modules. (USART is also known as a Serial Communications Interface or SCI.) The USART can be configured as a full-duplex asynchronous system that can communicate with peripheral devices, such as CRT terminals and personal computers, or it can be configured as a half-duplex synchronous system that can communicate with peripheral devices, such as A/D or D/A integrated circuits, serial EEPROMs, etc. The USART can be configured in the following modes:

- Asynchronous (full-duplex)
- Synchronous Master (half-duplex)
- Synchronous Slave (half-duplex)

Bit SPEN (RCSTA<7>) and bits TRISC<7:6> have to be set in order to configure pins RC6/TX/CK and RC7/RX/DT as the Universal Synchronous Asynchronous Receiver Transmitter.

The USART module also has a multi-processor communication capability using 9-bit address detection.

REGISTER 10-1: TXSTA: TRANSMIT STATUS AND CONTROL REGISTER (ADDRESS 98h)

							,	-						
	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0	R-1	R/W-0						
	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D						
	bit 7							bit 0						
bit 7	CSRC: Cloc	CSRC: Clock Source Select bit												
	Asynchronou	us mode:												
	Don't care.													
	<u>Synchronous</u> 1 = Master n		apporated i	ntornally fro										
	1 = Master fi0 = Slave mo				li broj									
bit 6	TX9 : 9-bit Tr	-		,										
	1 = Selects §													
	0 = Selects 8													
bit 5	TXEN: Trans	smit Enable	bit											
	1 = Transmit	enabled												
	0 = Transmit	disabled												
	Note: S	SREN/CRE	V overrides	TXEN in Sy	nc mode.									
bit 4	SYNC: USA	RT Mode Se	elect bit											
	1 = Synchro	nous mode												
	0 = Asynchro	onous mode	9											
bit 3	Unimpleme	nted: Read	as '0'											
bit 2	BRGH: High	Baud Rate	Select bit											
	Asynchronou													
	1 = High spe 0 = Low spe													
	0 = Low spe													
	Unused in th													
bit 1	TRMT: Trans	smit Shift Re	egister Statu	is bit										
	1 = TSR em		0											
	0 = TSR full													
bit 0	TX9D: 9th bi	it of Transm	it Data, can	be Parity bit	t									
	Legend:													
	R = Readabl	le bit	W = Wr	itable bit	U = Unimp	lemented b	it, read as ') '						
	- n = Value a	at POR	'1' = Bit	is set	'0' = Bit is	cleared	x = Bit is ur	nknown						

	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R-0	R-0	R-x			
	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D			
	bit 7		•				•	bit 0			
bit 7	SPEN: Serial Port Enable bit										
		oort enabled		RC7/RX/D	T and RC6/T	X/CK pins a	as serial port	: pins)			
bit 6	RX9 : 9-bit	Receive Ena	able bit								
		9-bit recep 8-bit recep									
bit 5	SREN: Sin	gle Receive	Enable bit								
	Asynchrone Don't care.										
		us mode – N									
		es single rec									
	 Disables single receive This bit is cleared after reception is complete. 										
	Synchronous mode – Slave:										
	Don't care.										
bit 4	CREN: Continuous Receive Enable bit										
	<u>Asynchronous mode:</u> 1 = Enables continuous receive										
	0 = Disables continuous receive										
	Synchronous mode:										
	 1 = Enables continuous receive until enable bit CREN is cleared (CREN overrides SREN) 0 = Disables continuous receive 										
bit 3	ADDEN: Address Detect Enable bit										
	Asynchron	ous mode 9	-bit (RX9 = :	1 <u>):</u>							
	1 = Enables address detection, enables interrupt and load of the receive buffer when RSR<8>										
	is set 0 = Disables address detection, all bytes are received and ninth bit can be used as parity bit										
bit 2		ming Error b									
~		-		by reading	RCREG regi	ister and red	ceive next va	alid byte)			
	 1 = Framing error (can be updated by reading RCREG register and receive next valid byte) 0 = No framing error 										
bit 1	OERR: Ov	errun Error	bit								
	 1 = Overrun error (can be cleared by clearing bit CREN) 0 = No overrun error 										
bit 0					h 14 h 4		ed by user f				

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented	l bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

10.1 USART Baud Rate Generator (BRG)

The BRG supports both the Asynchronous and Synchronous modes of the USART. It is a dedicated 8-bit baud rate generator. The SPBRG register controls the period of a free running 8-bit timer. In Asynchronous mode, bit BRGH (TXSTA<2>) also controls the baud rate. In Synchronous mode, bit BRGH is ignored. Table 10-1 shows the formula for computation of the baud rate for different USART modes which only apply in Master mode (internal clock).

Given the desired baud rate and Fosc, the nearest integer value for the SPBRG register can be calculated using the formula in Table 10-1. From this, the error in baud rate can be determined. It may be advantageous to use the high baud rate (BRGH = 1) even for slower baud clocks. This is because the Fosc/(16 (X + 1)) equation can reduce the baud rate error in some cases.

Writing a new value to the SPBRG register causes the BRG timer to be reset (or cleared). This ensures the BRG does not wait for a timer overflow before outputting the new baud rate.

10.1.1 SAMPLING

The data on the RC7/RX/DT pin is sampled three times by a majority detect circuit to determine if a high or a low level is present at the RX pin.

TABLE 10-1: BAUD RATE FORMULA

SYNC	BRGH = 0 (Low Speed)	BRGH = 1 (High Speed)
0	(Asynchronous) Baud Rate = Fosc/(64 (X + 1))	Baud Rate = Fosc/(16 (X + 1))
1	(Synchronous) Baud Rate = FOSC/(4 (X + 1))	N/A

Legend: X = value in SPBRG (0 to 255)

TABLE 10-2: REGISTERS ASSOCIATED WITH BAUD RATE GENERATOR

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value POR,		Valu all o Res	ther
98h	TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000	-010	0000	-010
18h	RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000	000x	0000	000x
99h	SPBRG	Baud Ra	Baud Rate Generator Register								0000	0000	0000

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used by the BRG.

BAUD	F	osc = 20 N	IHz	F	osc = 16 N	IHz	Fosc = 10 MHz			
RATE (K)	KBAUD	% ERROR	SPBRG value (decimal)	KBAUD	% ERROR	SPBRG value (decimal)	KBAUD	% ERROR	SPBRG value (decimal)	
0.3	-	-	-	-	-	-	-	-	-	
1.2	1.221	1.75	255	1.202	0.17	207	1.202	0.17	129	
2.4	2.404	0.17	129	2.404	0.17	103	2.404	0.17	64	
9.6	9.766	1.73	31	9.615	0.16	25	9.766	1.73	15	
19.2	19.531	1.72	15	19.231	0.16	12	19.531	1.72	7	
28.8	31.250	8.51	9	27.778	3.55	8	31.250	8.51	4	
33.6	34.722	3.34	8	35.714	6.29	6	31.250	6.99	4	
57.6	62.500	8.51	4	62.500	8.51	3	52.083	9.58	2	
HIGH	1.221	-	255	0.977	-	255	0.610	-	255	
LOW	312.500	-	0	250.000	-	0	156.250	-	0	

TABLE 10-3: BAUD RATES FOR ASYNCHRONOUS MODE (BRGH = 0)

DAUD		Fosc = 4 M	Hz	Fo	sc = 3.6864	MHz
BAUD RATE (K)	KBAUD	% ERROR	SPBRG value (decimal)	KBAUD	% ERROR	SPBRG value (decimal)
0.3	0.300	0	207	0.3	0	191
1.2	1.202	0.17	51	1.2	0	47
2.4	2.404	0.17	25	2.4	0	23
9.6	8.929	6.99	6	9.6	0	5
19.2	20.833	8.51	2	19.2	0	2
28.8	31.250	8.51	1	28.8	0	1
33.6	-	-	-	-	-	-
57.6	62.500	8.51	0	57.6	0	0
HIGH	0.244	-	255	0.225	-	255
LOW	62.500	-	0	57.6	-	0

TABLE 10-4: BAUD RATES FOR ASYNCHRONOUS MODE (BRGH = 1)

BAUD	F	osc = 20 M	Hz	F	osc = 16 M	Hz	Fosc = 10 MHz			
RATE (K)	KBAUD	% ERROR	SPBRG value (decimal)	KBAUD	% ERROR	SPBRG value (decimal)	KBAUD	% ERROR	SPBRG value (decimal)	
0.3	-	-	-	-	-	-	-	-	-	
1.2	-	-	-	-	-	-	-	-	-	
2.4	-	-	-	-	-	-	2.441	1.71	255	
9.6	9.615	0.16	129	9.615	0.16	103	9.615	0.16	64	
19.2	19.231	0.16	64	19.231	0.16	51	19.531	1.72	31	
28.8	29.070	0.94	42	29.412	2.13	33	28.409	1.36	21	
33.6	33.784	0.55	36	33.333	0.79	29	32.895	2.10	18	
57.6	59.524	3.34	20	58.824	2.13	16	56.818	1.36	10	
HIGH	4.883	-	255	3.906	-	255	2.441	-	255	
LOW	1250.000	-	0	1000.000		0	625.000	-	0	

BAUD	F	osc = 4 MH	łz	Fos	c = 3.6864	MHz
RATE (K)	KBAUD	% ERROR	SPBRG value (decimal)	KBAUD	% ERROR	SPBRG value (decimal)
0.3	-	-	-	-	-	-
1.2	1.202	0.17	207	1.2	0	191
2.4	2.404	0.17	103	2.4	0	95
9.6	9.615	0.16	25	9.6	0	23
19.2	19.231	0.16	12	19.2	0	11
28.8	27.798	3.55	8	28.8	0	7
33.6	35.714	6.29	6	32.9	2.04	6
57.6	62.500	8.51	3	57.6	0	3
HIGH	0.977	-	255	0.9	-	255
LOW	250.000	-	0	230.4	-	0

10.2 USART Asynchronous Mode

In this mode, the USART uses standard Non-Returnto-Zero (NRZ) format (one Start bit, eight or nine data bits and one Stop bit). The most common data format is 8 bits. An on-chip, dedicated, 8-bit Baud Rate Generator can be used to derive standard baud rate frequencies from the oscillator. The USART transmits and receives the LSb first. The transmitter and receiver are functionally independent but use the same data format and baud rate. The baud rate generator produces a clock, either x16 or x64 of the bit shift rate, depending on bit BRGH (TXSTA<2>). Parity is not supported by the hardware but can be implemented in software (and stored as the ninth data bit). Asynchronous mode is stopped during Sleep.

Asynchronous mode is selected by clearing bit SYNC (TXSTA<4>).

The USART Asynchronous module consists of the following important elements:

- Baud Rate Generator
- Sampling Circuit
- Asynchronous Transmitter
- Asynchronous Receiver

10.2.1 USART ASYNCHRONOUS TRANSMITTER

The USART transmitter block diagram is shown in Figure 10-1. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The shift register obtains its data from the Read/Write Transmit Buffer, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the Stop bit has been transmitted from the previous load. As soon as the Stop bit is transmitted, the TSR is loaded with new data from the TXREG register (if available). Once the TXREG register transfers the data to the TSR register (occurs in one TCY), the TXREG register is empty and flag bit, TXIF (PIR1<4>), is set. This interrupt can be

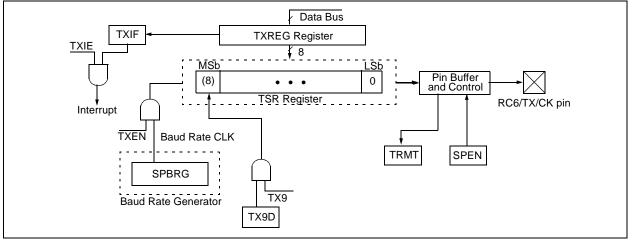
enabled/disabled by setting/clearing enable bit, TXIE (PIE1<4>). Flag bit TXIF will be set regardless of the state of enable bit TXIE and cannot be cleared in software. It will reset only when new data is loaded into the TXREG register. While flag bit TXIF indicates the status of the TXREG register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. Status bit TRMT is a read-only bit which is set when the TSR register is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty.

- Note 1: The TSR register is not mapped in data memory so it is not available to the user.
 - 2: Flag bit TXIF is set when enable bit TXEN is set. TXIF is cleared by loading TXREG.

Transmission is enabled by setting enable bit, TXEN (TXSTA<5>). The actual transmission will not occur until the TXREG register has been loaded with data and the Baud Rate Generator (BRG) has produced a shift clock (Figure 10-2). The transmission can also be started by first loading the TXREG register and then setting enable bit TXEN. Normally, when transmission is first started, the TSR register is empty. At that point, transfer to the TXREG register will result in an immediate transfer to TSR, resulting in an empty TXREG. A back-to-back transfer is thus possible (Figure 10-3). Clearing enable bit TXEN during a transmission will cause the transmission to be aborted and will reset the transmitter. As a result, the RC6/TX/CK pin will revert to high-impedance.

In order to select 9-bit transmission, transmit bit TX9 (TXSTA<6>) should be set and the ninth bit should be written to TX9D (TXSTA<0>). The ninth bit must be written before writing the 8-bit data to the TXREG register. This is because a data write to the TXREG register can result in an immediate transfer of the data to the TSR register (if the TSR is empty). In such a case, an incorrect ninth data bit may be loaded in the TSR register.





When setting up an Asynchronous Transmission, follow these steps:

- Initialize the SPBRG register for the appropriate baud rate. If a high-speed baud rate is desired, set bit BRGH (Section 10.1 "USART Baud Rate Generator (BRG)").
- 2. Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
- 3. If interrupts are desired, then set enable bit TXIE.
- 4. If 9-bit transmission is desired, then set transmit bit TX9.

- 5. Enable the transmission by setting bit TXEN, which will also set bit TXIF.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
- 7. Load data to the TXREG register (starts transmission).
- 8. If using interrupts, ensure that GIE and PEIE (bits 7 and 6) of the INTCON register are set.

FIGURE 10-2: ASYNCHRONOUS MASTER TRANSMISSION

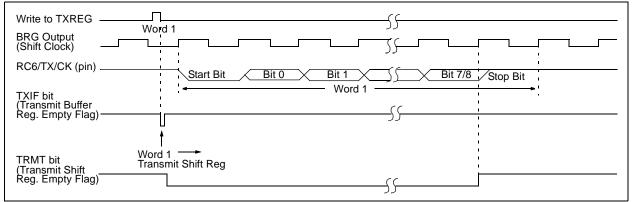


FIGURE 10-3: ASYNCHRONOUS MASTER TRANSMISSION (BACK TO BACK)

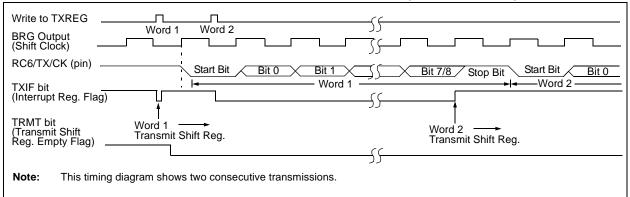


TABLE 10-5: REGISTERS ASSOCIATED WITH ASYNCHRONOUS TRANSMISSION

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	R0IF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
18h	RCSTA	SPEN	RX9	SREN	CREN		FERR	OERR	RX9D	0000 -00x	0000 -00x
19h	TXREG	USART Tra	nsmit Re	gister						0000 0000	0000 0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC		BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	PBRG Baud Rate Generator Register									0000 0000
Logond:		wn – unin	d for acym	chronous trans	mission						

Legend: x = unknown, - = unimplemented locations read as '0'. Shaded cells are not used for asynchronous transmission.
 Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices; always maintain these bits clear.

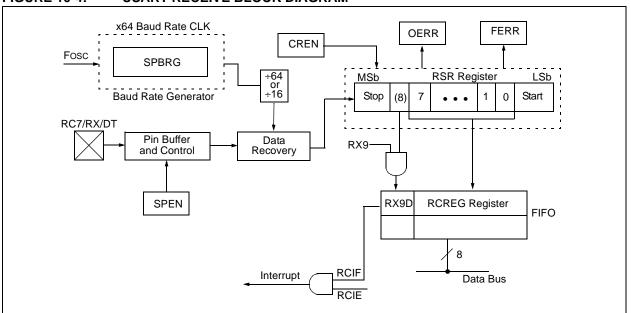
10.2.2 USART ASYNCHRONOUS RECEIVER

The receiver block diagram is shown in Figure 10-4. The data is received on the RC7/RX/DT pin and drives the data recovery block. The data recovery block is actually a high-speed shifter, operating at x16 times the baud rate; whereas the main receive serial shifter operates at the bit rate or at Fosc.

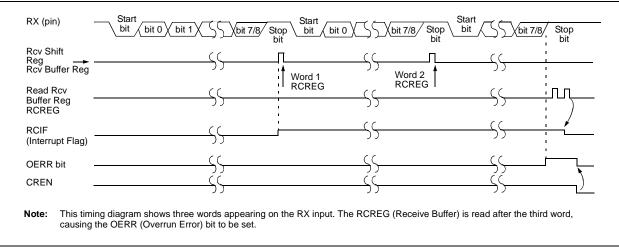
Once Asynchronous mode is selected, reception is enabled by setting bit CREN (RCSTA<4>).

The heart of the receiver is the Receive (Serial) Shift Register (RSR). After sampling the Stop bit, the received data in the RSR is transferred to the RCREG register (if it is empty). If the transfer is complete, flag bit, RCIF (PIR1<5>), is set. The actual interrupt can be enabled/disabled by setting/clearing enable bit, RCIE (PIE1<5>). Flag bit RCIF is a read-only bit which is cleared by the hardware. It is cleared when the RCREG register has been read and is empty. The RCREG is a double-buffered register (i.e., it is a two-deep FIFO). It is possible for two bytes of data to be received and transferred to the RCREG FIFO and a third byte to begin shifting to the RSR register. On the detection of the Stop bit of the third byte, if the RCREG register is still full, the Overrun Error bit, OERR (RCSTA<1>), will be set. The word in the RSR will be lost. The RCREG register can be read twice to retrieve the two bytes in the FIFO. Overrun bit OERR has to be cleared in software. This is done by resetting the receive logic (CREN is cleared and then set). If bit OERR is set, transfers from the RSR register to the RCREG register are inhibited and no further data will be received. It is, therefore, essential to clear error bit OERR if it is set. Framing error bit, FERR (RCSTA<2>), is set if a Stop bit is detected as clear. Bit FERR and the 9th receive bit are buffered the same way as the receive data. Reading the RCREG will load bits RX9D and FERR with new values, therefore, it is essential for the user to read the RCSTA register before reading the RCREG register in order not to lose the old FERR and RX9D information.

FIGURE 10-4: USART RECEIVE BLOCK DIAGRAM







When setting up an Asynchronous Reception, follow these steps:

- Initialize the SPBRG register for the appropriate baud rate. If a high-speed baud rate is desired, set bit BRGH (Section 10.1 "USART Baud Rate Generator (BRG)").
- 2. Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
- 3. If interrupts are desired, then set enable bit RCIE.
- 4. If 9-bit reception is desired, then set bit RX9.
- 5. Enable the reception by setting bit CREN.

- 6. Flag bit RCIF will be set when reception is complete and an interrupt will be generated if enable bit RCIE is set.
- 7. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
- 8. Read the 8-bit received data by reading the RCREG register.
- 9. If any error occurred, clear the error by clearing enable bit CREN.
- 10. If using interrupts, ensure that GIE and PEIE (bits 7 and 6) of the INTCON register are set.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	R0IF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
18h	RCSTA	SPEN	RX9	SREN	CREN		FERR	OERR	RX9D	x00- 0000	0000 -00x
1Ah	RCREG	USART R	eceive Reg	jister						0000 0000	0000 0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC		BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	Baud Rat	Baud Rate Generator Register								0000 0000

TABLE 10-6: REGISTERS ASSOCIATED WITH ASYNCHRONOUS RECEPTION

Legend: x = unknown, - = unimplemented locations read as '0'. Shaded cells are not used for asynchronous reception.

Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices; always maintain these bits clear.

10.2.3 SETTING UP 9-BIT MODE WITH ADDRESS DETECT

When setting up an Asynchronous Reception with address detect enabled:

- Initialize the SPBRG register for the appropriate baud rate. If a high-speed baud rate is desired, set bit BRGH.
- Enable the asynchronous serial port by clearing bit SYNC and setting bit SPEN.
- If interrupts are desired, then set enable bit RCIE.
- Set bit RX9 to enable 9-bit reception.
- Set ADDEN to enable address detect.
- Enable the reception by setting enable bit CREN.

- Flag bit RCIF will be set when reception is complete, and an interrupt will be generated if enable bit RCIE was set.
- Read the RCSTA register to get the ninth bit and determine if any error occurred during reception.
- Read the 8-bit received data by reading the RCREG register to determine if the device is being addressed.
- If any error occurred, clear the error by clearing enable bit CREN.
- If the device has been addressed, clear the ADDEN bit to allow data bytes and address bytes to be read into the receive buffer and interrupt the CPU.

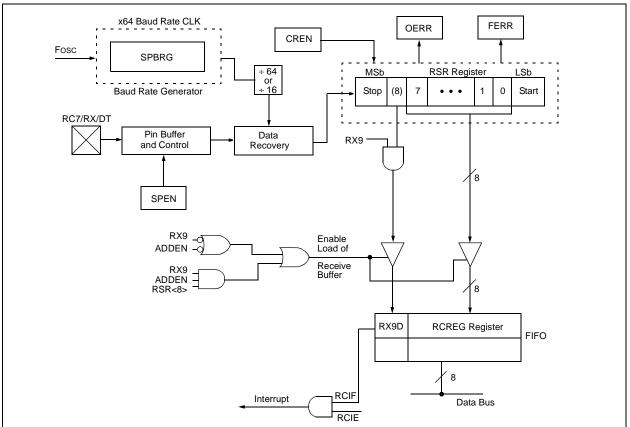
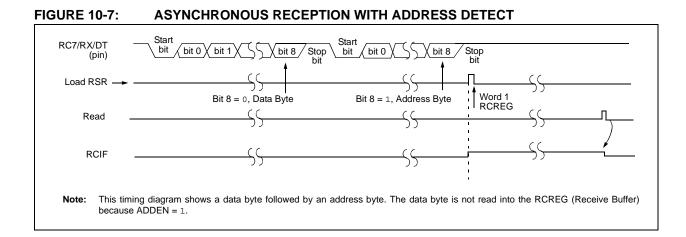
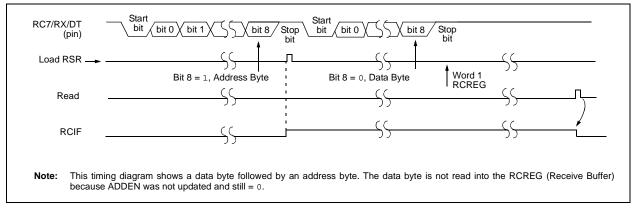


FIGURE 10-6: USART RECEIVE BLOCK DIAGRAM







Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value o POR, BO		Value all ot Rese	her
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	R0IF	0000 00	00x	0000	000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 00	000	0000	0000
18h	RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 00	200	0000	x000
1Ah	RCREG	USART Red	ceive Reo	gister						0000 00	000	0000	0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 00	000	0000	0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	0000 -0	010	0000	-010
99h	99h SPBRG Baud Rate Generator Register								0000 00	000	0000	0000	

Legend: x = unknown, - = unimplemented locations read as '0'. Shaded cells are not used for asynchronous reception.

Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices; always maintain these bits clear.

10.3 USART Synchronous Master Mode

In Synchronous Master mode, the data is transmitted in a half-duplex manner (i.e., transmission and reception do not occur at the same time). When transmitting data, the reception is inhibited and vice versa. Synchronous mode is entered by setting bit, SYNC (TXSTA<4>). In addition, enable bit, SPEN (RCSTA<7>), is set in order to configure the RC6/TX/CK and RC7/RX/DT I/O pins to CK (clock) and DT (data) lines, respectively. The Master mode indicates that the processor transmits the master clock on the CK line. The Master mode is entered by setting bit, CSRC (TXSTA<7>).

10.3.1 USART SYNCHRONOUS MASTER TRANSMISSION

The USART transmitter block diagram is shown in Figure 10-6. The heart of the transmitter is the Transmit (Serial) Shift Register (TSR). The shift register obtains its data from the Read/Write Transmit Buffer register, TXREG. The TXREG register is loaded with data in software. The TSR register is not loaded until the last bit has been transmitted from the previous load. As soon as the last bit is transmitted, the TSR is loaded with new data from the TXREG (if available). Once the TXREG register transfers the data to the TSR register (occurs in one TCYCLE), the TXREG is empty and interrupt bit, TXIF (PIR1<4>), is set. The interrupt can be enabled/disabled by setting/clearing enable bit TXIE (PIE1<4>). Flag bit TXIF will be set regardless of the state of enable bit TXIE and cannot be cleared in software. It will reset only when new data is loaded into the TXREG register. While flag bit TXIF indicates the status of the TXREG register, another bit, TRMT (TXSTA<1>), shows the status of the TSR register. TRMT is a readonly bit which is set when the TSR is empty. No interrupt logic is tied to this bit so the user has to poll this bit in order to determine if the TSR register is empty. The TSR is not mapped in data memory so it is not available to the user.

Transmission is enabled by setting enable bit, TXEN (TXSTA<5>). The actual transmission will not occur until the TXREG register has been loaded with data. The first data bit will be shifted out on the next available rising edge of the clock on the CK line. Data out is stable around the falling edge of the synchronous clock (Figure 10-9). The transmission can also be started by first loading the TXREG register and then setting bit TXEN (Figure 10-10). This is advantageous when slow baud rates are selected since the BRG is kept in Reset when bits TXEN, CREN and SREN are clear. Setting enable bit TXEN will start the BRG, creating a shift clock immediately. Normally, when transmission is first started, the TSR register is empty so a transfer to the TXREG register will result in an immediate transfer to TSR, resulting in an empty TXREG. Back-to-back transfers are possible.

Clearing enable bit TXEN during a transmission will cause the transmission to be aborted and will reset the transmitter. The DT and CK pins will revert to highimpedance. If either bit CREN or bit SREN is set during a transmission, the transmission is aborted and the DT pin reverts to a high-impedance state (for a reception). The CK pin will remain an output if bit CSRC is set (internal clock). The transmitter logic, however, is not reset, although it is disconnected from the pins. In order to reset the transmitter, the user has to clear bit TXEN. If bit SREN is set (to interrupt an on-going transmission and receive a single word), then after the single word is received, bit SREN will be cleared and the serial port will revert back to transmitting since bit TXEN is still set. The DT line will immediately switch from High-Impedance Receive mode to transmit and start driving. To avoid this, bit TXEN should be cleared.

In order to select 9-bit transmission, the TX9 (TXSTA<6>) bit should be set and the ninth bit should be written to bit TX9D (TXSTA<0>). The ninth bit must be written before writing the 8-bit data to the TXREG register. This is because a data write to the TXREG can result in an immediate transfer of the data to the TSR register (if the TSR is empty). If the TSR was empty and the TXREG was written before writing the "new" TX9D, the "present" value of bit TX9D is loaded.

Steps to follow when setting up a Synchronous Master Transmission:

- 1. Initialize the SPBRG register for the appropriate baud rate (Section 10.1 "USART Baud Rate Generator (BRG)").
- 2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
- 3. If interrupts are desired, set enable bit TXIE.
- 4. If 9-bit transmission is desired, set bit TX9.
- 5. Enable the transmission by setting bit TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
- 7. Start transmission by loading data to the TXREG register.
- 8. If using interrupts, ensure that GIE and PEIE (bits 7 and 6) of the INTCON register are set.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	R0IF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
18h	RCSTA	SPEN	RX9	SREN	CREN		FERR	OERR	RX9D	0000 -00x	0000 -00x
19h	TXREG	USART Tr	ansmit Re	egister						0000 0000	0000 0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC	—	BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	Baud Rate Generator Register							0000 0000	0000 0000	

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used for synchronous master transmission.

Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices; always maintain these bits clear.

FIGURE 10-9: SYNCHRONOUS TRANSMISSION

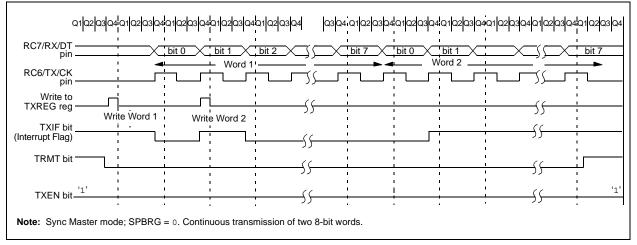
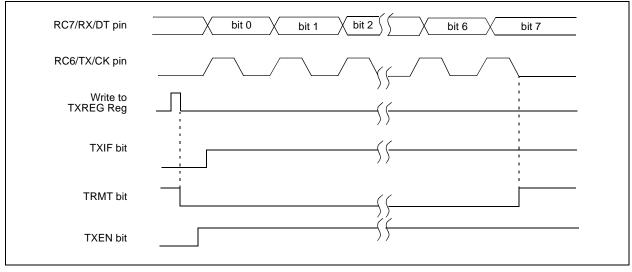


FIGURE 10-10: SYNCHRONOUS TRANSMISSION (THROUGH TXEN)



10.3.2 USART SYNCHRONOUS MASTER RECEPTION

Once Synchronous mode is selected, reception is enabled by setting either enable bit, SREN (RCSTA<5>), or enable bit, CREN (RCSTA<4>). Data is sampled on the RC7/RX/DT pin on the falling edge of the clock. If enable bit SREN is set, then only a single word is received. If enable bit CREN is set, the reception is continuous until CREN is cleared. If both bits are set, CREN takes precedence. After clocking the last bit, the received data in the Receive Shift Register (RSR) is transferred to the RCREG register (if it is empty). When the transfer is complete, interrupt flag bit, RCIF (PIR1<5>), is set. The actual interrupt can be enabled/ disabled by setting/clearing enable bit, RCIE (PIE1<5>). Flag bit RCIF is a read-only bit which is reset by the hardware. In this case, it is reset when the RCREG register has been read and is empty. The RCREG is a double-buffered register (i.e., it is a twodeep FIFO). It is possible for two bytes of data to be received and transferred to the RCREG FIFO and a third byte to begin shifting into the RSR register. On the clocking of the last bit of the third byte, if the RCREG register is still full, then Overrun Error bit, OERR (RCSTA<1>), is set. The word in the RSR will be lost. The RCREG register can be read twice to retrieve the two bytes in the FIFO. Bit OERR has to be cleared in software (by clearing bit CREN). If bit OERR is set, transfers from the RSR to the RCREG are inhibited so it is essential to clear bit OERR if it is set. The ninth receive bit is buffered the same way as the receive

data. Reading the RCREG register will load bit RX9D with a new value, therefore, it is essential for the user to read the RCSTA register before reading RCREG in order not to lose the old RX9D information.

When setting up a Synchronous Master Reception:

- 1. Initialize the SPBRG register for the appropriate baud rate (Section 10.1 "USART Baud Rate Generator (BRG)").
- 2. Enable the synchronous master serial port by setting bits SYNC, SPEN and CSRC.
- 3. Ensure bits CREN and SREN are clear.
- 4. If interrupts are desired, then set enable bit RCIE.
- 5. If 9-bit reception is desired, then set bit RX9.
- 6. If a single reception is required, set bit SREN. For continuous reception, set bit CREN.
- 7. Interrupt flag bit RCIF will be set when reception is complete and an interrupt will be generated if enable bit RCIE was set.
- 8. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
- 9. Read the 8-bit received data by reading the RCREG register.
- 10. If any error occurred, clear the error by clearing bit CREN.
- 11. If using interrupts, ensure that GIE and PEIE (bits 7 and 6) of the INTCON register are set.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	R0IF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
18h	RCSTA	SPEN	RX9	SREN	CREN		FERR	OERR	RX9D	0000 -00x	0000 -00x
1Ah	RCREG	USART Re	eceive Re	gister						0000 0000	0000 0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC		BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	RG Baud Rate Generator Register							0000 0000	0000 0000	

TABLE 10-9: REGISTERS ASSOCIATED WITH SYNCHRONOUS MASTER RECEPTION

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used for synchronous master reception.

Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices; always maintain these bits clear.

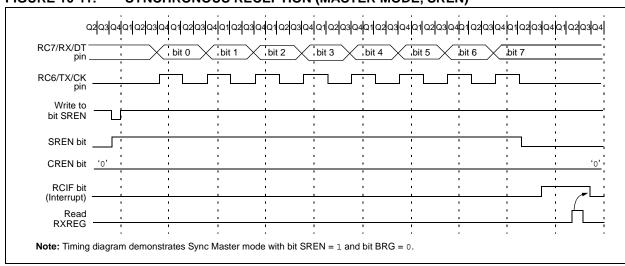


FIGURE 10-11: SYNCHRONOUS RECEPTION (MASTER MODE, SREN)

10.4 USART Synchronous Slave Mode

Synchronous Slave mode differs from the Master mode in the fact that the shift clock is supplied externally at the RC6/TX/CK pin (instead of being supplied internally in Master mode). This allows the device to transfer or receive data while in Sleep mode. Slave mode is entered by clearing bit, CSRC (TXSTA<7>).

10.4.1 USART SYNCHRONOUS SLAVE TRANSMIT

The operation of the Synchronous Master and Slave modes is identical, except in the case of the Sleep mode.

If two words are written to the TXREG and then the SLEEP instruction is executed, the following will occur:

- a) The first word will immediately transfer to the TSR register and transmit.
- b) The second word will remain in TXREG register.
- c) Flag bit TXIF will not be set.
- d) When the first word has been shifted out of TSR, the TXREG register will transfer the second word to the TSR and flag bit TXIF will now be set.
- e) If enable bit TXIE is set, the interrupt will wake the chip from Sleep and if the global interrupt is enabled, the program will branch to the interrupt vector (0004h).

When setting up a Synchronous Slave Transmission, follow these steps:

- Enable the synchronous slave serial port by setting bits SYNC and SPEN and clearing bit CSRC.
- 2. Clear bits CREN and SREN.
- If interrupts are desired, then set enable bit TXIE.
- 4. If 9-bit transmission is desired, then set bit TX9.
- 5. Enable the transmission by setting enable bit TXEN.
- 6. If 9-bit transmission is selected, the ninth bit should be loaded in bit TX9D.
- 7. Start transmission by loading data to the TXREG register.
- 8. If using interrupts, ensure that GIE and PEIE (bits 7 and 6) of the INTCON register are set.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value POR,	e on: BOR	all o	e on other sets
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	R0IF	0000	000x	0000	000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000	0000	0000	0000
18h	RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000	000x	0000	000x
19h	TXREG	USART Tr	ansmit R	egister						0000	0000	0000	0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000	0000	0000	0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	0000	-010	0000	-010
99h	SPBRG	G Baud Rate Generator Register								0000	0000	0000	0000

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used for synchronous slave transmission.

Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices; always maintain these bits clear.

10.4.2 USART SYNCHRONOUS SLAVE RECEPTION

The operation of the Synchronous Master and Slave modes is identical, except in the case of the Sleep mode. Bit SREN is a "don't care" in Slave mode.

If receive is enabled by setting bit CREN prior to the SLEEP instruction, then a word may be received during Sleep. On completely receiving the word, the RSR register will transfer the data to the RCREG register and if enable bit RCIE bit is set, the interrupt generated will wake the chip from Sleep. If the global interrupt is enabled, the program will branch to the interrupt vector (0004h).

When setting up a Synchronous Slave Reception, follow these steps:

- Enable the synchronous master serial port by setting bits SYNC and SPEN and clearing bit CSRC.
- 2. If interrupts are desired, set enable bit RCIE.
- 3. If 9-bit reception is desired, set bit RX9.
- 4. To enable reception, set enable bit CREN.
- 5. Flag bit RCIF will be set when reception is complete and an interrupt will be generated if enable bit RCIE was set.
- 6. Read the RCSTA register to get the ninth bit (if enabled) and determine if any error occurred during reception.
- 7. Read the 8-bit received data by reading the RCREG register.
- 8. If any error occurred, clear the error by clearing bit CREN.
- 9. If using interrupts, ensure that GIE and PEIE (bits 7 and 6) of the INTCON register are set.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on: POR, BOR	Value on all other Resets
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	R0IF	0000 000x	0000 000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000 0000	0000 0000
18h	RCSTA	SPEN	RX9	SREN	CREN	ADDEN	FERR	OERR	RX9D	0000 000x	x000 000x
1Ah	RCREG	USART R	eceive R	egister						0000 0000	0000 0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	TXIE	SSPIE	CCP1IE	TMR2IE	TMR1IE	0000 0000	0000 0000
98h	TXSTA	CSRC	TX9	TXEN	SYNC	_	BRGH	TRMT	TX9D	0000 -010	0000 -010
99h	SPBRG	Baud Rate	aud Rate Generator Register							0000 0000	0000 0000

TABLE 10-11: REGISTERS ASSOCIATED WITH SYNCHRONOUS SLAVE RECEPTION

Legend: x = unknown, - = unimplemented, read as '0'. Shaded cells are not used for synchronous slave reception.

Note 1: Bits PSPIE and PSPIF are reserved on 28-pin devices, always maintain these bits clear.

NOTES:

11.0 ANALOG-TO-DIGITAL CONVERTER (A/D) MODULE

The Analog-to-Digital (A/D) Converter module has five inputs for the 28-pin devices and eight for the 40/44-pin devices.

The conversion of an analog input signal results in a corresponding 10-bit digital number. The A/D module has high and low-voltage reference input that is software selectable to some combination of VDD, Vss, RA2 or RA3.

The A/D converter has a unique feature of being able to operate while the device is in Sleep mode. To operate in Sleep, the A/D clock must be derived from the A/D's internal RC oscillator. The A/D module has four registers. These registers are:

- A/D Result High Register (ADRESH)
- A/D Result Low Register (ADRESL)
- A/D Control Register 0 (ADCON0)
- A/D Control Register 1 (ADCON1)

The ADCON0 register, shown in Register 11-1, controls the operation of the A/D module. The ADCON1 register, shown in Register 11-2, configures the functions of the port pins. The port pins can be configured as analog inputs (RA3 can also be the voltage reference) or as digital I/O.

Additional information on using the A/D module can be found in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

REGISTER 11-1:	ADCON0 REGISTER (ADDRESS 1Fh)
----------------	-------------------------------

R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	R/W-0	U-0	R/W-0
ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/DONE		ADON
bit 7							bit 0

bit 7-6 ADCS1:ADCS0: A/D Conversion Clock Select bits (ADCON0 bits in **bold**)

ADCON1 <adcs2></adcs2>	ADCON0 <adcs1:adcs0></adcs1:adcs0>	Clock Conversion
0	00	Fosc/2
0	01	Fosc/8
0	10	Fosc/32
0	11	FRC (clock derived from the internal A/D RC oscillator)
1	00	Fosc/4
1	01	Fosc/16
1	10	Fosc/64
1	11	FRC (clock derived from the internal A/D RC oscillator)

bit 5-3 CHS2:CHS0: Analog Channel Select bits

- 000 = Channel 0 (AN0)
- 001 = Channel 1 (AN1)
- 010 = Channel 2 (AN2)
- 011 = Channel 3 (AN3)
- 100 = Channel 4 (AN4)
- 101 = Channel 5 (AN5)
- 110 = Channel 6 (AN6)
- 111 = Channel 7 (AN7)
 - **Note:** The PIC16F873A/876A devices only implement A/D channels 0 through 4; the unimplemented selections are reserved. Do not select any unimplemented channels with these devices.

bit 2 GO/DONE: A/D Conversion Status bit

When ADON = 1:

- 1 = A/D conversion in progress (setting this bit starts the A/D conversion which is automatically cleared by hardware when the A/D conversion is complete)
- 0 = A/D conversion not in progress
- bit 1 Unimplemented: Read as '0'

bit 0 ADON: A/D On bit

1 = A/D converter module is powered up

0 = A/D converter module is shut-off and consumes no operating current

Legend:

Legend:			
R = Readable bit	W = Writable bit	U = Unimplemented	bit, read as '0'
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown

REGISTER 11-2: ADCON1 REGISTER (ADDRESS 9Fh)

R/W-0	R/W-0	U-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0
ADFM	ADCS2	-	—	PCFG3	PCFG2	PCFG1	PCFG0
bit 7							bit 0

bit 7 ADFM: A/D Result Format Select bit

1 = Right justified. Six (6) Most Significant bits of ADRESH are read as '0'.
 0 = Left justified. Six (6) Least Significant bits of ADRESL are read as '0'.

bit 6 **ADCS2:** A/D Conversion Clock Select bit (ADCON1 bits in shaded area and in **bold**)

ADCON1 <adcs2></adcs2>	ADCON0 <adcs1:adcs0></adcs1:adcs0>	Clock Conversion
0	0 0	Fosc/2
0	01	Fosc/8
0	10	Fosc/32
0	11	FRC (clock derived from the internal A/D RC oscillator)
1	0 0	Fosc/4
1	01	Fosc/16
1	10	Fosc/64
1	11	FRC (clock derived from the internal A/D RC oscillator)

bit 5-4 Unimplemented: Read as '0'

bit 3-0 PCFG3:PCFG0: A/D Port Configuration Control bits

PCFG <3:0>	AN7	AN6	AN5	AN4	AN3	AN2	AN1	AN0	VREF+	VREF-	C/R
0000	Α	Α	Α	Α	А	А	Α	Α	Vdd	Vss	8/0
0001	А	А	А	Α	VREF+	А	А	А	AN3	Vss	7/1
0010	D	D	D	Α	А	А	А	А	Vdd	Vss	5/0
0011	D	D	D	Α	VREF+	А	Α	Α	AN3	Vss	4/1
0100	D	D	D	D	А	D	А	А	Vdd	Vss	3/0
0101	D	D	D	D	VREF+	D	Α	Α	AN3	Vss	2/1
011x	D	D	D	D	D	D	D	D	—	—	0/0
1000	А	А	А	Α	VREF+	VREF-	А	А	AN3	AN2	6/2
1001	D	D	А	А	А	А	Α	Α	Vdd	Vss	6/0
1010	D	D	А	Α	Vref+	А	А	А	AN3	Vss	5/1
1011	D	D	А	Α	VREF+	VREF-	А	А	AN3	AN2	4/2
1100	D	D	D	А	VREF+	VREF-	Α	Α	AN3	AN2	3/2
1101	D	D	D	D	VREF+	VREF-	А	А	AN3	AN2	2/2
1110	D	D	D	D	D	D	D	А	Vdd	Vss	1/0
1111	D	D	D	D	VREF+	VREF-	D	А	AN3	AN2	1/2

A = Analog input D = Digital I/O

C/R = # of analog input channels/# of A/D voltage references

Legend:

Legena.							
R = Readable bit	W = Writable bit	U = Unimplemented bit, read as '0'					
- n = Value at POR	'1' = Bit is set	'0' = Bit is cleared	x = Bit is unknown				

Note: On any device Reset, the port pins that are multiplexed with analog functions (ANx) are forced to be an analog input.

The ADRESH:ADRESL registers contain the 10-bit result of the A/D conversion. When the A/D conversion is complete, the result is loaded into this A/D Result register pair, the GO/DONE bit (ADCON0<2>) is cleared and the A/D interrupt flag bit ADIF is set. The block diagram of the A/D module is shown in Figure 11-1.

After the A/D module has been configured as desired, the selected channel must be acquired before the conversion is started. The analog input channels must have their corresponding TRIS bits selected as inputs.

To determine sample time, see **Section 11.1 "A/D Acquisition Requirements"**. After this acquisition time has elapsed, the A/D conversion can be started.

To do an A/D Conversion, follow these steps:

- 1. Configure the A/D module:
 - Configure analog pins/voltage reference and digital I/O (ADCON1)
 - Select A/D input channel (ADCON0)
 - Select A/D conversion clock (ADCON0)
 - Turn on A/D module (ADCON0)

- 2. Configure A/D interrupt (if desired):
 - Clear ADIF bit
 - Set ADIE bit
 - Set PEIE bit
 - Set GIE bit
- 3. Wait the required acquisition time.
- 4. Start conversion:
 - Set GO/DONE bit (ADCON0)
- 5. Wait for A/D conversion to complete by either:
 - Polling for the GO/DONE bit to be cleared (interrupts disabled); OR
 - Waiting for the A/D interrupt
- 6. Read A/D Result register pair (ADRESH:ADRESL), clear bit ADIF if required.
- 7. For the next conversion, go to step 1 or step 2 as required. The A/D conversion time per bit is defined as TAD.

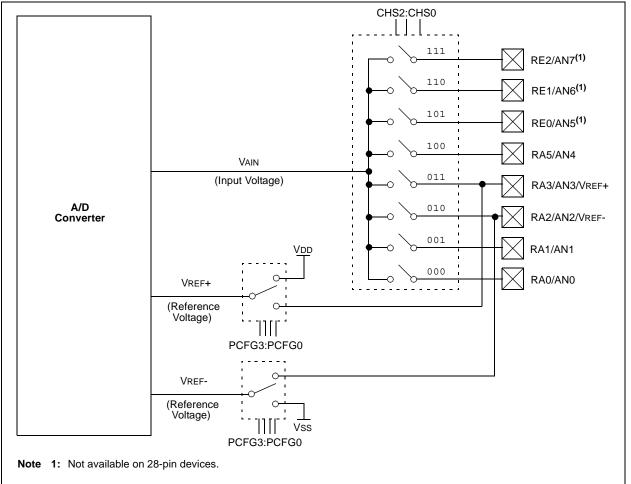


FIGURE 11-1: A/D BLOCK DIAGRAM

11.1 A/D Acquisition Requirements

For the A/D converter to meet its specified accuracy, the charge holding capacitor (CHOLD) must be allowed to fully charge to the input channel voltage level. The analog input model is shown in Figure 11-2. The source impedance (Rs) and the internal sampling switch impedance (Rss) directly affect the time required to charge the capacitor CHOLD. The sampling switch (Rss) impedance varies over the device voltage (VDD); see Figure 11-2. The maximum recommended impedance for analog sources is 2.5 k Ω . As the impedance is decreased, the acquisition time may be

decreased. After the analog input channel is selected (changed), this acquisition must be done before the conversion can be started.

To calculate the minimum acquisition time, Equation 11-1 may be used. This equation assumes that 1/2 LSb error is used (1024 steps for the A/D). The 1/2 LSb error is the maximum error allowed for the A/D to meet its specified resolution.

To calculate the minimum acquisition time, TACQ, see the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

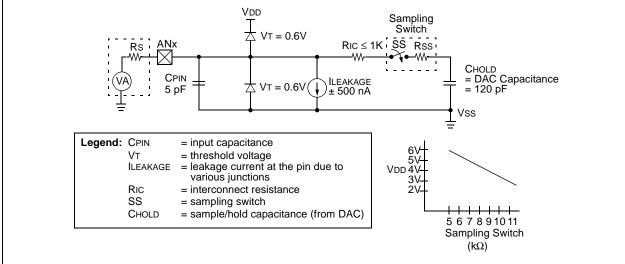
EQUATION 11-1: ACQUISITION TIME

TACQ	= Amplifier Settling Time + Hold Capacitor Charging Time + Temperature Coefficient
Тс	= TAMP + TC + TCOFF = $2 \mu s$ + TC + [(Temperature - 25°C)(0.05 $\mu s/^{\circ}$ C)] = CHOLD (RIC + RSS + RS) In(1/2047) = -120 pF (1 k Ω + 7 k Ω + 10 k Ω) In(0.0004885)
Tacq	= $16.47 \mu s$ = $2 \mu s + 16.47 \mu s + [(50^{\circ}C - 25^{\circ}C)(0.05 \mu s/^{\circ}C)]$ = $19.72 \mu s$

Note 1: The reference voltage (VREF) has no effect on the equation since it cancels itself out.

- 2: The charge holding capacitor (CHOLD) is not discharged after each conversion.
- **3:** The maximum recommended impedance for analog sources is 2.5 k Ω . This is required to meet the pin leakage specification.





11.2 Selecting the A/D Conversion Clock

The A/D conversion time per bit is defined as TAD. The A/D conversion requires a minimum 12 TAD per 10-bit conversion. The source of the A/D conversion clock is software selected. The seven possible options for TAD are:

- 2 Tosc
- 4 Tosc
- 8 Tosc
- 16 Tosc
- 32 Tosc
- 64 Tosc
- Internal A/D module RC oscillator (2-6 μs)

For correct A/D conversions, the A/D conversion clock (TAD) must be selected to ensure a minimum TAD time of 1.6 $\mu s.$

Table 11-1 shows the resultant TAD times derived from the device operating frequencies and the A/D clock source selected.

11.3 Configuring Analog Port Pins

The ADCON1 and TRIS registers control the operation of the A/D port pins. The port pins that are desired as analog inputs must have their corresponding TRIS bits set (input). If the TRIS bit is cleared (output), the digital output level (VOH or VOL) will be converted.

The A/D operation is independent of the state of the CHS2:CHS0 bits and the TRIS bits.

- Note 1: When reading the port register, any pin configured as an analog input channel will read as cleared (a low level). Pins configured as digital inputs will convert an analog input. Analog levels on a digitally configured input will not affect the conversion accuracy.
 - 2: Analog levels on any pin that is defined as a digital input (including the AN7:AN0 pins) may cause the input buffer to consume current that is out of the device specifications.

AD Clo	ck Source (TAD)	Maximum Davias Francesou
Operation	ADCS2:ADCS1:ADCS0	Maximum Device Frequency
2 Tosc	000	1.25 MHz
4 Tosc	100	2.5 MHz
8 Tosc	001	5 MHz
16 Tosc	101	10 MHz
32 Tosc	010	20 MHz
64 Tosc	110	20 MHz
RC ^(1, 2, 3)	x11	(Note 1)

TABLE 11-1: TAD VS. MAXIMUM DEVICE OPERATING FREQUENCIES (STANDARD DEVICES (F))

Note 1: The RC source has a typical TAD time of 4 μ s but can vary between 2-6 μ s.

2: When the device frequencies are greater than 1 MHz, the RC A/D conversion clock source is only recommended for Sleep operation.

3: For extended voltage devices (LF), please refer to Section 17.0 "Electrical Characteristics".

11.4 A/D Conversions

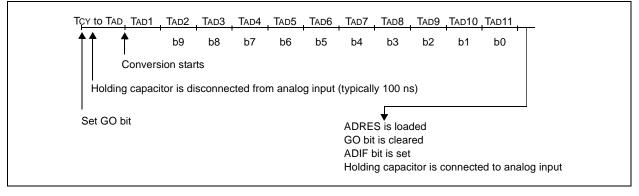
Clearing the GO/DONE bit during a conversion will abort the current conversion. The A/D Result register pair will NOT be updated with the partially completed A/D conversion sample. That is, the ADRESH:ADRESL registers will continue to contain the value of the last completed conversion (or the last value written to the ADRESH:ADRESL registers). After the A/D conversion

FIGURE 11-3: A/D CONVERSION TAD CYCLES

is aborted, the next acquisition on the selected channel is automatically started. The GO/DONE bit can then be set to start the conversion.

In Figure 11-3, after the GO bit is set, the first time segment has a minimum of TCY and a maximum of TAD.

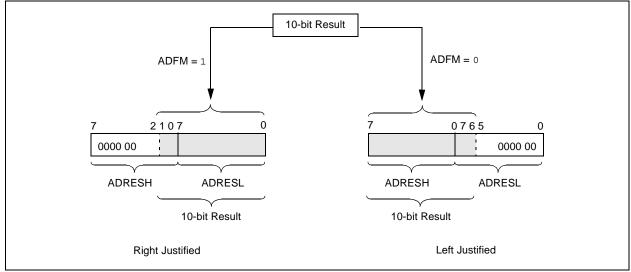
Note: The GO/DONE bit should NOT be set in the same instruction that turns on the A/D.



11.4.1 A/D RESULT REGISTERS

The ADRESH:ADRESL register pair is the location where the 10-bit A/D result is loaded at the completion of the A/D conversion. This register pair is 16 bits wide. The A/D module gives the flexibility to left or right justify the 10-bit result in the 16-bit result register. The A/D Format Select bit (ADFM) controls this justification. Figure 11-4 shows the operation of the A/D result justification. The extra bits are loaded with '0's. When an A/D result will not overwrite these locations (A/D disable), these registers may be used as two general purpose 8-bit registers.

FIGURE 11-4: A/D RESULT JUSTIFICATION



11.5 A/D Operation During Sleep

The A/D module can operate during Sleep mode. This requires that the A/D clock source be set to RC (ADCS1:ADCS0 = 11). When the RC clock source is selected, the A/D module waits one instruction cycle before starting the conversion. This allows the SLEEP instruction to be executed which eliminates all digital switching noise from the conversion. When the conversion is completed, the GO/DONE bit will be cleared and the result loaded into the ADRES register. If the A/D interrupt is enabled, the device will wake-up from Sleep. If the A/D interrupt is not enabled, the ADON bit will remain set.

When the A/D clock source is another clock option (not RC), a SLEEP instruction will cause the present conversion to be aborted and the A/D module to be turned off, though the ADON bit will remain set.

Turning off the A/D places the A/D module in its lowest current consumption state.

Note:	For the A/D module to operate in Sleep,
	the A/D clock source must be set to RC
	(ADCS1:ADCS0 = 11). To allow the con-
	version to occur during Sleep, ensure the
	SLEEP instruction immediately follows the
	instruction that sets the GO/DONE bit.

11.6 Effects of a Reset

A device Reset forces all registers to their Reset state. This forces the A/D module to be turned off and any conversion is aborted. All A/D input pins are configured as analog inputs.

The value that is in the ADRESH:ADRESL registers is not modified for a Power-on Reset. The ADRESH:ADRESL registers will contain unknown data after a Power-on Reset.

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value POR, I		<u>Valu</u> MCLR	e on , WDT
0Bh,8Bh, 10Bh,18Bh	INTCON	GIE	PEIE	TMR0IE	INTE	RBIE	TMR0IF	INTF	RBIF	0000 (000x	0000	000u
0Ch	PIR1	PSPIF ⁽¹⁾	ADIF	RCIF	TXIF	SSPIF	CCP1IF	TMR2IF	TMR1IF	0000	0000	0000	0000
8Ch	PIE1	PSPIE ⁽¹⁾	ADIE	RCIE	RCIE TXIE SSPIE CCP1IE TMR2IE TMR1IE						0000	0000	0000
1Eh	ADRESH	A/D Resu	A/D Result Register High Byte									uuuu	uuuu
9Eh	ADRESL	A/D Resu	lt Registe	r Low Byte	•					XXXX X	xxxx	uuuu	uuuu
1Fh	ADCON0	ADCS1	ADCS0	CHS2	CHS1	CHS0	GO/DONE	-	ADON	0000	0 0 - 0	0000	00-0
9Fh	ADCON1	ADFM	ADCS2	_	_	PCFG3	PCFG2	PCFG1	PCFG0	00 0	0000	00	0000
85h	TRISA	_	_	PORTA D	ata Direction	Register				11 3	1111	11	1111
05h	PORTA	_	_	PORTA D	ata Latch wh	en written	: PORTA pin	s when re	ad	0x (0000	0u	0000
89h ⁽¹⁾	TRISE	IBF	OBF	IBOV	IBOV PSPMODE — PORTE Data Direction bits						-111	0000	-111
09h ⁽¹⁾	PORTE	—	—	—	_	—	RE2	RE1	RE0		-xxx		-uuu

 TABLE 11-2:
 REGISTERS/BITS ASSOCIATED WITH A/D

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used for A/D conversion.

Note 1: These registers are not available on 28-pin devices.

NOTES:

12.0 COMPARATOR MODULE

The comparator module contains two analog comparators. The inputs to the comparators are multiplexed with I/O port pins RA0 through RA3, while the outputs are multiplexed to pins RA4 and RA5. The on-chip voltage reference (Section 13.0 "Comparator Voltage Reference Module") can also be an input to the comparators. The CMCON register (Register 12-1) controls the comparator input and output multiplexers. A block diagram of the various comparator configurations is shown in Figure 12-1.

	R-0	R-0	R/W-0	R/W-0	R/W-0	R/W-1	R/W-1	R/W-1
	C2OUT	C1OUT	C2INV	C1INV	CIS	CM2	CM1	CM0
	bit 7							bit 0
bit 7		omparator 2	Output bit					
	<u>When C2IN</u> 1 = C2 V(N)	<u>IV = 0:</u> + > C2 VIN-						
		+ > C2 VIN- + < C2 VIN-						
	When C2IN	V = 1:						
		+ < C2 VIN-						
		+ > C2 VIN-						
bit 6		omparator 1	Output bit					
	<u>When C1IN</u> 1 = C1 VIN	<u>NV = 0:</u> + > C1 VIN-						
		+ < C1 VIN-						
	When C1IN	V = 1:						
		+ < C1 VIN-						
		+ > C1 VIN-						
bit 5		mparator 2 (Jutput Inver	sion bit				
		out inverted out not inver	ted					
bit 4	•	mparator 1 (sion bit				
		out inverted						
		out not inver	ted					
bit 3	CIS: Comp	arator Input	Switch bit					
		2:CM0 = 110						
		I- connects t I- connects t						
		- connects t						
	C2 VIN	I- connects t	o RA1/AN1					
bit 2	CM2:CM0:	Comparato	r Mode bits					
	Figure 12-1	I shows the	Comparator	modes and	CM2:CM0	bit settings.		
	Legend:							
	R = Reada	ble bit	W = W	ritable bit	U = Unin	nplemented	bit, read as	'0'
	- n = Value	at POR	'1' = B	it is set	'0' = Bit i	s cleared	x = Bit is u	nknown

REGISTER 12-1: CMCON REGISTER

12.1 Comparator Configuration

There are eight modes of operation for the comparators. The CMCON register is used to select these modes. Figure 12-1 shows the eight possible modes. The TRISA register controls the data direction of the comparator pins for each mode. If the Comparator mode is changed, the comparator output level may not be valid for the specified mode change delay shown in **Section 17.0 "Electrical Characteristics"**.

Note:	Compa	rato	or interrupts sh	nould be	disabled					
	during	а	Comparator	mode change.						
	Otherwise, a false interrupt may occur.									

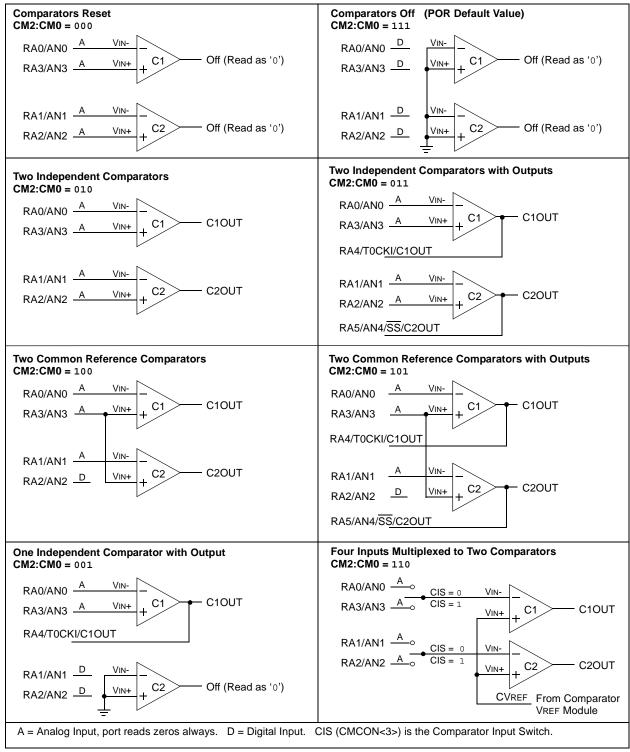


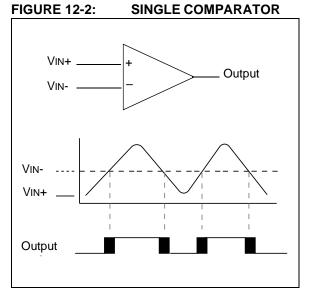
FIGURE 12-1: COMPARATOR I/O OPERATING MODES

12.2 Comparator Operation

A single comparator is shown in Figure 12-2 along with the relationship between the analog input levels and the digital output. When the analog input at VIN+ is less than the analog input VIN-, the output of the comparator is a digital low level. When the analog input at VIN+ is greater than the analog input VIN-, the output of the comparator is a digital high level. The shaded areas of the output of the comparator in Figure 12-2 represent the uncertainty due to input offsets and response time.

12.3 Comparator Reference

An external or internal reference signal may be used depending on the comparator operating mode. The analog signal present at VIN- is compared to the signal at VIN+ and the digital output of the comparator is adjusted accordingly (Figure 12-2).



12.3.1 EXTERNAL REFERENCE SIGNAL

When external voltage references are used, the comparator module can be configured to have the comparators operate from the same or different reference sources. However, threshold detector applications may require the same reference. The reference signal must be between VSS and VDD and can be applied to either pin of the comparator(s).

12.3.2 INTERNAL REFERENCE SIGNAL

The comparator module also allows the selection of an internally generated voltage reference for the comparators. Section 13.0 "Comparator Voltage Reference Module" contains a detailed description of the Comparator Voltage Reference module that provides this signal. The internal reference signal is used when comparators are in mode, CM<2:0> = 110 (Figure 12-1). In this mode, the internal voltage reference is applied to the VIN+ pin of both comparators.

12.4 Comparator Response Time

Response time is the minimum time, after selecting a new reference voltage or input source, before the comparator output has a valid level. If the internal reference is changed, the maximum delay of the internal voltage reference must be considered when using the comparator outputs. Otherwise, the maximum delay of the comparators should be used (Section 17.0 "Electrical Characteristics").

12.5 Comparator Outputs

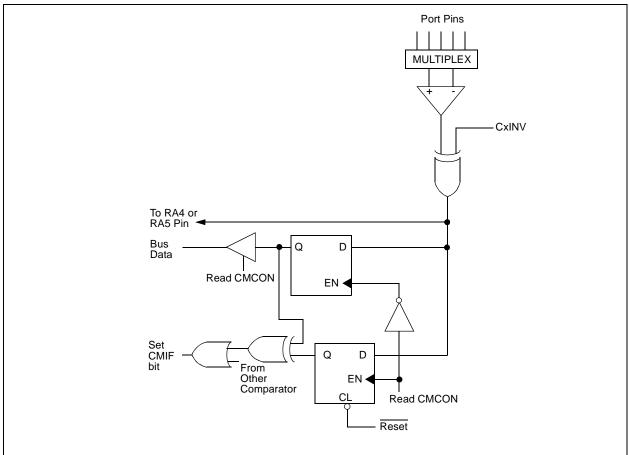
The comparator outputs are read through the CMCON register. These bits are read-only. The comparator outputs may also be directly output to the RA4 and RA5 I/O pins. When enabled, multiplexors in the output path of the RA4 and RA5 pins will switch and the output of each pin will be the unsynchronized output of the comparator. The uncertainty of each of the comparators is related to the input offset voltage and the response time given in the specifications. Figure 12-3 shows the comparator output block diagram.

The TRISA bits will still function as an output enable/ disable for the RA4 and RA5 pins while in this mode.

The polarity of the comparator outputs can be changed using the C2INV and C1INV bits (CMCON<4:5>).

- Note 1: When reading the Port register, all pins configured as analog inputs will read as a '0'. Pins configured as digital inputs will convert an analog input according to the Schmitt Trigger input specification.
 - 2: Analog levels on any pin defined as a digital input may cause the input buffer to consume more current than is specified.
 - **3:** RA4 is an open collector I/O pin. When used as an output, a pull-up resistor is required.





12.6 Comparator Interrupts

The comparator interrupt flag is set whenever there is a change in the output value of either comparator. Software will need to maintain information about the status of the output bits, as read from CMCON<7:6>, to determine the actual change that occurred. The CMIF bit (PIR registers) is the Comparator Interrupt Flag. The CMIF bit must be reset by clearing it ('0'). Since it is also possible to write a '1' to this register, a simulated interrupt may be initiated.

The CMIE bit (PIE registers) and the PEIE bit (INTCON register) must be set to enable the interrupt. In addition, the GIE bit must also be set. If any of these bits are clear, the interrupt is not enabled, though the CMIF bit will still be set if an interrupt condition occurs.

Note: If a change in the CMCON register (C1OUT or C2OUT) should occur when a read operation is being executed (start of the Q2 cycle), then the CMIF (PIR registers) interrupt flag may not get set.

The user, in the Interrupt Service Routine, can clear the interrupt in the following manner:

- a) Any read or write of CMCON will end the mismatch condition.
- b) Clear flag bit CMIF.

A mismatch condition will continue to set flag bit CMIF. Reading CMCON will end the mismatch condition and allow flag bit CMIF to be cleared.

12.7 Comparator Operation During Sleep

When a comparator is active and the device is placed in Sleep mode, the comparator remains active and the interrupt is functional if enabled. This interrupt will wake-up the device from Sleep mode when enabled. While the comparator is powered up, higher Sleep currents than shown in the power-down current specification will occur. Each operational comparator will consume additional current as shown in the comparator specifications. To minimize power consumption while in Sleep mode, turn off the comparators, CM<2:0> = 111, before entering Sleep. If the device wakes up from Sleep, the contents of the CMCON register are not affected.

12.8 Effects of a Reset

A device Reset forces the CMCON register to its Reset state, causing the comparator module to be in the Comparator Off mode, CM<2:0> = 111. This ensures compatibility to the PIC16F87X devices.

12.9 Analog Input Connection Considerations

A simplified circuit for an analog input is shown in Figure 12-4. Since the analog pins are connected to a digital output, they have reverse biased diodes to VDD and Vss. The analog input, therefore, must be between Vss and VDD. If the input voltage deviates from this range by more than 0.6V in either direction, one of the diodes is forward biased and a latch-up condition may occur. A maximum source impedance of 10 k Ω is recommended for the analog sources. Any external component connected to an analog input pin, such as a capacitor or a Zener diode, should have very little leakage current.

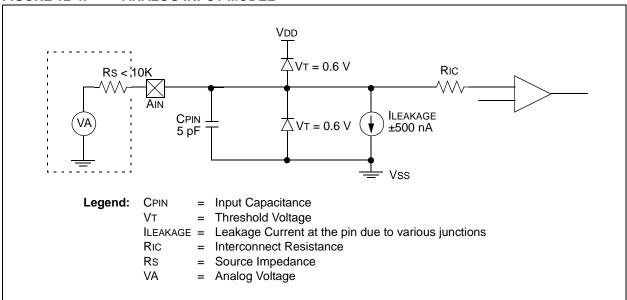


FIGURE 12-4: ANALOG INPUT MODEL

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR	Value on all other Resets
9Ch	CMCON	C2OUT	C1OUT	C2INV	C1INV	CIS	CM2	CM1	CM0	0000 0111	0000 0111
9Dh	CVRCON	CVREN	CVROE	CVRR	_	CVR3	CVR2	CVR1	CVR0	000- 0000	000- 0000
0Bh, 8Bh, 10Bh,18Bh	INTCON	GIE/ GIEH	PEIE/ GIEL	TMR0IE	INTIE	RBIE	TMR0IF	INTIF	RBIF	x000 000x	0000 000u
0Dh	PIR2	—	CMIF	—	_	BCLIF	LVDIF	TMR3IF	CCP2IF	-0 0000	-0 0000
8Dh	PIE2	—	CMIE	—	_	BCLIE	LVDIE	TMR3IE	CCP2IE	-0 0000	-0 0000
05h	PORTA	—	_	RA5	RA4	RA3	RA2	RA1	RA0	0x 0000	0u 0000
85h	TRISA	_	_	PORTA D	ata Direc	tion Regis	11 1111	11 1111			

TABLE 12-1: REGISTERS ASSOCIATED WITH COMPARATOR MODULE

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are unused by the comparator module.

13.0 COMPARATOR VOLTAGE REFERENCE MODULE

The Comparator Voltage Reference Generator is a 16-tap resistor ladder network that provides a fixed voltage reference when the comparators are in mode '110'. A programmable register controls the function of the reference generator. Register 13-1 lists the bit functions of the CVRCON register.

As shown in Figure 13-1, the resistor ladder is segmented to provide two ranges of CVREF values and has a power-down function to conserve power when the reference is not being used. The comparator reference

bit

bit

bit

bit bit supply voltage (also referred to as CVRSRC) comes directly from VDD. It should be noted, however, that the voltage at the top of the ladder is CVRSRC – VSAT, where VSAT is the saturation voltage of the power switch transistor. This reference will only be as accurate as the values of CVRSRC and VSAT.

The output of the reference generator may be connected to the RA2/AN2/VREF-/CVREF pin. This can be used as a simple D/A function by the user if a very highimpedance load is used. The primary purpose of this function is to provide a test path for testing the reference generator function.

REGISTER 13-1: CVRCON CONTROL REGISTER (ADDRESS 9Dh)

R/W-0	R/W-0	R/W-0	U-0	R/W-0	R/W-0	R/W-0	R/W-0				
CVREN	CVROE	CVRR		CVR3	CVR2	CVR1	CVR0				
bit 7	·						bit 0				
CVREN: (Comparator V	oltage Refe	rence Enabl	e bit							
	F circuit powe										
CVROE: (Comparator V	REF Output	Enable bit								
	•										
CVRR: Co											
	1 = 0 to 0.75 CVRsRc, with CVRsRc/24 step size 0 = 0.25 CVRsRc to 0.75 CVRsRc, with CVRsRc/32 step size										
Unimplen	nented: Read	d as '0'									
CVR3:CV	R0: Compara	ator VREF Va	lue Selectio	n bits $0 \le VF$	R3:VR0 ≤ 15	5					
<u>When CVI</u> CVREF = () • (CVRSRC)								
	CVREF = (VR<3:0>/ 24) • (CVRSRC) <u>When CVRR = 0:</u> CVREF = 1/4 • (CVRSRC) + (VR3:VR0/ 32) • (CVRSRC)										
Legend:											
R = Reada	able bit	W = W	ritable bit	U = Unim	plemented	bit, read as '	0'				
-n = Value	at POR	'1' = Bi	t is set	'0' = Bit is	s cleared	x = Bit is u	nknown				

FIGURE 13-1: COMPARATOR VOLTAGE REFERENCE BLOCK DIAGRAM

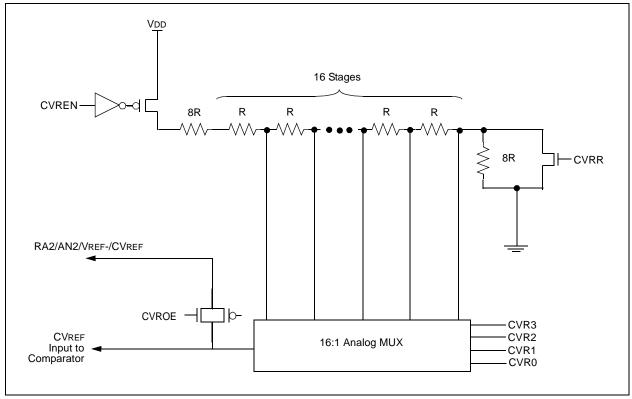


TABLE 13-1: REGISTERS ASSOCIATED WITH COMPARATOR VOLTAGE REFERENCE

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Value on POR	Value on all other Resets	
9Dh	CVRCON	CVREN	CVROE	CVRR		CVR3	CVR2	CVR1	CVR0	000- 0000	000- 0000	
9Ch	CMCON	C2OUT	C1OUT	C2INV	C1INV	CIS	CM2	CM1	CM0	0000 0111	0000 0111	

Legend: x = unknown, u = unchanged, - = unimplemented, read as '0'. Shaded cells are not used with the comparator voltage reference.

14.0 SPECIAL FEATURES OF THE CPU

All PIC16F87XA devices have a host of features intended to maximize system reliability, minimize cost through elimination of external components, provide power saving operating modes and offer code protection. These are:

- Oscillator Selection
- Reset
 - Power-on Reset (POR)
 - Power-up Timer (PWRT)
 - Oscillator Start-up Timer (OST)
 - Brown-out Reset (BOR)
- Interrupts
- Watchdog Timer (WDT)
- Sleep
- Code Protection
- ID Locations
- In-Circuit Serial Programming
- Low-Voltage In-Circuit Serial Programming
- In-Circuit Debugger

PIC16F87XA devices have a Watchdog Timer which can be shut-off only through configuration bits. It runs off its own RC oscillator for added reliability.

There are two timers that offer necessary delays on power-up. One is the Oscillator Start-up Timer (OST), intended to keep the chip in Reset until the crystal oscillator is stable. The other is the Power-up Timer (PWRT), which provides a fixed delay of 72 ms (nominal) on power-up only. It is designed to keep the part in Reset while the power supply stabilizes. With these two timers on-chip, most applications need no external Reset circuitry. Sleep mode is designed to offer a very low current power-down mode. The user can wake-up from Sleep through external Reset, Watchdog Timer wake-up or through an interrupt.

Several oscillator options are also made available to allow the part to fit the application. The RC oscillator option saves system cost while the LP crystal option saves power. A set of configuration bits is used to select various options.

Additional information on special features is available in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

14.1 Configuration Bits

The configuration bits can be programmed (read as '0'), or left unprogrammed (read as '1') to select various device configurations. The erased or unprogrammed value of the Configuration Word register is 3FFFh. These bits are mapped in program memory location 2007h.

It is important to note that address 2007h is beyond the user program memory space which can be accessed only during programming.

CONFIGURATION WORD (ADDRESS 2007h)⁽¹⁾ REGISTER 14-1: R/P-1 **R/P-1 R/P-1** R/P-1 **R/P-1** R/P-1 R/P-1 **R/P-1** R/P-1 U-0 R/P-1 R/P-1 U-0 U-0 CP DEBUG WRT1 WRT0 CPD LVP BOREN PWRTEN WDTEN Fosc1 Fosc0 bit 13 bit0 bit 13 CP: Flash Program Memory Code Protection bit 1 = Code protection off0 = All program memory code-protected bit 12 Unimplemented: Read as '1' bit 11 DEBUG: In-Circuit Debugger Mode bit 1 = In-Circuit Debugger disabled, RB6 and RB7 are general purpose I/O pins 0 = In-Circuit Debugger enabled, RB6 and RB7 are dedicated to the debugger bit 10-9 WRT1:WRT0 Flash Program Memory Write Enable bits For PIC16F876A/877A: 11 = Write protection off; all program memory may be written to by EECON control 10 = 0000h to 00FFh write-protected; 0100h to 1FFFh may be written to by EECON control 01 = 0000h to 07FFh write-protected; 0800h to 1FFFh may be written to by EECON control 00 = 0000h to 0FFFh write-protected; 1000h to 1FFFh may be written to by EECON control For PIC16F873A/874A: 11 = Write protection off; all program memory may be written to by EECON control 10 = 0000h to 00FFh write-protected; 0100h to 0FFFh may be written to by EECON control 01 = 0000h to 03FFh write-protected; 0400h to 0FFFh may be written to by EECON control 00 = 0000h to 07FFh write-protected; 0800h to 0FFFh may be written to by EECON control bit 8 CPD: Data EEPROM Memory Code Protection bit 1 = Data EEPROM code protection off 0 = Data EEPROM code-protected LVP: Low-Voltage (Single-Supply) In-Circuit Serial Programming Enable bit bit 7 1 = RB3/PGM pin has PGM function; low-voltage programming enabled 0 = RB3 is digital I/O, HV on MCLR must be used for programming bit 6 BOREN: Brown-out Reset Enable bit 1 = BOR enabled 0 = BOR disabled bit 5-4 Unimplemented: Read as '1' bit 3 **PWRTEN:** Power-up Timer Enable bit 1 = PWRT disabled 0 = PWRT enabled bit 2 WDTEN: Watchdog Timer Enable bit 1 = WDT enabled 0 = WDT disabled bit 1-0 Fosc1:Fosc0: Oscillator Selection bits 11 = RC oscillator 10 = HS oscillator 01 = XT oscillator 00 = LP oscillator

Legend:

R = Readable bit P = Programmable bit - n = Value when device is unprogrammed U = Unimplemented bit, read as '0'

u = Unchanged from programmed state

Note 1: The erased (unprogrammed) value of the Configuration Word is 3FFFh.

14.2 Oscillator Configurations

14.2.1 OSCILLATOR TYPES

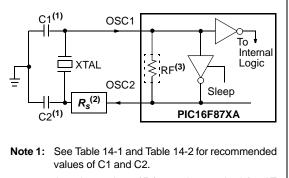
The PIC16F87XA can be operated in four different oscillator modes. The user can program two configuration bits (FOsc1 and FOsc0) to select one of these four modes:

- LP Low-Power Crystal
- XT Crystal/Resonator
- HS High-Speed Crystal/Resonator
- RC Resistor/Capacitor

14.2.2 CRYSTAL OSCILLATOR/CERAMIC RESONATORS

In XT, LP or HS modes, a crystal or ceramic resonator is connected to the OSC1/CLKI and OSC2/CLKO pins to establish oscillation (Figure 14-1). The PIC16F87XA oscillator design requires the use of a parallel cut crystal. Use of a series cut crystal may give a frequency out of the crystal manufacturer's specifications. When in XT, LP or HS modes, the device can have an external clock source to drive the OSC1/CLKI pin (Figure 14-2).

FIGURE 14-1: CRYSTAL/CERAMIC RESONATOR OPERATION (HS, XT OR LP OSC CONFIGURATION)



- **2:** A series resistor (*R*_s) may be required for AT strip cut crystals.
- 3: RF varies with the crystal chosen.

FIGURE 14-2:

EXTERNAL CLOCK INPUT OPERATION (HS, XT OR LP OSC CONFIGURATION)

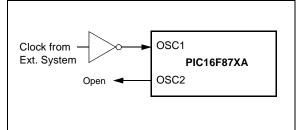


TABLE 14-1: CERAMIC RESONATORS

Ranges Tested:									
Mode	Mode Freq. OSC1 OSC2								
XT	455 kHz	68-100 pF	68-100 pF						
	2.0 MHz	15-68 pF	15-68 pF						
	4.0 MHz	15-68 pF	15-68 pF						
HS	8.0 MHz	10-68 pF	10-68 pF						
	16.0 MHz	10-22 pF	10-22 pF						

These values are for design guidance only. See notes following Table 14-2.

Resonators Used:						
2.0 MHz	Murata Erie CSA2.00MG	$\pm 0.5\%$				
4.0 MHz	Murata Erie CSA4.00MG	± 0.5%				
8.0 MHz	Murata Erie CSA8.00MT	± 0.5%				
16.0 MHz	Murata Erie CSA16.00MX	$\pm 0.5\%$				
All resonators used did not have built-in capacitors.						

TABLE 14-2:CAPACITOR SELECTION FOR
CRYSTAL OSCILLATOR

Osc Type	Crystal Freq.	Cap. Range C1	Cap. Range C2
LP	32 kHz	33 pF	33 pF
	200 kHz	15 pF	15 pF
XT	200 kHz	47-68 pF	47-68 pF
	1 MHz	15 pF	15 pF
	4 MHz	15 pF	15 pF
HS	4 MHz	15 pF	15 pF
	8 MHz	15-33 pF	15-33 pF
	20 MHz	15-33 pF	15-33 pF

These values are for design guidance only. See notes following this table.

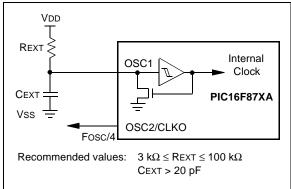
Crystals Used						
32 kHz	Epson C-001R32.768K-A	± 20 PPM				
200 kHz	STD XTL 200.000KHz	± 20 PPM				
1 MHz	ECS ECS-10-13-1	± 50 PPM				
4 MHz	ECS ECS-40-20-1	± 50 PPM				
8 MHz	EPSON CA-301 8.000M-C	± 30 PPM				
20 MHz	EPSON CA-301 20.000M-C	± 30 PPM				

- Note 1: Higher capacitance increases the stability of oscillator but also increases the start-up time.
 - Since each resonator/crystal has its own characteristics, the user should consult the resonator/crystal manufacturer for appropriate values of external components.
 - **3:** *R*_s may be required in HS mode, as well as XT mode, to avoid overdriving crystals with low drive level specification.
 - 4: When migrating from other PICmicro[®] devices, oscillator performance should be verified.

14.2.3 RC OSCILLATOR

For timing insensitive applications, the "RC" device option offers additional cost savings. The RC oscillator frequency is a function of the supply voltage, the resistor (REXT) and capacitor (CEXT) values and the operating temperature. In addition to this, the oscillator frequency will vary from unit to unit due to normal process parameter variation. Furthermore, the difference in lead frame capacitance between package types will also affect the oscillation frequency, especially for low CEXT values. The user also needs to take into account variation due to tolerance of external R and C components used. Figure 14-3 shows how the R/C combination is connected to the PIC16F87XA.

FIGURE 14-3: RC OSCILLATOR MODE



14.3 Reset

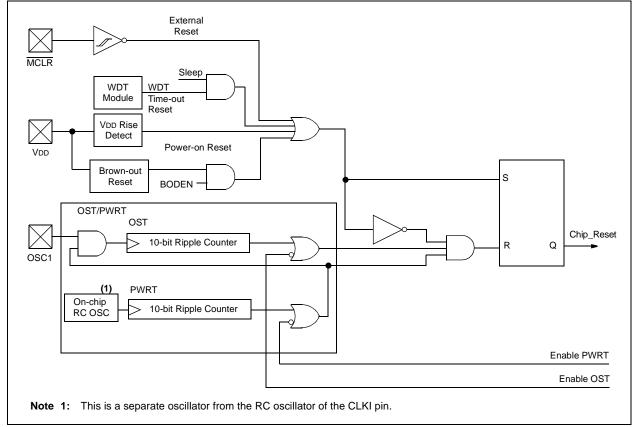
The PIC16F87XA differentiates between various kinds of Reset:

- Power-on Reset (POR)
- MCLR Reset during normal operation
- MCLR Reset during Sleep
- WDT Reset (during normal operation)
- WDT Wake-up (during Sleep)
- Brown-out Reset (BOR)

Some registers are not affected in any Reset condition. Their status is unknown on POR and unchanged in any other Reset. Most other registers are reset to a "Reset state" on Power-on Reset (POR), on the MCLR and WDT Reset, on MCLR Reset during Sleep and Brownout Reset (BOR). They are not affected by a WDT wake-up which is viewed as the resumption of normal operation. The TO and PD bits are set or cleared differently in different Reset situations as indicated in Table 14-4. These bits are used in software to determine the nature of the Reset. See Table 14-6 for a full description of Reset states of all registers.

A simplified block diagram of the on-chip Reset circuit is shown in Figure 14-4.

FIGURE 14-4: SIMPLIFIED BLOCK DIAGRAM OF ON-CHIP RESET CIRCUIT



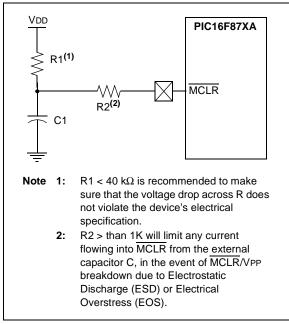
14.4 MCLR

PIC16F87XA devices have a noise filter in the $\overline{\text{MCLR}}$ Reset path. The filter will detect and ignore small pulses.

It should be noted that a WDT Reset does not drive $\overline{\text{MCLR}}$ pin low.

The behavior of the ESD protection on the MCLR pin differs from previous devices of this family. Voltages applied to the pin that exceed its specification can result in both Resets and current consumption outside of device specification during the Reset event. For this reason, Microchip recommends that the MCLR pin no longer be tied directly to VDD. The use of an RCR network, as shown in Figure 14-5, is suggested.

FIGURE 14-5: RECOMMENDED MCLR CIRCUIT



14.5 Power-on Reset (POR)

A Power-on Reset pulse is generated on-chip when VDD rise is detected (in the range of 1.2V-1.7V). To take advantage of the POR, tie the MCLR pin to VDD through an RC network, as described in **Section 14.4** "**MCLR**". A maximum rise time for VDD is specified. See **Section 17.0** "**Electrical Characteristics**" for details.

When the device starts normal operation (exits the Reset condition), device operating parameters (voltage, frequency, temperature, etc.) must be met to ensure operation. If these conditions are not met, the device must be held in Reset until the operating conditions are met. Brown-out Reset may be used to meet the start-up conditions. For additional information, refer to application note, *AN607, "Power-up Trouble Shooting*" (DS00607).

14.6 Power-up Timer (PWRT)

The Power-up Timer provides a fixed 72 ms nominal time-out on power-up only from the POR. The Power-up Timer operates on an internal RC oscillator. The chip is kept in Reset as long as the PWRT is active. The PWRT's time delay allows VDD to rise to an acceptable level. A configuration bit is provided to enable or disable the PWRT.

The power-up time delay will vary from chip to chip due to VDD, temperature and process variation. See **Section 17.0 "Electrical Characteristics"** for details (TPWRT, parameter #33).

14.7 Oscillator Start-up Timer (OST)

The Oscillator Start-up Timer (OST) provides a delay of 1024 oscillator cycles (from OSC1 input) after the PWRT delay is over (if PWRT is enabled). This helps to ensure that the crystal oscillator or resonator has started and stabilized.

The OST time-out is invoked only for XT, LP and HS modes and only on Power-on Reset or wake-up from Sleep.

14.8 Brown-out Reset (BOR)

The configuration bit, BODEN, can enable or disable the Brown-out Reset circuit. If VDD falls below VBOR (parameter D005, about 4V) for longer than TBOR (parameter #35, about 100 μ S), the brown-out situation will reset the device. If VDD falls below VBOR for less than TBOR, a Reset may not occur.

Once the brown-out occurs, the device will remain in Brown-out Reset until VDD rises above VBOR. The Power-up Timer then keeps the device in Reset for TPWRT (parameter #33, about 72 mS). If VDD should fall below VBOR during TPWRT, the Brown-out Reset process will restart when VDD rises above VBOR with the Power-up Timer Reset. The Power-up Timer is always enabled when the Brown-out Reset circuit is enabled, regardless of the state of the PWRT configuration bit.

14.9 Time-out Sequence

On power-up, the time-out sequence is as follows: the PWRT delay starts (if enabled) when a POR Reset occurs. Then, OST starts counting 1024 oscillator cycles when PWRT ends (LP, XT, HS). When the OST ends, the device comes out of Reset.

If MCLR is kept low long enough, the time-outs will expire. Bringing MCLR high will begin execution immediately. This is useful for testing purposes or to synchronize more than one PIC16F87XA device operating in parallel.

Table 14-5 shows the Reset conditions for the Status, PCON and PC registers, while Table 14-6 shows the Reset conditions for all the registers.

14.10 Power Control/Status Register (PCON)

The Power Control/Status Register, PCON, has up to two bits depending upon the device.

Bit 0 is the Brown-out Reset Status bit, BOR. The BOR bit is unknown on a Power-on Reset. It must then be set by the user and checked on subsequent Resets to see if it has been cleared, indicating that a BOR has occurred.

When the Brown-out Reset is disabled, the state of the BOR bit is unpredictable and is, therefore, not valid at any time.

Bit 1 is the Power-on Reset Status bit, $\overline{\text{POR}}$. It is cleared on a Power-on Reset and unaffected otherwise. The user must set this bit following a Power-on Reset.

TABLE 14-3: TIME-OUT IN VARIOUS SITUATIONS

Oscillator Configuration	Power	·up	Brown-out	Wake-up from Sleep	
	PWRTE = 0	PWRTE = 1	Brown-out		
XT, HS, LP	72 ms + 1024 Tosc	1024 Tosc	72 ms + 1024 Tosc	1024 Tosc	
RC	72 ms		72 ms	_	

TABLE 14-4: STATUS BITS AND THEIR SIGNIFICANCE

POR	BOR	то	PD	Condition
0	x	1	1	Power-on Reset
0	x	0	x	Illegal, TO is set on POR
0	x	x	0	Illegal, PD is set on POR
1	0	1	1	Brown-out Reset
1	1	0	1	WDT Reset
1	1	0	0	WDT Wake-up
1	1	u	u	MCLR Reset during normal operation
1	1	1	0	MCLR Reset during Sleep or Interrupt Wake-up from Sleep

Legend: x = don't care, u = unchanged

TABLE 14-5: RESET CONDITIONS FOR SPECIAL REGISTERS

Condition	Program Counter	Status Register	PCON Register
Power-on Reset	000h	0001 1xxx	0x
MCLR Reset during normal operation	000h	000u uuuu	uu
MCLR Reset during Sleep	000h	0001 0uuu	uu
WDT Reset	000h	0000 luuu	uu
WDT Wake-up	PC + 1	uuu0 0uuu	uu
Brown-out Reset	000h	0001 luuu	u0
Interrupt Wake-up from Sleep	PC + 1 ⁽¹⁾	uuul Ouuu	uu

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0'

Note 1: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

					ITIONS FOR ALL RE Power-on Reset,	MCLR Resets,	Wake-up via WDT or
Register	Devices			Brown-out Reset	WDT Reset	Interrupt	
W	73A	74A	76A	77A	xxxx xxxx	uuuu uuuu	uuuu uuuu
INDF	73A	74A	76A	77A	N/A	N/A	N/A
TMR0	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
PCL	73A	74A	76A	77A	0000 0000	0000 0000	PC + 1 ⁽²⁾
STATUS	73A	74A	76A	77A	0001 1xxx	000q quuu (3)	uuuq quuu ⁽³⁾
FSR	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
PORTA	73A	74A	76A	77A	0x 0000	0u 0000	uu uuuu
PORTB	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
PORTC	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
PORTD	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
PORTE	73A	74A	76A	77A	xxx	uuu	uuu
PCLATH	73A	74A	76A	77A	0 0000	0 0000	u uuuu
INTCON	73A	74A	76A	77A	0000 000x	0000 000u	uuuu uuuu (1)
PIR1	73A	74A	76A	77A	r000 0000	r000 0000	ruuu uuuu (1)
PIKI	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu (1)
PIR2	73A	74A	76A	77A	-0-0 00	-0-0 00	-u-u uu (1)
TMR1L	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
TMR1H	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
T1CON	73A	74A	76A	77A	00 0000	uu uuuu	uu uuuu
TMR2	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu
T2CON	73A	74A	76A	77A	-000 0000	-000 0000	-uuu uuuu
SSPBUF	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
SSPCON	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu
CCPR1L	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCPR1H	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCP1CON	73A	74A	76A	77A	00 0000	00 0000	uu uuuu
RCSTA	73A	74A	76A	77A	0000 000x	0000 000x	uuuu uuuu
TXREG	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu
RCREG	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu
CCPR2L	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCPR2H	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
CCP2CON	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu
ADRESH	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
ADCON0	73A	74A	76A	77A	0000 00-0	0000 00-0	uuuu uu-u
OPTION_REG	73A	74A	76A	77A	1111 1111	1111 1111	uuuu uuuu
TRISA	73A	74A	76A	77A	11 1111	11 1111	uu uuuu
TRISB	73A	74A	76A	77A	1111 1111	1111 1111	uuuu uuuu
TRISC	73A	74A	76A	77A	1111 1111	1111 1111	uuuu uuuu

TABLE 14-6: INITIALIZATION CONDITIONS FOR ALL REGISTERS

Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition, r = reserved, maintain clear. Shaded cells indicate conditions do not apply for the designated device.

Note 1: One or more bits in INTCON, PIR1 and/or PIR2 will be affected (to cause wake-up).

- 2: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).
- 3: See Table 14-5 for Reset value for specific condition.

TABLE 14-6: INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)							
Register		Devices		Power-on Reset, Brown-out Reset	MCLR Resets, WDT Reset	Wake-up via WDT or Interrupt	
TRISD	73A	74A	76A	77A	1111 1111	1111 1111	uuuu uuuu
TRISE	73A	74A	76A	77A	0000 -111	0000 -111	uuuu -uuu
PIE1	73A	74A	76A	77A	r000 0000	r000 0000	ruuu uuuu
	73A	74A	76A	77A	0000 0000	0000 0000	սսսս սսսս
PIE2	73A	74A	76A	77A	-0-0 00	-0-0 00	-u-u uu
PCON	73A	74A	76A	77A	dd	uu	uu
SSPCON2	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu
PR2	73A	74A	76A	77A	1111 1111	1111 1111	1111 1111
SSPADD	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu
SSPSTAT	73A	74A	76A	77A	00 0000	00 0000	uu uuuu
TXSTA	73A	74A	76A	77A	0000 -010	0000 -010	uuuu -uuu
SPBRG	73A	74A	76A	77A	0000 0000	0000 0000	uuuu uuuu
CMCON	73A	974	76A	77A	0000 0111	0000 0111	uuuu uuuu
CVRCON	73A	74A	76A	77A	000- 0000	000- 0000	uuu- uuuu
ADRESL	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
ADCON1	73A	74A	76A	77A	00 0000	00 0000	uu uuuu
EEDATA	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
EEADR	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	uuuu uuuu
EEDATH	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	นนนน นนนน
EEADRH	73A	74A	76A	77A	XXXX XXXX	uuuu uuuu	นนนน นนนน
EECON1	73A	74A	76A	77A	x x000	u u000	u uuuu
EECON2	73A	74A	76A	77A			

TABLE 14-6 :	INITIALIZATION CONDITIONS FOR ALL REGISTERS (CONTINUED)	

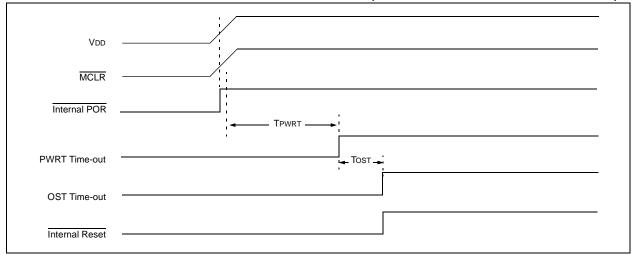
Legend: u = unchanged, x = unknown, - = unimplemented bit, read as '0', q = value depends on condition, <math>r = reserved, maintain clear. Shaded cells indicate conditions do not apply for the designated device.

Note 1: One or more bits in INTCON, PIR1 and/or PIR2 will be affected (to cause wake-up).

2: When the wake-up is due to an interrupt and the GIE bit is set, the PC is loaded with the interrupt vector (0004h).

3: See Table 14-5 for Reset value for specific condition.

FIGURE 14-6: TIME-OUT SEQUENCE ON POWER-UP (MCLR TIED TO VDD VIA RC NETWORK)





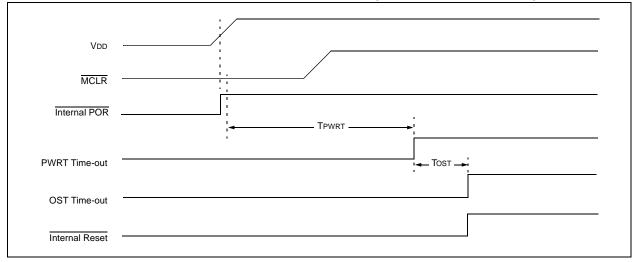


FIGURE 14-8: TIME-OUT SEQUENCE ON POWER-UP (MCLR NOT TIED TO VDD): CASE 2

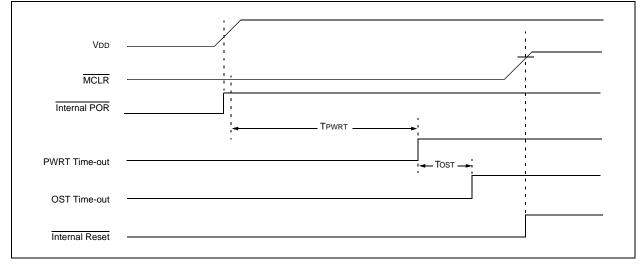
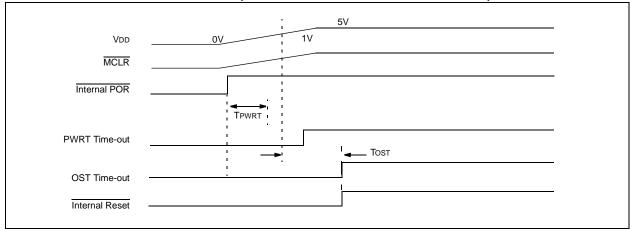


FIGURE 14-9: SLOW RISE TIME (MCLR TIED TO VDD VIA RC NETWORK)



14.11 Interrupts

The PIC16F87XA family has up to 15 sources of interrupt. The Interrupt Control register (INTCON) records individual interrupt requests in flag bits. It also has individual and global interrupt enable bits.

Note:	Individual interrupt flag bits are set regard-			
	less of the status of their corresponding			
	mask bit or the GIE bit.			

A global interrupt enable bit, GIE (INTCON<7>), enables (if set) all unmasked interrupts or disables (if cleared) all interrupts. When bit GIE is enabled and an interrupt's flag bit and mask bit are set, the interrupt will vector immediately. Individual interrupts can be disabled through their corresponding enable bits in various registers. Individual interrupt bits are set regardless of the status of the GIE bit. The GIE bit is cleared on Reset.

The "return from interrupt" instruction, RETFIE, exits the interrupt routine, as well as sets the GIE bit, which re-enables interrupts.

The RB0/INT pin interrupt, the RB port change interrupt and the TMR0 overflow interrupt flags are contained in the INTCON register.

The peripheral interrupt flags are contained in the Special Function Registers, PIR1 and PIR2. The corresponding interrupt enable bits are contained in Special Function Registers, PIE1 and PIE2, and the peripheral interrupt enable bit is contained in Special Function Register, INTCON.

When an interrupt is responded to, the GIE bit is cleared to disable any further interrupt, the return address is pushed onto the stack and the PC is loaded with 0004h. Once in the Interrupt Service Routine, the source(s) of the interrupt can be determined by polling the interrupt flag bits. The interrupt flag bit(s) must be cleared in software before re-enabling interrupts to avoid recursive interrupts.

For external interrupt events, such as the INT pin or PORTB change interrupt, the interrupt latency will be three or four instruction cycles. The exact latency depends when the interrupt event occurs. The latency is the same for one or two-cycle instructions. Individual interrupt flag bits are set regardless of the status of their corresponding mask bit, PEIE bit or GIE bit.

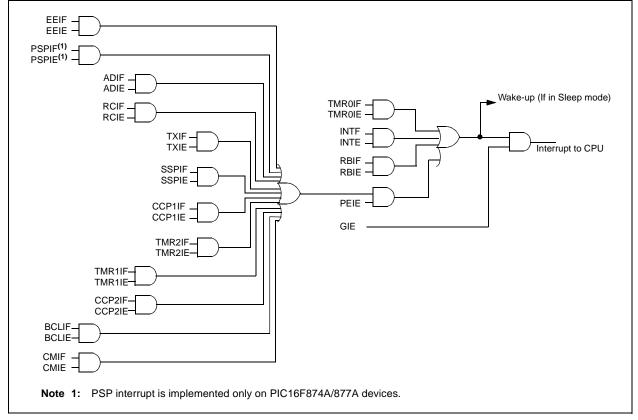


FIGURE 14-10: INTERRUPT LOGIC

14.11.1 INT INTERRUPT

External interrupt on the RB0/INT pin is edge triggered, either rising if bit INTEDG (OPTION_REG<6>) is set or falling if the INTEDG bit is clear. When a valid edge appears on the RB0/INT pin, flag bit, INTF (INTCON<1>), is set. This interrupt can be disabled by clearing enable bit, INTE (INTCON<4>). Flag bit INTF must be cleared in software in the Interrupt Service Routine before re-enabling this interrupt. The INT interrupt can wake-up the processor from Sleep if bit INTE was set prior to going into Sleep. The status of global interrupt enable bit, GIE, decides whether or not the processor branches to the interrupt vector following wake-up. See Section 14.14 "Power-down Mode (Sleep)" for details on Sleep mode.

14.11.2 TMR0 INTERRUPT

An overflow (FFh \rightarrow 00h) in the TMR0 register will set flag bit, TMR0IF (INTCON<2>). The interrupt can be enabled/disabled by setting/clearing enable bit, TMR0IE (INTCON<5>). See **Section 5.0 "Timer0 Module"**.

14.11.3 PORTB INTCON CHANGE

An input change on PORTB<7:4> sets flag bit, RBIF (INTCON<0>). The interrupt can be enabled/disabled by setting/clearing enable bit, RBIE (INTCON<4>). See Section 4.2 "PORTB and the TRISB Register".

14.12 Context Saving During Interrupts

During an interrupt, only the return PC value is saved on the stack. Typically, users may wish to save key registers during an interrupt (i.e., W register and Status register). This will have to be implemented in software.

For the PIC16F873A/874A devices, the register W_TEMP must be defined in both Banks 0 and 1 and must be defined at the same offset from the bank base address (i.e., If W_TEMP is defined at 0x20 in Bank 0, it must also be defined at 0xA0 in Bank 1). The registers, PCLATH_TEMP and STATUS_TEMP, are only defined in Bank 0.

Since the upper 16 bytes of each bank are common in the PIC16F876A/877A devices, temporary holding registers, W_TEMP, STATUS_TEMP and PCLATH_TEMP, should be placed in here. These 16 locations don't require banking and therefore, make it easier for context save and restore. The same code shown in Example 14-1 can be used.

EXAMPLE 14-1: SAVING STATUS, W AND PCLATH REGISTERS IN RAM

MOVWF	W_TEMP	;Copy W to TEMP register
SWAPF	STATUS,W	;Swap status to be saved into W
CLRF	STATUS	;bank 0, regardless of current bank, Clears IRP,RP1,RP0
MOVWF	STATUS_TEMP	;Save status to bank zero STATUS_TEMP register
MOVF	PCLATH, W	;Only required if using pages 1, 2 and/or 3
MOVWF	PCLATH_TEMP	;Save PCLATH into W
CLRF	PCLATH	;Page zero, regardless of current page
:		
:(ISR)		;(Insert user code here)
:		
MOVF	PCLATH_TEMP, W	;Restore PCLATH
MOVWF	PCLATH	;Move W into PCLATH
SWAPF	STATUS_TEMP,W	;Swap STATUS_TEMP register into W
		;(sets bank to original state)
MOVWF	STATUS	;Move W into STATUS register
SWAPF	W_TEMP,F	;Swap W_TEMP
SWAPF	W_TEMP,W	;Swap W_TEMP into W

14.13 Watchdog Timer (WDT)

The Watchdog Timer is a free running, on-chip RC oscillator which does not require any external components. This RC oscillator is separate from the RC oscillator of the OSC1/CLKI pin. That means that the WDT will run even if the clock on the OSC1/CLKI and OSC2/CLKO pins of the device has been stopped, for example, by execution of a SLEEP instruction.

During normal operation, a WDT time-out generates a device Reset (Watchdog Timer Reset). If the device is in Sleep mode, a WDT time-out causes the device to wake-up and continue with normal operation (Watchdog Timer Wake-up). The TO bit in the Status register will be cleared upon a Watchdog Timer time-out.

The WDT can be permanently disabled by clearing configuration bit, WDTE (Section 14.1 "Configuration Bits").

WDT time-out period values may be found in **Section 17.0** "**Electrical Characteristics**" under parameter #31. Values for the WDT prescaler (actually a postscaler but shared with the Timer0 prescaler) may be assigned using the OPTION_REG register.

- **Note 1:** The CLRWDT and SLEEP instructions clear the WDT and the postscaler, if assigned to the WDT and prevent it from timing out and generating a device Reset condition.
 - 2: When a CLRWDT instruction is executed and the prescaler is assigned to the WDT, the prescaler count will be cleared but the prescaler assignment is not changed.

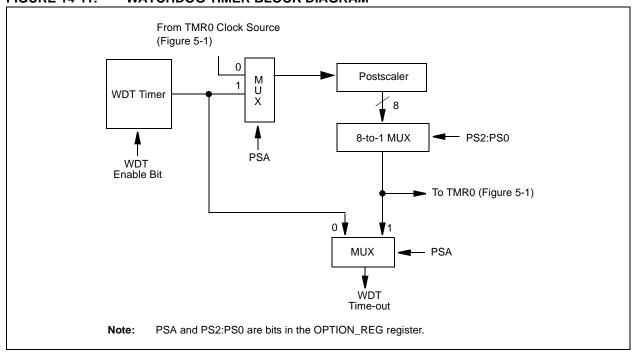


FIGURE 14-11: WATCHDOG TIMER BLOCK DIAGRAM

TABLE 14-7: SUMMARY OF WATCHDOG TIMER REGISTERS

Address	Name	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
2007h	Config. bits	(1)	BODEN ⁽¹⁾	CP1	CP0	PWRTE ⁽¹⁾	WDTE	Fosc1	Fosc0
81h, 181h	OPTION_REG	RBPU	INTEDG	TOCS	TOSE	PSA	PS2	PS1	PS0

Legend: Shaded cells are not used by the Watchdog Timer.

Note 1: See Register 14-1 for operation of these bits.

14.14 Power-down Mode (Sleep)

Power-down mode is entered by executing a SLEEP instruction.

If enabled, the Watchdog Timer will be cleared but keeps running, the PD bit (Status<3>) is cleared, the TO (Status<4>) bit is set and the oscillator driver is turned off. The I/O ports maintain the status they had before the SLEEP instruction was executed (driving high, low or high-impedance).

For lowest current consumption in this mode, place all I/O pins at either VDD or VSS, ensure no external circuitry is drawing current from the I/O pin, powerdown the A/D and disable external clocks. Pull all I/O pins that are high-impedance inputs, high or low externally, to avoid switching currents caused by floating inputs. The TOCKI input should also be at VDD or VSS for lowest current consumption. The contribution from on-chip pull-ups on PORTB should also be considered.

The $\overline{\text{MCLR}}$ pin must be at a logic high level (VIHMC).

14.14.1 WAKE-UP FROM SLEEP

The device can wake-up from Sleep through one of the following events:

- 1. External Reset input on MCLR pin.
- 2. Watchdog Timer wake-up (if WDT was enabled).
- 3. Interrupt from INT pin, RB port change or peripheral interrupt.

External MCLR Reset will cause a device Reset. All other events are considered a continuation of program execution and cause a "wake-up". The TO and PD bits in the Status register can be used to determine the cause of device Reset. The PD bit, which is <u>set</u> on power-up, is cleared when Sleep is invoked. The TO bit is cleared if a WDT time-out occurred and caused wake-up.

The following peripheral interrupts can wake the device from Sleep:

- 1. PSP read or write (PIC16F874/877 only).
- 2. TMR1 interrupt. Timer1 must be operating as an asynchronous counter.
- 3. CCP Capture mode interrupt.
- 4. Special event trigger (Timer1 in Asynchronous mode using an external clock).
- 5. SSP (Start/Stop) bit detect interrupt.
- 6. SSP transmit or receive in Slave mode (SPI/I²C).
- 7. USART RX or TX (Synchronous Slave mode).
- 8. A/D conversion (when A/D clock source is RC).
- 9. EEPROM write operation completion.
- 10. Comparator output changes state.

Other peripherals cannot generate interrupts since during Sleep, no on-chip clocks are present.

When the SLEEP instruction is being executed, the next instruction (PC + 1) is prefetched. For the device to wake-up through an interrupt event, the corresponding interrupt enable bit must be set (enabled). Wake-up is regardless of the state of the GIE bit. If the GIE bit is clear (disabled), the device continues execution at the instruction after the SLEEP instruction. If the GIE bit is set (enabled), the device executes the instruction after the SLEEP instruction and then branches to the interrupt address (0004h). In cases where the execution of the instruction following SLEEP is not desirable, the user should have a NOP after the SLEEP instruction.

14.14.2 WAKE-UP USING INTERRUPTS

When global interrupts are disabled (GIE cleared) and any interrupt source has both its interrupt enable bit and interrupt flag bit set, one of the following will occur:

- If the interrupt occurs **before** the execution of a SLEEP instruction, the SLEEP instruction will complete as a NOP. Therefore, the WDT and WDT postscaler will not be cleared, the TO bit will not be set and PD bits will not be cleared.
- If the interrupt occurs during or after the execution of a SLEEP instruction, the device will immediately wake-up from Sleep. The SLEEP instruction will be completely executed before the wake-up. Therefore, the WDT and WDT postscaler will be cleared, the TO bit will be set and the PD bit will be cleared.

Even if the flag bits were checked before executing a SLEEP instruction, it may be possible for flag bits to become set before the SLEEP instruction completes. To determine whether a SLEEP instruction executed, test the PD bit. If the PD bit is set, the SLEEP instruction was executed as a NOP.

To ensure that the WDT is cleared, a CLRWDT instruction should be executed before a SLEEP instruction.

FIGURE 14-12: WAKE-UP FROM SLEEP THROUGH INTERRUPT

OSC1	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1			Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4	Q1 Q2 Q3 Q4
CLKO ⁽⁴⁾			1	TOST(2)	,	; ;/		
INT pin		 		· · ·	 	1 1 1		, I I I I
INTF Flag (INTCON<1>	>)	i 	<u>````</u>		, , ,	Interrupt Latency	(2)	
GIE bit (INTCON<7>	>)	I I I	Processor in		I I I	<u>.</u>		. I I I
INSTRUCTIO	ON FLOW	' I I	Sleep		' I I		· · · · · · · · · · · · · · · · · · ·	· · · · · · · · · · · · · · · · · · ·
PC	с Х РС	X PC+1	<u>Х РС</u>	+2	X PC+2	X PC + 2	<u>χ 0004h</u>	X 0005h
Instruction Fetched	<pre>{ Inst(PC) = Sleep</pre>	Inst(PC + 1)	1 1 1	• •	Inst(PC + 2)	1 1 1	Inst(0004h)	Inst(0005h)
Instruction Executed	<pre>{ Inst(PC - 1)</pre>	Sleep	1 1 1		Inst(PC + 1)	Dummy cycle	Dummy cycle	Inst(0004h)

Note 1: XT, HS or LP Oscillator mode assumed.

2: TOST = 1024 TOSC (drawing not to scale). This delay will not be there for RC Oscillator mode.

3: GIE = 1 assumed. In this case, after wake- up, the processor jumps to the interrupt routine.

- If GIE = 0, execution will continue in-line.
- 4: CLKO is not available in these oscillator modes but shown here for timing reference.

14.15 In-Circuit Debugger

When the DEBUG bit in the configuration word is programmed to a '0', the in-circuit debugger functionality is enabled. This function allows simple debugging functions when used with MPLAB[®] ICD. When the microcontroller has this feature enabled, some of the resources are not available for general use. Table 14-8 shows which features are consumed by the background debugger.

TABLE 14-8: DEBUGGER RESOURCES	TABLE 14-8:	DEBUGGER RESOURCES
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I/O pins	RB6, RB7
Stack	1 level
Program Memory	Address 0000h must be NOP
	Last 100h words
Data Memory	0x070 (0x0F0, 0x170, 0x1F0) 0x1EB-0x1EF

To use the in-circuit debugger function of the microcontroller, the design must implement In-Circuit Serial Programming connections to MCLR/VPP, VDD, GND, RB7 and RB6. This will interface to the in-circuit debugger module available from Microchip or one of the third party development tool companies.

14.16 Program Verification/Code Protection

If the code protection bit(s) have not been programmed, the on-chip program memory can be read out for verification purposes.

14.17 ID Locations

Four memory locations (2000h-2003h) are designated as ID locations, where the user can store checksum or other code identification numbers. These locations are not accessible during normal execution but are readable and writable during program/verify. It is recommended that only the 4 Least Significant bits of the ID location are used.

14.18 In-Circuit Serial Programming

PIC16F87XA microcontrollers can be serially programmed while in the end application circuit. This is simply done with two lines for clock and data and three other lines for power, ground and the programming voltage. This allows customers to manufacture boards with unprogrammed devices and then program the microcontroller just before shipping the product. This also allows the most recent firmware or a custom firmware to be programmed.

When using ICSP, the part must be supplied at 4.5V to 5.5V if a bulk erase will be executed. This includes reprogramming of the code-protect, both from an on state to an off state. For all other cases of ICSP, the part may be programmed at the normal operating voltages. This means calibration values, unique user IDs or user code can be reprogrammed or added.

For complete details of serial programming, please refer to the "*PIC16F87XA Flash Memory Programming Specification*" (DS39589).

14.19 Low-Voltage (Single-Supply) ICSP Programming

The LVP bit of the configuration word enables lowvoltage ICSP programming. This mode allows the microcontroller to be programmed via ICSP using a VDD source in the operating voltage range. This only means that VPP does not have to be brought to VIHH but can instead be left at the normal operating voltage. In this mode, the RB3/PGM pin is dedicated to the programming function and ceases to be a general purpose I/O pin. During programming, VDD is applied to the MCLR pin. To enter Programming mode, VDD must be applied to the RB3/PGM provided the LVP bit is set. The LVP bit defaults to on ('1') from the factory.

- Note 1: The High-Voltage Programming mode is always available, regardless of the state of the LVP bit, by applying VIHH to the MCLR pin.
 - 2: While in Low-Voltage ICSP mode, the RB3 pin can no longer be used as a general purpose I/O pin.
 - 3: When using Low-Voltage ICSP Programming (LVP) and the pull-ups on PORTB are enabled, bit 3 in the TRISB register must be cleared to disable the pull-up on RB3 and ensure the proper operation of the device.
 - 4: RB3 should not be allowed to float if LVP is enabled. An external pull-down device should be used to default the device to normal operating mode. If RB3 floats high, the PIC16F87XA device will enter Programming mode.
 - LVP mode is enabled by default on all devices shipped from Microchip. It can be disabled by clearing the LVP bit in the CONFIG register.
 - 6: Disabling LVP will provide maximum compatibility to other PIC16CXXX devices.

If Low-Voltage Programming mode is not used, the LVP bit can be programmed to a '0' and RB3/PGM becomes a digital I/O pin. However, the LVP bit may only be programmed when programming is entered with VIHH on MCLR. The LVP bit can only be charged when using high voltage on MCLR.

It should be noted, that once the LVP bit is programmed to '0', only the High-Voltage Programming mode is available and only High-Voltage Programming mode can be used to program the device.

When using low-voltage ICSP, the part must be supplied at 4.5V to 5.5V if a bulk erase will be executed. This includes reprogramming of the code-protect bits from an on state to an off state. For all other cases of low-voltage ICSP, the part may be programmed at the normal operating voltage. This means calibration values, unique user IDs or user code can be reprogrammed or added.

15.0 INSTRUCTION SET SUMMARY

The PIC16 instruction set is highly orthogonal and is comprised of three basic categories:

- Byte-oriented operations
- Bit-oriented operations
- Literal and control operations

Each PIC16 instruction is a 14-bit word divided into an **opcode** which specifies the instruction type and one or more **operands** which further specify the operation of the instruction. The formats for each of the categories is presented in Figure 15-1, while the various opcode fields are summarized in Table 15-1.

Table 15-2 lists the instructions recognized by the MPASM[™] Assembler. A complete description of each instruction is also available in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

For **byte-oriented** instructions, 'f' represents a file register designator and 'd' represents a destination designator. The file register designator specifies which file register is to be used by the instruction.

The destination designator specifies where the result of the operation is to be placed. If 'd' is zero, the result is placed in the W register. If 'd' is one, the result is placed in the file register specified in the instruction.

For **bit-oriented** instructions, 'b' represents a bit field designator which selects the bit affected by the operation, while 'f' represents the address of the file in which the bit is located.

For **literal and control** operations, 'k' represents an eight or eleven-bit constant or literal value

One instruction cycle consists of four oscillator periods; for an oscillator frequency of 4 MHz, this gives a normal instruction execution time of 1 μ s. All instructions are executed within a single instruction cycle, unless a conditional test is true, or the program counter is changed as a result of an instruction. When this occurs, the execution takes two instruction cycles with the second cycle executed as a NOP.

Note:	To maintain upward compatibility with
	future PIC16F87XA products, do not use
	the OPTION and TRIS instructions.

All instruction examples use the format '0xhh' to represent a hexadecimal number, where 'h' signifies a hexadecimal digit.

15.1 READ-MODIFY-WRITE OPERATIONS

Any instruction that specifies a file register as part of the instruction performs a Read-Modify-Write (R-M-W) operation. The register is read, the data is modified, and the result is stored according to either the instruction or the destination designator 'd'. A read operation is performed on a register even if the instruction writes to that register. For example, a "CLRF PORTB" instruction will read PORTB, clear all the data bits, then write the result back to PORTB. This example would have the unintended result that the condition that sets the RBIF flag would be cleared.

TABLE 15-1: OPCODE FIELD DESCRIPTIONS

Field	Description
f	Register file address (0x00 to 0x7F)
W	Working register (accumulator)
b	Bit address within an 8-bit file register
k	Literal field, constant data or label
x	Don't care location (= 0 or 1). The assembler will generate code with x = 0 . It is the recommended form of use for compatibility with all Microchip software tools.
d	Destination select; $d = 0$: store result in W, d = 1: store result in file register f. Default is $d = 1$.
PC	Program Counter
ТО	Time-out bit
PD	Power-down bit

FIGURE 15-1: GENERAL FORMAT FOR INSTRUCTIONS

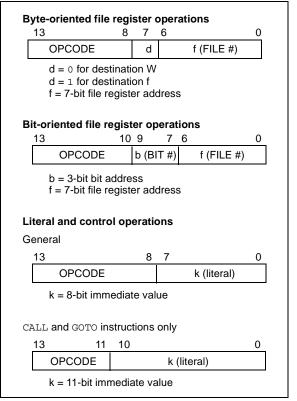


TABLE 15-2:	PIC16F87XA INSTRUCTION SET
-------------	----------------------------

Mnemonic, Operands		Description	Cycles	14-Bit Opcode				Status	Natas
		Description		MSb			LSb	Affected	Notes
		BYTE-ORIENTED FILE	REGISTER OPE	RATIC	NS				
ADDWF	f, d	Add W and f	1	00	0111	dfff	ffff	C,DC,Z	1,2
ANDWF	f, d	AND W with f	1	00	0101	dfff	ffff	Z	1,2
CLRF	f	Clear f	1	00	0001	lfff	ffff	Z	2
CLRW	-	Clear W	1	00	0001	0xxx	xxxx	Z	
COMF	f, d	Complement f	1	00	1001	dfff	ffff	Z	1,2
DECF	f, d	Decrement f	1	00	0011	dfff	ffff	Z	1,2
DECFSZ	f, d	Decrement f, Skip if 0	1(2)	00	1011	dfff	ffff		1,2,3
INCF	f, d	Increment f	1	00	1010	dfff	ffff	Z	1,2
INCFSZ	f, d	Increment f, Skip if 0	1(2)	00	1111	dfff	ffff		1,2,3
IORWF	f, d	Inclusive OR W with f	1	00	0100	dfff	ffff	Z	1,2
MOVF	f, d	Move f	1	00	1000	dfff	ffff	Z	1,2
MOVWF	f	Move W to f	1	00	0000	lfff	ffff		
NOP	-	No Operation	1	00	0000	0xx0	0000		
RLF	f, d	Rotate Left f through Carry	1	00	1101	dfff	ffff	С	1,2
RRF	f, d	Rotate Right f through Carry	1	00	1100	dfff	ffff	С	1,2
SUBWF	f, d	Subtract W from f	1	00	0010	dfff	ffff	C,DC,Z	1,2
SWAPF	f, d	Swap nibbles in f	1	00	1110	dfff	ffff		1,2
XORWF	f, d	Exclusive OR W with f	1	00	0110	dfff	ffff	Z	1,2
		BIT-ORIENTED FILE		ATIO	NS				
BCF	f, b	Bit Clear f	1	01	00bb	bfff	ffff		1,2
BSF	f, b	Bit Set f	1	01	01bb	bfff	ffff		1,2
BTFSC	f, b	Bit Test f, Skip if Clear	1 (2)	01	10bb	bfff	ffff		3
BTFSS	f, b	Bit Test f, Skip if Set	1 (2)	01	11bb	bfff	ffff		3
		LITERAL AND CO	NTROL OPERAT	IONS					
ADDLW	k	Add Literal and W	1	11	111x	kkkk	kkkk	C,DC,Z	
ANDLW	k	AND Literal with W	1	11	1001	kkkk	kkkk	Z	
CALL	k	Call Subroutine	2	10	0kkk	kkkk	kkkk		
CLRWDT	-	Clear Watchdog Timer	1	00	0000	0110	0100	TO,PD	
GOTO	k	Go to Address	2	10	1kkk	kkkk	kkkk		
IORLW	k	Inclusive OR Literal with W	1	11	1000	kkkk	kkkk	Z	
MOVLW	k	Move Literal to W	1	11	00xx	kkkk	kkkk		
RETFIE	-	Return from Interrupt	2	00	0000	0000	1001		
RETLW	k	Return with Literal in W	2	11	01xx	kkkk	kkkk		
RETURN	-	Return from Subroutine	2	00	0000	0000	1000		
SLEEP	-	Go into Standby mode	1	00	0000	0110	0011	TO,PD	
SUBLW	k	Subtract W from Literal	1	11		kkkk		C,DC,Z	
XORLW	k	Exclusive OR Literal with W	1	11	1010		kkkk	Z	
Note 1:		I/O register is modified as a function of itse		L					L

Note 1: When an I/O register is modified as a function of itself (e.g., MOVF PORTE, 1), the value used will be that value present on the pins themselves. For example, if the data latch is '1' for a pin configured as input and is driven low by an external device, the data will be written back with a '0'.

2: If this instruction is executed on the TMR0 register (and where applicable, d = 1), the prescaler will be cleared if assigned to the Timer0 module.

3: If Program Counter (PC) is modified, or a conditional test is true, the instruction requires two cycles. The second cycle is executed as a NOP.

Note: Additional information on the mid-range instruction set is available in the PICmicro[®] Mid-Range MCU Family Reference Manual (DS33023).

15.2 Instruction Descriptions

ADDLW	Add Literal and W
Syntax:	[<i>label</i>] ADDLW k
Operands:	$0 \le k \le 255$
Operation:	$(W) + k \to (W)$
Status Affected:	C, DC, Z
Description:	The contents of the W register are added to the eight-bit literal 'k' and the result is placed in the W register.

ADDWF	Add W and f
Syntax:	[<i>label</i>] ADDWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \left[0,1 \right] \end{array}$
Operation:	(W) + (f) \rightarrow (destination)
Status Affected:	C, DC, Z
Description:	Add the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

BCF	Bit Clear f
Syntax:	[label] BCF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$0 \rightarrow (f < b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is cleared.

BSF	Bit Set f
Syntax:	[label] BSF f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	$1 \rightarrow (f < b >)$
Status Affected:	None
Description:	Bit 'b' in register 'f' is set.

ANDLW	AND Literal with W
Syntax:	[<i>label</i>] ANDLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .AND. (k) \rightarrow (W)
Status Affected:	Z
Description:	The contents of W register are AND'ed with the eight-bit literal 'k'. The result is placed in the W register.

BTFSS	Bit Test f, Skip if Set
Syntax:	[label] BTFSS f,b
Operands:	$0 \le f \le 127$ $0 \le b < 7$
Operation:	skip if (f) = 1
Status Affected:	None
Description:	If bit 'b' in register 'f' is '0', the next instruction is executed. If bit 'b' is '1', then the next instruction is discarded and a NOP is executed instead, making this a 2 TCY instruction.

ANDWF	AND W with f
Syntax:	[label] ANDWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) .AND. (f) \rightarrow (destination)
Status Affected:	Z
Description:	AND the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

BTFSC	Bit Test, Skip if Clear
Syntax:	[label] BTFSC f,b
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ 0 \leq b \leq 7 \end{array}$
Operation:	skip if (f) = 0
Status Affected:	None
Description:	If bit 'b' in register 'f' is '1', the next instruction is executed. If bit 'b' in register 'f' is '0', the next instruction is discarded and a NOP is executed instead, making this a 2 TCY instruction.

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CALL	Call Subroutine
Syntax:	[<i>label</i>] CALL k
Operands:	$0 \le k \le 2047$
Operation:	$\begin{array}{l} (PC)+1 \rightarrow TOS, \\ k \rightarrow PC < 10:0>, \\ (PCLATH < 4:3>) \rightarrow PC < 12:11> \end{array}$
Status Affected:	None
Description:	Call Subroutine. First, return address (PC+1) is pushed onto the stack. The eleven-bit immediate address is loaded into PC bits <10:0>. The upper bits of the PC are loaded from PCLATH. CALL is a two-cycle instruction.

CLRWDT	Clear Watchdog Timer
Syntax:	[label] CLRWDT
Operands:	None
Operation: Status Affected:	$\begin{array}{l} 00h \rightarrow WDT \\ 0 \rightarrow WDT \ prescaler, \\ 1 \rightarrow \overline{TO} \\ 1 \rightarrow PD \\ \overline{TO}, \ \overline{PD} \end{array}$
Description:	CLRWDT instruction resets the Watchdog Timer. It also resets the prescaler of the WDT. Status bits, TO and PD, are set.

CLRF	Clear f	
Syntax:	[label] CLRF f	
Operands:	$0 \le f \le 127$	
Operation:	$\begin{array}{l} 00h \rightarrow (f) \\ 1 \rightarrow Z \end{array}$	
Status Affected:	Z	
Description:	The contents of register 'f' are cleared and the Z bit is set.	

COMF	Complement f
Syntax:	[label] COMF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \left[0,1\right] \end{array}$
Operation:	$(\overline{f}) \rightarrow (destination)$
Status Affected:	Z
Description:	The contents of register 'f' are complemented. If 'd' is '0', the result is stored in W. If 'd' is '1', the result is stored back in register 'f'.

CLRW	Clear W	DECF	Decrement f
Syntax:	[label] CLRW	Syntax:	[label] DECF f,d
Operands:		Operands:	0 ≤ f ≤ 127 d ∈ [0,1]
Operation:	$\begin{array}{l} 00h \rightarrow (W) \\ 1 \rightarrow Z \end{array}$	Operation:	(f) - 1 \rightarrow (destination)
Status Affected:	Z	Status Affected:	Z
Description:	W register is cleared. Zero bit (Z) is set.	Description:	Decrement register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

DECFSZ	Decrement f, Skip if 0
Syntax:	[label] DECFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$
Operation:	(f) - 1 \rightarrow (destination); skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are decremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruction is executed. If the result is '0', then a NOP is executed instead, making it a 2 TCY instruction.

INCFSZ	Increment f, Skip if 0
Syntax:	[label] INCFSZ f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$
Operation:	(f) + 1 \rightarrow (destination), skip if result = 0
Status Affected:	None
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'. If the result is '1', the next instruc- tion is executed. If the result is '0', a NOP is executed instead, making it a 2 TCY instruction.

GOTO	Unconditional Branch
Syntax:	[<i>label</i>] GOTO k
Operands:	$0 \le k \le 2047$
Operation:	$k \rightarrow PC<10:0>$ PCLATH<4:3> \rightarrow PC<12:11>
Status Affected:	None
Description:	GOTO is an unconditional branch. The eleven-bit immediate value is loaded into PC bits <10:0>. The upper bits of PC are loaded from PCLATH<4:3>. GOTO is a two-cycle instruction.

IORLW	Inclusive OR Literal with W
Syntax:	[label] IORLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .OR. $k \rightarrow$ (W)
Status Affected:	Z
Description:	The contents of the W register are OR'ed with the eight-bit literal 'k'. The result is placed in the W register.

INCF	Increment f	IORWF	Inclusive OR W with f
Syntax:	[label] INCF f,d	Syntax:	[label] IORWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$	Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$
Operation:	(f) + 1 \rightarrow (destination)	Operation:	(W) .OR. (f) \rightarrow (destination)
Status Affected:	Z	Status Affected:	Z
Description:	The contents of register 'f' are incremented. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.	Description:	Inclusive OR the W register with register 'f'. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

RLF	Rotate Left f through Carry
Syntax:	[<i>label</i>] RLF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$
Operation:	See description below
Status Affected:	С
Description:	The contents of register 'f' are rotated one bit to the left through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is stored back in register 'f'.

SLEEP

Syntax:	[label] SLEEP
Operands:	None
Operation:	$\begin{array}{l} 00h \rightarrow WDT, \\ 0 \rightarrow \underline{W}DT \text{ prescaler}, \\ 1 \rightarrow \underline{TO}, \\ 0 \rightarrow \overline{PD} \end{array}$
Status Affected:	TO, PD
Description:	The power-down status bit, $\overline{\text{PD}}$, is cleared. Time-out status bit, $\overline{\text{TO}}$, is set. Watchdog Timer and its prescaler are cleared. The processor is put into Sleep mode with the oscillator stopped.

RETURN	Return from Subroutine
Syntax:	[label] RETURN
Operands:	None
Operation:	$TOS\toPC$
Status Affected:	None
Description:	Return from subroutine. The stack is POPed and the top of the stack (TOS) is loaded into the program counter. This is a two-cycle instruction.

SUBLW	Subtract W from Literal
Syntax:	[<i>label</i>] SUBLW k
Operands:	$0 \le k \le 255$
Operation:	$k \text{ - (W)} \to (W)$
Status Affected:	C, DC, Z
Description:	The W register is subtracted (2's complement method) from the eight-bit literal 'k'. The result is placed in the W register.

RRF	Rotate Right f through Carry
Syntax:	[label] RRF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$
Operation:	See description below
Status Affected:	С
Description:	The contents of register 'f' are rotated one bit to the right through the Carry flag. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed back in register 'f'.

SUBWF	Subtract W from f
Syntax:	[<i>label</i>] SUBWF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$
Operation:	(f) - (W) \rightarrow (destination)
Status Affected:	C, DC, Z
Description:	Subtract (2's complement method) W register from register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.



SWAPF	Swap Nibbles in f
Syntax:	[label] SWAPF f,d
Operands:	$\begin{array}{l} 0 \leq f \leq 127 \\ d \in \ [0,1] \end{array}$
Operation:	$(f<3:0>) \rightarrow (destination<7:4>),$ $(f<7:4>) \rightarrow (destination<3:0>)$
Status Affected:	None
Description:	The upper and lower nibbles of register 'f' are exchanged. If 'd' is '0', the result is placed in the W register. If 'd' is '1', the result is placed in register 'f'.

XORWF	Exclusive OR W with f
Syntax:	[label] XORWF f,d
Operands:	$0 \le f \le 127$ $d \in [0,1]$
Operation:	(W) .XOR. (f) \rightarrow (destination)
Status Affected:	Z
Description:	Exclusive OR the contents of the W register with register 'f'. If 'd' is '0', the result is stored in the W register. If 'd' is '1', the result is stored back in register 'f'.

XORLW	Exclusive OR Literal with W
Syntax:	[label] XORLW k
Operands:	$0 \le k \le 255$
Operation:	(W) .XOR. $k \rightarrow (W)$
Status Affected:	Z
Description:	The contents of the W register are XOR'ed with the eight-bit literal 'k'. The result is placed in the W register.

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NOTES:

16.0 DEVELOPMENT SUPPORT

The PICmicro $^{\mbox{\tiny B}}$ microcontrollers are supported with a full range of hardware and software development tools:

- Integrated Development Environment
 - MPLAB® IDE Software
- Assemblers/Compilers/Linkers
 - MPASM[™] Assembler
 - MPLAB C17 and MPLAB C18 C Compilers
 - MPLINK[™] Object Linker/ MPLIB[™] Object Librarian
 - MPLAB C30 C Compiler
 - MPLAB ASM30 Assembler/Linker/Library
- Simulators
 - MPLAB SIM Software Simulator
 - MPLAB dsPIC30 Software Simulator
- Emulators
 - MPLAB ICE 2000 In-Circuit Emulator
 - MPLAB ICE 4000 In-Circuit Emulator
- In-Circuit Debugger
- MPLAB ICD 2
- Device Programmers
 - PRO MATE[®] II Universal Device Programmer
 - PICSTART[®] Plus Development Programmer
- Low Cost Demonstration Boards
 - PICDEM[™] 1 Demonstration Board
 - PICDEM.net[™] Demonstration Board
 - PICDEM 2 Plus Demonstration Board
 - PICDEM 3 Demonstration Board
 - PICDEM 4 Demonstration Board
 - PICDEM 17 Demonstration Board
 - PICDEM 18R Demonstration Board
 - PICDEM LIN Demonstration Board
 - PICDEM USB Demonstration Board
- Evaluation Kits
 - KEELOQ[®]
 - PICDEM MSC
 - microID®
 - CAN
 - PowerSmart®
 - Analog

16.1 MPLAB Integrated Development Environment Software

The MPLAB IDE software brings an ease of software development previously unseen in the 8/16-bit microcontroller market. The MPLAB IDE is a Windows[®] based application that contains:

- An interface to debugging tools
 - simulator
 - programmer (sold separately)
 - emulator (sold separately)
 - in-circuit debugger (sold separately)
- · A full-featured editor with color coded context
- A multiple project manager
- Customizable data windows with direct edit of contents
- High level source code debugging
- Mouse over variable inspection
- Extensive on-line help

The MPLAB IDE allows you to:

- Edit your source files (either assembly or C)
- One touch assemble (or compile) and download to PICmicro emulator and simulator tools (automatically updates all project information)
- Debug using:
 - source files (assembly or C)
 - absolute listing file (mixed assembly and C)
 - machine code

MPLAB IDE supports multiple debugging tools in a single development paradigm, from the cost effective simulators, through low cost in-circuit debuggers, to full-featured emulators. This eliminates the learning curve when upgrading to tools with increasing flexibility and power.

16.2 MPASM Assembler

The MPASM assembler is a full-featured, universal macro assembler for all PICmicro MCUs.

The MPASM assembler generates relocatable object files for the MPLINK object linker, Intel[®] standard HEX files, MAP files to detail memory usage and symbol reference, absolute LST files that contain source lines and generated machine code and COFF files for debugging.

The MPASM assembler features include:

- Integration into MPLAB IDE projects
- User defined macros to streamline assembly code
- Conditional assembly for multi-purpose source files
- Directives that allow complete control over the assembly process

16.3 MPLAB C17 and MPLAB C18 C Compilers

The MPLAB C17 and MPLAB C18 Code Development Systems are complete ANSI C compilers for Microchip's PIC17CXXX and PIC18CXXX family of microcontrollers. These compilers provide powerful integration capabilities, superior code optimization and ease of use not found with other compilers.

For easy source level debugging, the compilers provide symbol information that is optimized to the MPLAB IDE debugger.

16.4 MPLINK Object Linker/ MPLIB Object Librarian

The MPLINK object linker combines relocatable objects created by the MPASM assembler and the MPLAB C17 and MPLAB C18 C compilers. It can link relocatable objects from precompiled libraries, using directives from a linker script.

The MPLIB object librarian manages the creation and modification of library files of pre-compiled code. When a routine from a library is called from a source file, only the modules that contain that routine will be linked in with the application. This allows large libraries to be used efficiently in many different applications.

The object linker/library features include:

- Efficient linking of single libraries instead of many smaller files
- Enhanced code maintainability by grouping related modules together
- Flexible creation of libraries with easy module listing, replacement, deletion and extraction

16.5 MPLAB C30 C Compiler

The MPLAB C30 C compiler is a full-featured, ANSI compliant, optimizing compiler that translates standard ANSI C programs into dsPIC30F assembly language source. The compiler also supports many command-line options and language extensions to take full advantage of the dsPIC30F device hardware capabilities, and afford fine control of the compiler code generator.

MPLAB C30 is distributed with a complete ANSI C standard library. All library functions have been validated and conform to the ANSI C library standard. The library includes functions for string manipulation, dynamic memory allocation, data conversion, timekeeping, and math functions (trigonometric, exponential and hyperbolic). The compiler provides symbolic information for high level source debugging with the MPLAB IDE.

16.6 MPLAB ASM30 Assembler, Linker, and Librarian

MPLAB ASM30 assembler produces relocatable machine code from symbolic assembly language for dsPIC30F devices. MPLAB C30 compiler uses the assembler to produce it's object file. The assembler generates relocatable object files that can then be archived or linked with other relocatable object files and archives to create an executable file. Notable features of the assembler include:

- Support for the entire dsPIC30F instruction set
- Support for fixed-point and floating-point data
- Command line interface
- Rich directive set
- Flexible macro language
- MPLAB IDE compatibility

16.7 MPLAB SIM Software Simulator

The MPLAB SIM software simulator allows code development in a PC hosted environment by simulating the PICmicro series microcontrollers on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a file, or user defined key press, to any pin. The execution can be performed in Single-Step, Execute Until Break, or Trace mode.

The MPLAB SIM simulator fully supports symbolic debugging using the MPLAB C17 and MPLAB C18 C Compilers, as well as the MPASM assembler. The software simulator offers the flexibility to develop and debug code outside of the laboratory environment, making it an excellent, economical software development tool.

16.8 MPLAB SIM30 Software Simulator

The MPLAB SIM30 software simulator allows code development in a PC hosted environment by simulating the dsPIC30F series microcontrollers on an instruction level. On any given instruction, the data areas can be examined or modified and stimuli can be applied from a file, or user defined key press, to any of the pins.

The MPLAB SIM30 simulator fully supports symbolic debugging using the MPLAB C30 C Compiler and MPLAB ASM30 assembler. The simulator runs in either a Command Line mode for automated tasks, or from MPLAB IDE. This high speed simulator is designed to debug, analyze and optimize time intensive DSP routines.

16.9 MPLAB ICE 2000 High Performance Universal In-Circuit Emulator

The MPLAB ICE 2000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for PICmicro microcontrollers. Software control of the MPLAB ICE 2000 in-circuit emulator is advanced by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICE 2000 is a full-featured emulator system with enhanced trace, trigger and data monitoring features. Interchangeable processor modules allow the system to be easily reconfigured for emulation of different processors. The universal architecture of the MPLAB ICE in-circuit emulator allows expansion to support new PICmicro microcontrollers.

The MPLAB ICE 2000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft[®] Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

16.10 MPLAB ICE 4000 High Performance Universal In-Circuit Emulator

The MPLAB ICE 4000 universal in-circuit emulator is intended to provide the product development engineer with a complete microcontroller design tool set for highend PICmicro microcontrollers. Software control of the MPLAB ICE in-circuit emulator is provided by the MPLAB Integrated Development Environment, which allows editing, building, downloading and source debugging from a single environment.

The MPLAB ICD 4000 is a premium emulator system, providing the features of MPLAB ICE 2000, but with increased emulation memory and high speed performance for dsPIC30F and PIC18XXXX devices. Its advanced emulator features include complex triggering and timing, up to 2 Mb of emulation memory, and the ability to view variables in real-time.

The MPLAB ICE 4000 in-circuit emulator system has been designed as a real-time emulation system with advanced features that are typically found on more expensive development tools. The PC platform and Microsoft Windows 32-bit operating system were chosen to best make these features available in a simple, unified application.

16.11 MPLAB ICD 2 In-Circuit Debugger

Microchip's In-Circuit Debugger, MPLAB ICD 2, is a powerful, low-cost, run-time development tool, connecting to the host PC via an RS-232 or high-speed USB interface. This tool is based on the Flash PICmicro MCUs and can be used to develop for these and other PICmicro microcontrollers. The MPLAB ICD 2 utilizes the in-circuit debugging capability built into the Flash devices. This feature, along with Microchip's In-Circuit Serial Programming[™] (ICSP[™]) protocol, offers cost effective in-circuit Flash debugging from the graphical user interface of the MPLAB Integrated Development Environment. This enables a designer to develop and debug source code by setting breakpoints, single-stepping and watching variables, CPU status and peripheral registers. Running at full speed enables testing hardware and applications in real-time. MPLAB ICD 2 also serves as a development programmer for selected PICmicro devices.

16.12 PRO MATE II Universal Device Programmer

The PRO MATE II is a universal, CE compliant device programmer with programmable voltage verification at VDDMIN and VDDMAX for maximum reliability. It features an LCD display for instructions and error messages and a modular detachable socket assembly to support various package types. In Stand-Alone mode, the PRO MATE II device programmer can read, verify, and program PICmicro devices without a PC connection. It can also set code protection in this mode.

16.13 PICSTART Plus Development Programmer

The PICSTART Plus development programmer is an easy-to-use, low-cost, prototype programmer. It connects to the PC via a COM (RS-232) port. MPLAB Integrated Development Environment software makes using the programmer simple and efficient. The PICSTART Plus development programmer supports most PICmicro devices up to 40 pins. Larger pin count devices, such as the PIC16C92X and PIC17C76X, may be supported with an adapter socket. The PICSTART Plus development programmer is CE compliant.

16.14 PICDEM 1 PICmicro Demonstration Board

The PICDEM 1 demonstration board demonstrates the capabilities of the PIC16C5X (PIC16C54 to PIC16C58A), PIC16C61, PIC16C62X, PIC16C71, PIC16C8X, PIC17C42, PIC17C43 and PIC17C44. All necessary hardware and software is included to run basic demo programs. The sample microcontrollers provided with the PICDEM 1 demonstration board can be programmed with a PRO MATE II device programmer, or a PICSTART Plus development programmer. The PICDEM 1 demonstration board can be connected to the MPLAB ICE in-circuit emulator for testing. A prototype area extends the circuitry for additional application components. Features include an RS-232 interface, a potentiometer for simulated analog input, push button switches and eight LEDs.

16.15 PICDEM.net Internet/Ethernet Demonstration Board

The PICDEM.net demonstration board is an Internet/ Ethernet demonstration board using the PIC18F452 microcontroller and TCP/IP firmware. The board supports any 40-pin DIP device that conforms to the standard pinout used by the PIC16F877 or PIC18C452. This kit features a user friendly TCP/IP stack, web server with HTML, a 24L256 Serial EEPROM for Xmodem download to web pages into Serial EEPROM, ICSP/MPLAB ICD 2 interface connector, an Ethernet interface, RS-232 interface, and a 16 x 2 LCD display. Also included is the book and CD-ROM *"TCP/IP Lean, Web Servers for Embedded Systems,"* by Jeremy Bentham.

16.16 PICDEM 2 Plus Demonstration Board

The PICDEM 2 Plus demonstration board supports many 18-, 28-, and 40-pin microcontrollers, including PIC16F87X and PIC18FXX2 devices. All the necessary hardware and software is included to run the demonstration programs. The sample microcontrollers provided with the PICDEM 2 demonstration board can be programmed with a PRO MATE II device programmer, PICSTART Plus development programmer, or MPLAB ICD 2 with a Universal Programmer Adapter. The MPLAB ICD 2 and MPLAB ICE in-circuit emulators may also be used with the PICDEM 2 demonstration board to test firmware. A prototype area extends the circuitry for additional application components. Some of the features include an RS-232 interface, a 2 x 16 LCD display, a piezo speaker, an on-board temperature sensor, four LEDs, and sample PIC18F452 and PIC16F877 Flash microcontrollers.

16.17 PICDEM 3 PIC16C92X Demonstration Board

The PICDEM 3 demonstration board supports the PIC16C923 and PIC16C924 in the PLCC package. All the necessary hardware and software is included to run the demonstration programs.

16.18 PICDEM 4 8/14/18-Pin Demonstration Board

The PICDEM 4 can be used to demonstrate the capabilities of the 8, 14, and 18-pin PIC16XXXX and PIC18XXXX MCUs, including the PIC16F818/819, PIC16F87/88, PIC16F62XA and the PIC18F1320 family of microcontrollers. PICDEM 4 is intended to showcase the many features of these low pin count parts, including LIN and Motor Control using ECCP. Special provisions are made for low power operation with the supercapacitor circuit, and jumpers allow on-board hardware to be disabled to eliminate current draw in this mode. Included on the demo board are provisions for Crystal, RC or Canned Oscillator modes, a five volt regulator for use with a nine volt wall adapter or battery, DB-9 RS-232 interface, ICD connector for programming via ICSP and development with MPLAB ICD 2, 2x16 liquid crystal display, PCB footprints for H-Bridge motor driver, LIN transceiver and EEPROM. Also included are: header for expansion, eight LEDs, four potentiometers, three push buttons and a prototyping area. Included with the kit is a PIC16F627A and a PIC18F1320. Tutorial firmware is included along with the User's Guide.

16.19 PICDEM 17 Demonstration Board

The PICDEM 17 demonstration board is an evaluation board that demonstrates the capabilities of several Microchip microcontrollers, including PIC17C752, PIC17C756A, PIC17C762 and PIC17C766. A programmed sample is included. The PRO MATE II device programmer, or the PICSTART Plus development programmer, can be used to reprogram the device for user tailored application development. The PICDEM 17 demonstration board supports program download and execution from external on-board Flash memory. A generous prototype area is available for user hardware expansion.

16.20 PICDEM 18R PIC18C601/801 Demonstration Board

The PICDEM 18R demonstration board serves to assist development of the PIC18C601/801 family of Microchip microcontrollers. It provides hardware implementation of both 8-bit Multiplexed/Demultiplexed and 16-bit Memory modes. The board includes 2 Mb external Flash memory and 128 Kb SRAM memory, as well as serial EEPROM, allowing access to the wide range of memory types supported by the PIC18C601/801.

16.21 PICDEM LIN PIC16C43X Demonstration Board

The powerful LIN hardware and software kit includes a series of boards and three PICmicro microcontrollers. The small footprint PIC16C432 and PIC16C433 are used as slaves in the LIN communication and feature on-board LIN transceivers. A PIC16F874 Flash micro-controller serves as the master. All three microcontrollers are programmed with firmware to provide LIN bus communication.

16.22 PICkit[™] 1 Flash Starter Kit

A complete "development system in a box", the PICkit Flash Starter Kit includes a convenient multi-section board for programming, evaluation and development of 8/14-pin Flash PIC[®] microcontrollers. Powered via USB, the board operates under a simple Windows GUI. The PICkit 1 Starter Kit includes the user's guide (on CD ROM), PICkit 1 tutorial software and code for various applications. Also included are MPLAB[®] IDE (Integrated Development Environment) software, software and hardware "Tips 'n Tricks for 8-pin Flash PIC[®] Microcontrollers" Handbook and a USB Interface Cable. Supports all current 8/14-pin Flash PIC microcontrollers, as well as many future planned devices.

16.23 PICDEM USB PIC16C7X5 Demonstration Board

The PICDEM USB Demonstration Board shows off the capabilities of the PIC16C745 and PIC16C765 USB microcontrollers. This board provides the basis for future USB products.

16.24 Evaluation and Programming Tools

In addition to the PICDEM series of circuits, Microchip has a line of evaluation kits and demonstration software for these products.

- KEELOQ evaluation and programming tools for Microchip's HCS Secure Data Products
- CAN developers kit for automotive network applications
- Analog design boards and filter design software
- PowerSmart battery charging evaluation/ calibration kits
- IrDA[®] development kit
- microID development and rfLab[™] development software
- SEEVAL[®] designer kit for memory evaluation and endurance calculations
- PICDEM MSC demo boards for Switching mode power supply, high power IR driver, delta sigma ADC, and flow rate sensor

Check the Microchip web page and the latest Product Line Card for the complete list of demonstration and evaluation kits.

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NOTES:

17.0 ELECTRICAL CHARACTERISTICS

Absolute Maximum Ratings †

Ambient temperature under bias	55 to +125°C
Storage temperature	
Voltage on any pin with respect to Vss (except VDD, MCLR. and RA4)	0.3V to (VDD + 0.3V)
Voltage on VDD with respect to Vss	0.3 to +7.5V
Voltage on MCLR with respect to Vss (Note 2)	0 to +14V
Voltage on RA4 with respect to Vss	0 to +8.5V
Total power dissipation (Note 1)	1.0W
Maximum current out of Vss pin	300 mA
Maximum current into Vod pin	250 mA
Input clamp current, Iк (Vi < 0 or Vi > VDD)	±20 mA
Output clamp current, loк (Vo < 0 or Vo > VDD)	±20 mA
Maximum output current sunk by any I/O pin	25 mA
Maximum output current sourced by any I/O pin	25 mA
Maximum current sunk by PORTA, PORTB and PORTE (combined) (Note 3)	200 mA
Maximum current sourced by PORTA, PORTB and PORTE (combined) (Note 3)	200 mA
Maximum current sunk by PORTC and PORTD (combined) (Note 3)	200 mA
Maximum current sourced by PORTC and PORTD (combined) (Note 3)	200 mA
Note 1: Power dissipation is calculated as follows: Pdis = VDD x {IDD - \sum IOH} + \sum {(VDD - VOH)	H) X IOH} + Σ (VOI X IOL)
2: Voltage spikes below Vss at the $\overline{\text{MCLR}}$ pin, inducing currents greater than 80 mA	<u>, may cause latch-up.</u>

- 2: Voltage spikes below Vss at the MCLR pin, inducing currents greater than 80 mA, may cause latch-up. Thus, a series resistor of 50-100Ω should be used when applying a "low" level to the MCLR pin rather than pulling this pin directly to Vss.
- 3: PORTD and PORTE are not implemented on PIC16F873A/876A devices.

† NOTICE: Stresses above those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operation listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

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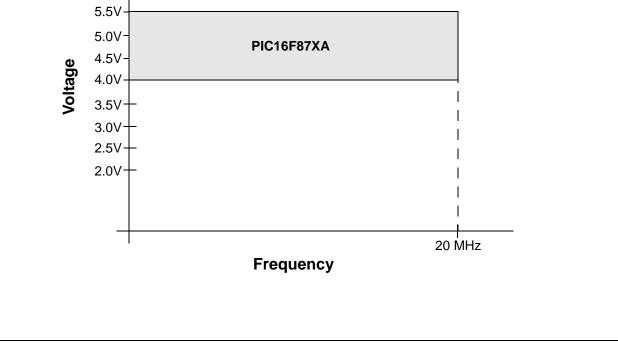
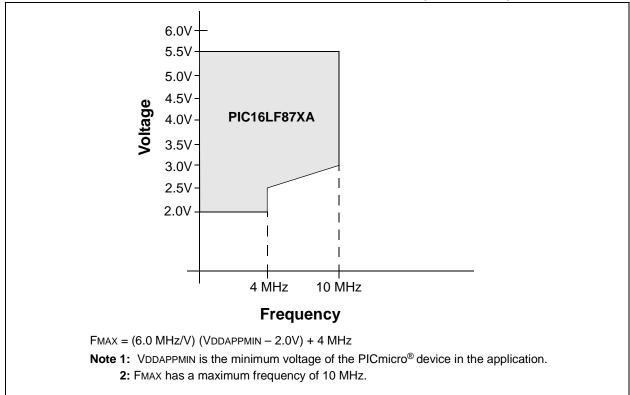


FIGURE 17-2: PIC16LF87XA VOLTAGE-FREQUENCY GRAPH (INDUSTRIAL)



17.1 DC Characteristics: PIC16F873A/874A/876A/877A (Industrial, Extended) PIC16LF873A/874A/876A/877A (Industrial)

(Industrial)		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrialStandard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial $-40^{\circ}C \le TA \le +85^{\circ}C$ for extended						
	Vdd	Supply Voltage						
D001		16LF87XA	2.0	—	5.5	V	All configurations (DC to 10 MHz)	
D001		16F87XA	4.0		5.5	V	All configurations	
D001A			VBOR		5.5	V	BOR enabled, FMAX = 14 MHz ⁽⁷⁾	
D002	Vdr	RAM Data Retention Voltage ⁽¹⁾	—	1.5	_	V		
D003	VPOR	VDD Start Voltage to ensure internal Power-on Reset signal	—	Vss	_	V	See Section 14.5 "Power-on Reset (POR)" for details	
D004	SVDD	VDD Rise Rate to ensure internal Power-on Reset signal	0.05	_	_	V/ms	See Section 14.5 "Power-on Reset (POR)" for details	
D005	VBOR	Brown-out Reset Voltage	3.65	4.0	4.35	V	BODEN bit in configuration word enabled	

Legend: Rows with standard voltage device data only are shaded for improved readability.

- † Data in "Typ" column is at 5V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.
- **Note 1:** This is the limit to which VDD can be lowered without losing RAM data.
 - 2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading, switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

 \overline{MCLR} = VDD; WDT enabled/disabled as specified.

- **3:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD and Vss.
- 4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in k Ω .
- **5:** Timer1 oscillator (when enabled) adds approximately 20 μA to the specification. This value is from characterization and is for design guidance only. This is not tested.
- 6: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.
- 7: When BOR is enabled, the device will operate correctly until the VBOR voltage trip point is reached.

17.1 DC Characteristics: PIC16F873A/874A/876A/877A (Industrial, Extended) PIC16LF873A/874A/876A/877A (Industrial) (Continued)

PIC16LF873A/874A/876A/877A (Industrial) PIC16F873A/874A/876A/877A (Industrial, Extended)		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
		Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial $-40^{\circ}C \le TA \le +125^{\circ}C$ for extended					
Param No.	Symbol	Characteristic/ Device	Min	Тур†	Max	Units	Conditions
	Idd	Supply Current ^(2,5)					
D010		16LF87XA		0.6	2.0	mA	XT, RC osc configurations, Fosc = 4 MHz, VDD = 3.0V
D010		16F87XA	_	1.6	4	mA	XT, RC osc configurations, Fosc = 4 MHz, VDD = 5.5V
D010A		16LF87XA		20	35	μA	LP osc configuration, Fosc = 32 kHz, VDD = 3.0V, WDT disabled
D013		16F87XA		7	15	mA	HS osc configuration, Fosc = 20 MHz, VDD = 5.5V
D015	Δ Ibor	Brown-out Reset Current ⁽⁶⁾		85	200	μA	BOR enabled, VDD = 5.0V

Legend: Rows with standard voltage device data only are shaded for improved readability.

† Data in "Typ" column is at 5V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.

- Note 1: This is the limit to which VDD can be lowered without losing RAM data.
 - 2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading, switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.
 - The test conditions for all IDD measurements in active operation mode are:
 - OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

 $\overline{MCLR} = VDD$; WDT enabled/disabled as specified.

- **3:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD and Vss.
- 4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in kΩ.
- **5:** Timer1 oscillator (when enabled) adds approximately 20 μA to the specification. This value is from characterization and is for design guidance only. This is not tested.
- 6: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.
- 7: When BOR is enabled, the device will operate correctly until the VBOR voltage trip point is reached.

17.1 DC Characteristics: PIC16F873A/874A/876A/877A (Industrial, Extended) PIC16LF873A/874A/876A/877A (Industrial) (Continued)

PIC16LF873A/874A/876A/877A (Industrial) PIC16F873A/874A/876A/877A (Industrial, Extended)			Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial					
			Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial $-40^{\circ}C \le TA \le +125^{\circ}C$ for extended					
Param No. Symbol Characteristic/ Device			Min	Тур†	Max	Units	Conditions	
	IPD	Power-down Current ^(3,5)						
D020		16LF87XA	_	7.5	30	μA	VDD = 3.0V, WDT enabled, -40°C to +85°C	
D020		16F87XA	_	10.5	42 60	μΑ μΑ	VDD = 4.0V, WDT enabled, -40°C to +85°C VDD = 4.0V, WDT enabled, -40°C to +125°C (extended)	
D021		16LF87XA	_	0.9	5	μA	VDD = 3.0V, WDT disabled, 0°C to +70°C	
D021		16F87XA	_	1.5	16 20	μΑ μΑ	VDD = 4.0V, WDT disabled, -40°C to +85°C VDD = 4.0V, WDT disabled, -40°C to +125°C (extended)	
D021A		16LF87XA		0.9	5	μA	VDD = 3.0V, WDT disabled, -40°C to +85°C	
D021A		16F87XA		1.5	19	μA	VDD = 4.0V, WDT disabled, -40°C to +85°C	
D023	ΔIBOR	Brown-out Reset Current ⁽⁶⁾	_	85	200	μA	BOR enabled, VDD = 5.0V	

Legend: Rows with standard voltage device data only are shaded for improved readability.

† Data in "Typ" column is at 5V, 25°C, unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: This is the limit to which VDD can be lowered without losing RAM data.

2: The supply current is mainly a function of the operating voltage and frequency. Other factors, such as I/O pin loading, switching rate, oscillator type, internal code execution pattern and temperature, also have an impact on the current consumption.

The test conditions for all IDD measurements in active operation mode are:

OSC1 = external square wave, from rail-to-rail; all I/O pins tri-stated, pulled to VDD;

 $\overline{MCLR} = VDD$; WDT enabled/disabled as specified.

- **3:** The power-down current in Sleep mode does not depend on the oscillator type. Power-down current is measured with the part in Sleep mode, with all I/O pins in high-impedance state and tied to VDD and Vss.
- 4: For RC osc configuration, current through REXT is not included. The current through the resistor can be estimated by the formula Ir = VDD/2REXT (mA) with REXT in k Ω .
- **5:** Timer1 oscillator (when enabled) adds approximately 20 μA to the specification. This value is from characterization and is for design guidance only. This is not tested.
- 6: The Δ current is the additional current consumed when this peripheral is enabled. This current should be added to the base IDD or IPD measurement.
- 7: When BOR is enabled, the device will operate correctly until the VBOR voltage trip point is reached.

17.2 DC Characteristics: PIC16F873A/874A/876A/877A (Industrial, Extended) PIC16LF873A/874A/876A/877A (Industrial)

DC CHARACTERISTICS			$\begin{array}{llllllllllllllllllllllllllllllllllll$						
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions		
	VIL	Input Low Voltage							
		I/O ports:							
D030		with TTL buffer	Vss	—	0.15 Vdd	V	For entire VDD range		
D030A			Vss	—	0.8V	V	$4.5V \le VDD \le 5.5V$		
D031		with Schmitt Trigger buffer	Vss	—	0.2 Vdd	V			
D032		MCLR, OSC1 (in RC mode)	Vss	—	0.2 Vdd	V			
D033		OSC1 (in XT and LP modes)	Vss	—	0.3V	V	(Note 1)		
		OSC1 (in HS mode)	Vss	—	0.3 Vdd	V			
		Ports RC3 and RC4:		—					
D034		with Schmitt Trigger buffer	Vss	—	0.3 Vdd	V	For entire VDD range		
D034A		with SMBus	-0.5	—	0.6	V	For VDD = 4.5 to 5.5V		
	Viн	Input High Voltage							
		I/O ports:		—					
D040		with TTL buffer	2.0	—	Vdd	V	$4.5V \le VDD \le 5.5V$		
D040A			0.25 VDD + 0.8V		Vdd	V	For entire VDD range		
D041		with Schmitt Trigger buffer	0.8 Vdd	—	Vdd	V	For entire VDD range		
D042		MCLR	0.8 Vdd	—	Vdd	V			
D042A		OSC1 (in XT and LP modes)	1.6V	—	Vdd	V	(Note 1)		
		OSC1 (in HS mode)	0.7 Vdd	—	Vdd	V			
D043		OSC1 (in RC mode)	0.9 Vdd	—	Vdd	V			
		Ports RC3 and RC4:							
D044		with Schmitt Trigger buffer	0.7 Vdd	—	Vdd	V	For entire VDD range		
D044A		with SMBus	1.4	_	5.5	V	For VDD = 4.5 to 5.5V		
D070	IPURB	PORTB Weak Pull-up Current	50	250	400	μA	Vdd = 5V, Vpin = Vss, -40°C to +85°C		
	liL	Input Leakage Current ^(2, 3)							
D060		I/O ports		—	±1	μA	$Vss \le VPIN \le VDD,$ pin at high-impedance		
D061		MCLR, RA4/T0CKI	_	—	±5	μA	$VSS \le VPIN \le VDD$		
D063		OSC1	-	_	±5	μA	Vss \leq VPIN \leq VDD, XT, HS and LP osc configuration		

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKI pin is a Schmitt Trigger input. It is not recommended that the PIC16F87XA be driven with external clock in RC mode.

2: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

3: Negative current is defined as current sourced by the pin.

17.2 DC Characteristics: PIC16F873A/874A/876A/877A (Industrial, Extended) PIC16LF873A/874A/876A/877A (Industrial) (Continued)

		RISTICS	Standard Operating Conditions (unless otherwise stated)Operating temperature $-40^{\circ}C \le TA \le +85^{\circ}C$ for industrial $-40^{\circ}C \le TA \le +125^{\circ}C$ for extendedOperating voltage VDD range as described in DC specification (Section 17.1)					
Param No.	Sym	Characteristic	Min	Тур†	Max	Units	Conditions	
	Vol	Output Low Voltage						
D080		I/O ports	—		0.6	V	IOL = 8.5 mA, VDD = 4.5V, -40°C to +85°C	
D083		OSC2/CLKO (RC osc config)	—	—	0.6	V	IOL = 1.6 mA, VDD = 4.5V, -40°C to +85°C	
	Vон	Output High Voltage						
D090		I/O ports ⁽³⁾	Vdd - 0.7	—	—	V	IOH = -3.0 mA, VDD = 4.5V, -40°C to +85°C	
D092		OSC2/CLKO (RC osc config)	Vdd - 0.7	—	—	V	IOH = -1.3 mA, VDD = 4.5V, -40°С to +85°С	
D150*	Vod	Open-Drain High Voltage	—		8.5	V	RA4 pin	
		Capacitive Loading Specs on Output Pins						
D100	Cosc2	OSC2 pin	—	—	15	pF	In XT, HS and LP modes when external clock is used to drive OSC1	
D101	Сю	All I/O pins and OSC2 (RC mode)	—		50	pF		
D102	Св	SCL, SDA (I ² C mode)	—	—	400	pF		
D120	ED	Data EEPROM Memory Endurance	100K	1M		E/W	-40°C to +85°C	
D120 D121	Vdrw	VDD for read/write	VMIN		 5.5	V	Using EECON to read/write, VMIN = min. operating voltage	
D122	TDEW	Erase/write cycle time	_	4	8	ms		
		Program Flash Memory						
D130	Eр	Endurance	10K	100K	_	E/W	-40°C to +85°C	
D131	Vpr	VDD for read	VMIN	—	5.5	V	VMIN = min. operating voltage	
D132A		VDD for erase/write	VMIN		5.5	V	Using EECON to read/write, VMIN = min. operating voltage	
D133	TPEW	Erase/Write cycle time	—	4	8	ms		

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: In RC oscillator configuration, the OSC1/CLKI pin is a Schmitt Trigger input. It is not recommended that the PIC16F87XA be driven with external clock in RC mode.

2: The leakage current on the MCLR pin is strongly dependent on the applied voltage level. The specified levels represent normal operating conditions. Higher leakage current may be measured at different input voltages.

3: Negative current is defined as current sourced by the pin.

COMPARATOR SPECIFICATIONS TABLE 17-1:

Operating	Operating Conditions: $3.0V < VDD < 5.5V, -40^{\circ}C < TA < +85^{\circ}C$ (unless otherwise stated) $4.0V < VDD < 5.5V, -40^{\circ}C < TA < +125^{\circ}C$ (unless otherwise stated)											
Param No.	Sym	Characteristics	Min	Тур	Max	Units	Comments					
D300	VIOFF	Input Offset Voltage		± 5.0	± 10	mV						
D301	VICM	Input Common Mode Voltage*	0	-	Vdd - 1.5	V						
D302	CMRR	Common Mode Rejection Ratio*	55	-	_	dB						
300 300A	TRESP	Response Time* ⁽¹⁾	_	150	400 600	ns ns	PIC16F87XA PIC16LF87XA					
301	TMC20V	Comparator Mode Change to Output Valid*		—	10	μs						

These parameters are characterized but not tested.

Note 1: Response time measured with one comparator input at (VDD - 1.5)/2 while the other input transitions from Vss to VDD.

TABLE 17-2: VOLTAGE REFERENCE SPECIFICATIONS

Operating	Operating Conditions: 3.0V < VDD < 5.5V, -40°C < TA < +85°C (unless otherwise stated) 4.0V < VDD < 5.5V, -40°C < TA < +125°C (unless otherwise stated)										
Spec No.	Sym	Characteristics	Min	Тур	Мах	Units	Comments				
D310	Vres	Resolution	Vdd/24		Vdd/32	LSb					
D311	VRAA	Absolute Accuracy			1/2 1/2	LSb LSb	Low Range (VRR = 1) High Range (VRR = 0)				
D312	VRur	Unit Resistor Value (R)*	—	2k	—	Ω					
310	TSET	Settling Time* ⁽¹⁾	—		10	μs					

* These parameters are characterized but not tested.

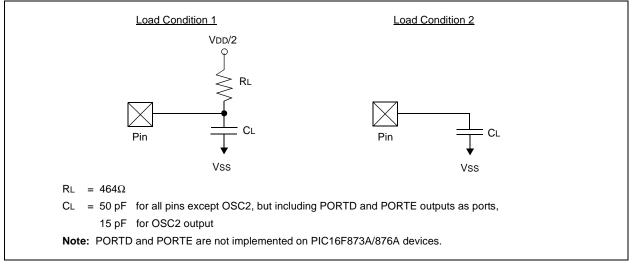
Note 1: Settling time measured while VRR = 1 and VR<3:0> transitions from '0000' to '1111'.

17.3 Timing Parameter Symbology

The timing parameter symbols have been created following one of the following formats:

1. TppS2pp	oS	3. Tcc:st	(I ² C specifications only)
2. TppS		4. Ts	(I ² C specifications only)
Т			
F	Frequency	Т	Time
Lowerca	se letters (pp) and their meanings:		
рр			
СС	CCP1	OSC	OSC1
ck	CLKO	rd	RD
CS	CS	rw	RD or WR
di	SDI	SC	SCK
do	SDO	SS	SS
dt	Data in	tO	TOCKI
io	I/O port	t1	T1CKI
mc	MCLR	wr	WR
Upperca	se letters and their meanings:		
S			
F	Fall	Р	Period
Н	High	R	Rise
I	Invalid (High-impedance)	V	Valid
L	Low	Z	High-impedance
I ² C only			
AA	output access	High	High
BUF	Bus free	Low	Low
Tcc:st (I	² C specifications only)		
CC			
HD	Hold	SU	Setup
ST			
DAT	Data input hold	STO	Stop condition
STA	Start condition		

FIGURE 17-3: LOAD CONDITIONS



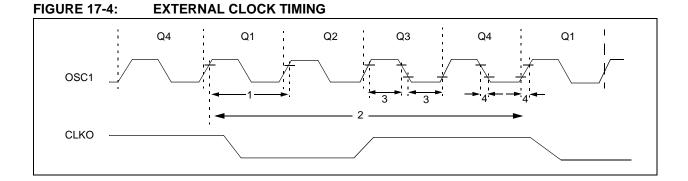


TABLE 17-3: EXTERNAL CLOCK TIMING REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Тур†	Max	Units	Conditions
	Fosc	External CLKI Frequency	DC		1	MHz	XT and RC Osc mode
		(Note 1)	DC	—	20	MHz	HS Osc mode
			DC	—	32	kHz	LP Osc mode
		Oscillator Frequency	DC		4	MHz	RC Osc mode
		(Note 1)	0.1	—	4	MHz	XT Osc mode
			4 5	_	20 200		HS Osc mode LP Osc mode
1	Tosc	External CLKI Period	1000			ns	XT and RC Osc mode
		(Note 1)	50	—	_	ns	HS Osc mode
			5	—	—	μs	LP Osc mode
		Oscillator Period	250	_	—	ns	RC Osc mode
		(Note 1)	250	—	1	μs	XT Osc mode
			100	—	250	ns	HS Osc mode
			50	—	250	ns	HS Osc mode
			31.25	_	—	μs	LP Osc mode
2	Тсү	Instruction Cycle Time (Note 1)	200	Тсү	DC	ns	Tcy = 4/Fosc
3	TosL,	External Clock in (OSC1) High or	100	_	—	ns	XT oscillator
	TosH	Low Time	2.5	—	—	μs	LP oscillator
			15	—	—	ns	HS oscillator
4	TosR,	External Clock in (OSC1) Rise or	—	—	25	ns	XT oscillator
	TosF	Fall Time	—	—	50	ns	LP oscillator
			—	—	15	ns	HS oscillator

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: Instruction cycle period (TCY) equals four times the input oscillator time base period. All specified values are based on characterization data for that particular oscillator type, under standard operating conditions, with the device executing code. Exceeding these specified limits may result in an unstable oscillator operation and/or higher than expected current consumption. All devices are tested to operate at "min." values with an external clock applied to the OSC1/CLKI pin. When an external clock input is used, the "max." cycle time limit is "DC" (no clock) for all devices.



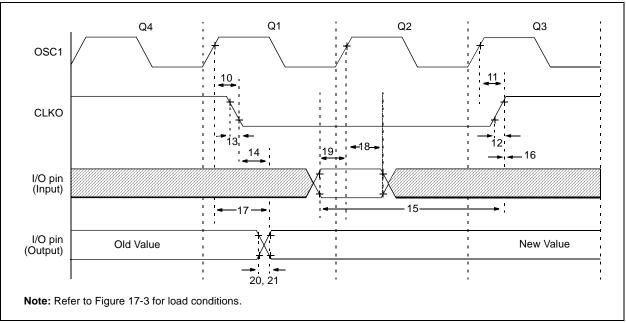


TABLE 17-4: CL	KO AND I/O TIMING REQUIREMENTS
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Param No.	Symbol	Characteristic		Min	Тур†	Max	Units	Conditions
10*	TosH2ckL	OSC1 ↑ to CLKO \downarrow		_	75	200	ns	(Note 1)
11*	TosH2ckH	OSC1 ↑ to CLKO ↑		—	75	200	ns	(Note 1)
12*	ТскR	CLKO Rise Time		—	35	100	ns	(Note 1)
13*	ТскF	CLKO Fall Time	LKO Fall Time		35	100	ns	(Note 1)
14*	TCKL2IOV	CLKO ↓ to Port Out Valid		—	_	0.5 TCY + 20	ns	(Note 1)
15*	TIOV2CKH	Port In Valid before CLKO ↑ T		Tosc + 200		—	ns	(Note 1)
16*	TCKH2IOI	Port In Hold after CLKO ↑		0		—	ns	(Note 1)
17*	TosH2IoV	OSC1 ↑ (Q1 cycle) to Port Out Valid		—	100	255	ns	
18*	TosH2iol	OSC1 ↑ (Q2 cycle) to Port Input	Standard (F)	100	_	—	ns	
		Invalid (I/O in hold time)	Extended (LF)	200	_	—	ns	
19*	TIOV20sH	Port Input Valid to OSC1 \uparrow (I/O in	setup time)	0	_	—	ns	
20*	TIOR	Port Output Rise Time	Standard (F)	—	10	40	ns	
			Extended (LF)	—		145	ns	
21*	TIOF	Port Output Fall Time	Standard (F)	—	10	40	ns	
			Extended (LF)	—		145	ns	
22††*	TINP	INT pin High or Low Time		TCY	—	—	ns	
23††*	Trbp	RB7:RB4 Change INT High or Lo	w Time	Тсү	—	—	ns	

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

these parameters are asynchronous events not related to any internal clock edges.

Note 1: Measurements are taken in RC mode where CLKO output is 4 x Tosc.

*

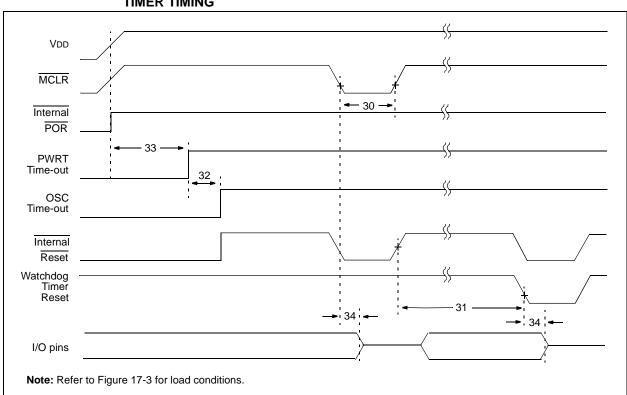


FIGURE 17-6: RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER AND POWER-UP TIMER TIMING

FIGURE 17-7: BROWN-OUT RESET TIMING

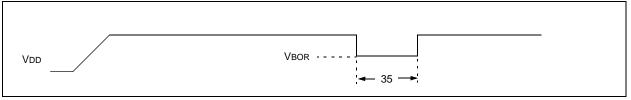


TABLE 17-5:RESET, WATCHDOG TIMER, OSCILLATOR START-UP TIMER, POWER-UP TIMER
AND BROWN-OUT RESET REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Тур†	Max	Units	Conditions
30	TMCL	MCLR Pulse Width (low)	2	_		μs	VDD = 5V, -40°C to +85°C
31*	Twdt	Watchdog Timer Time-out Period (no prescaler)	7	18	33	ms	VDD = 5V, -40°C to +85°C
32	Tost	Oscillation Start-up Timer Period	—	1024 Tosc	_	_	Tosc = OSC1 period
33*	TPWRT	Power-up Timer Period	28	72	132	ms	VDD = 5V, -40°C to +85°C
34	Tıoz	I/O High-Impedance from MCLR Low or Watchdog Timer Reset	—		2.1	μs	
35	TBOR	Brown-out Reset Pulse Width	100	—		μs	$VDD \leq VBOR (D005)$

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.



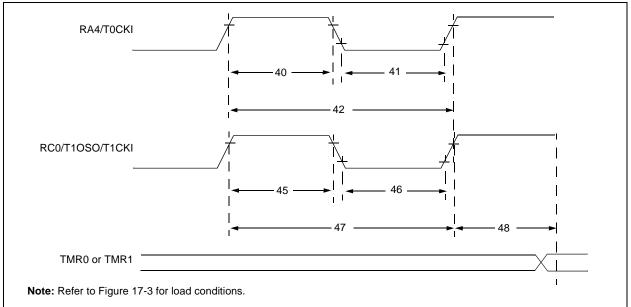


TABLE 17-6: TIMER0 AND TIMER1 EXTERNAL CLOCK REQUIREMENTS

Param No.	Symbol		Characteristic		Min	Тур†	Max	Units	Conditions	
40*	T⊤0H	T0CKI High Pulse	e Width	No Prescaler	0.5 TCY + 20	—	_	ns	Must also meet	
		_		With Prescaler	10	—		ns	parameter 42	
41*	T⊤0L	T0CKI Low Pulse	Width	No Prescaler	0.5 TCY + 20	_	_	ns	Must also meet	
				With Prescaler	10	_	_	ns	parameter 42	
42*	T⊤0P	T0CKI Period		No Prescaler	Tcy + 40	—	_	ns		
				With Prescaler	Greater of:	—	_	ns	N = prescale value	
					20 or <u>TCY + 40</u>				(2, 4,, 256)	
					N					
45*	T⊤1H	T1CKI High	Synchronous, Pre	scaler = 1	0.5 TCY + 20	—	_	ns	Must also meet	
		Time	Synchronous,	Standard(F)	15	—	_	ns	parameter 47	
			Prescaler = 2, 4, 8	Extended(LF)	25			ns		
			Asynchronous	Standard(F)	30	-		ns		
				Extended(LF)	50	Ι	—	ns		
46*	T⊤1L	T1CKI Low Time	Synchronous, Pre	scaler = 1	0.5 TCY + 20			ns	Must also meet	
			Synchronous, Prescaler = 2, 4, 8 Asynchronous	Standard(F)	15			ns	parameter 47	
				Extended(LF)	25	-		ns		
				Standard(F)	30	Ι	—	ns		
				Extended(LF)	50	Ι	—	ns		
47*	Tt1P	T1CKI Input Period	Synchronous	Standard(F)	Greater of: 30 or <u>Tcy + 40</u> N	_	_	ns	N = prescale value (1, 2, 4, 8)	
				Extended(LF)	Greater of: 50 or <u>Tcy + 40</u> N				N = prescale value (1, 2, 4, 8)	
			Asynchronous	Standard(F)	60	—	_	ns		
				Extended(LF)	100	—	—	ns		
	F⊤1		Input Frequency R d by setting bit T10	0	DC	_	200	kHz		
48	TCKEZTMR1	Delay from Extern	nal Clock Edge to T	Fimer Increment	2 Tosc	—	7 Tosc	—		

These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

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FIGURE 17-9: CAPTURE/COMPARE/PWM TIMINGS (CCP1 AND CCP2)

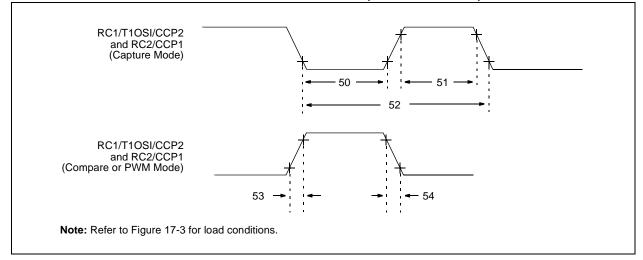


TABLE 17-7: CAPTURE/COMPARE/PWM REQUIREMENTS (CCP1 AND CCP2)

Param No.	Symbol	Characteristic			Min	Тур†	Max	Units	Conditions
50*	TccL	CCP1 and CCP2 Input Low Time	No Prescaler	No Prescaler			_	ns	
			With Prescaler	Standard(F)	10	_	_	ns	
			with Prescaler	Extended(LF)	20	_	—	ns	
51* To	ТссН		No Prescaler	No Prescaler		_	_	ns	
		Input High Time	With Prescaler	Standard(F)	10	_	_	ns	
				Extended(LF)	20	_		ns	
52*	TCCP	CCP1 and CCP2 In	put Period	•	<u>3 TCY + 40</u>			ns	N = prescale value
			N				(1, 4 or 16)		
53*	TCCR	CCP1 and CCP2 Out	utput Rise Time	Standard(F)	—	10	25	ns	
		Extended(LF)			—	25	50	ns	
54*	TCCF	CCP1 and CCP2 Ou	utput Fall Time	Standard(F)	—	10	25	ns	
			Extended(LF)	_	25	45	ns		

These parameters are characterized but not tested.

t Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.



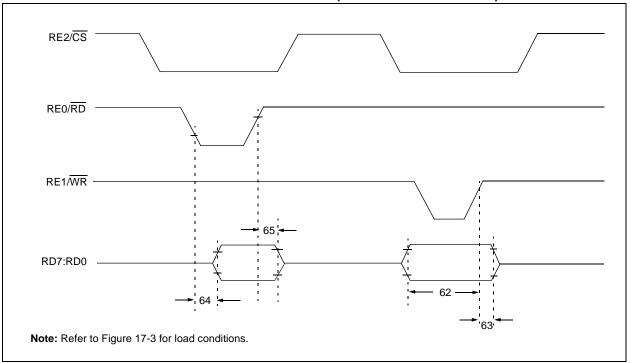


TABLE 17-8: PARALLEL SLAVE PORT REQUIREMENTS (PIC16F874A/877A ONLY)

Param No.	Symbol	Characteristic			Тур†	Max	Units	Conditions
62	TdtV2wrH	Data In Valid before $\overline{WR} \uparrow$ or $\overline{CS} \uparrow$ (setup time)			—		ns	
63*	TwrH2dtl	\overline{WR} \uparrow or \overline{CS} \uparrow to Data–in Invalid	Standard(F)	20	_	_	ns	
		(hold time) Extended(LF)		35	_		ns	
64	TrdL2dtV	$\overline{RD} \downarrow$ and $\overline{CS} \downarrow$ to Data–out Valid			—	80	ns	
65	TrdH2dtI	$\overline{RD} \uparrow or \overline{CS} \downarrow to Data-out Invalid$		10	_	30	ns	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

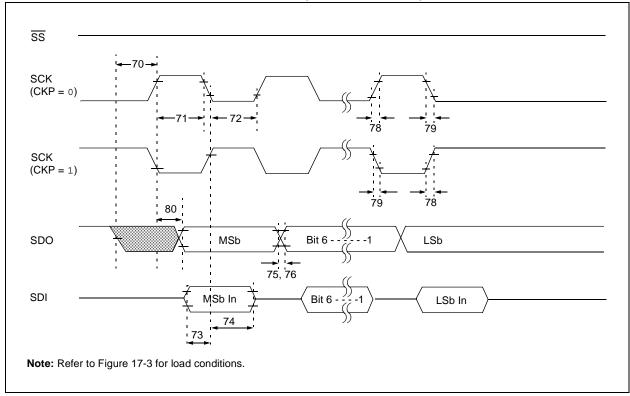
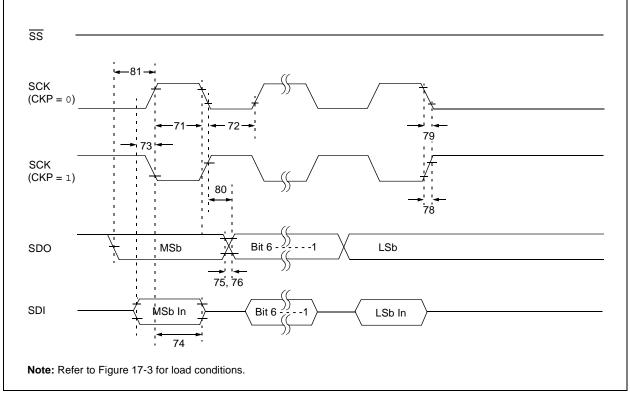
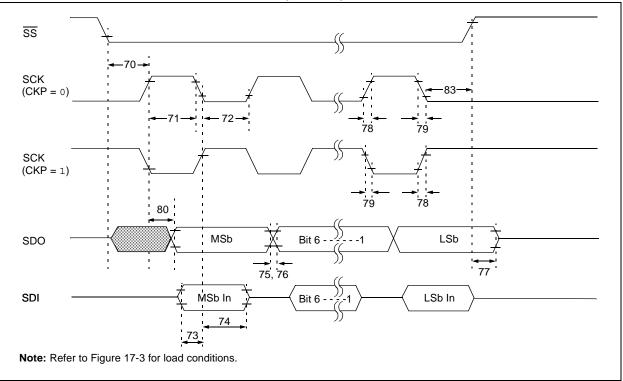


FIGURE 17-11: SPI MASTER MODE TIMING (CKE = 0, SMP = 0)

FIGURE 17-12:SPI MASTER MODE TIMING (CKE = 1, SMP = 1)







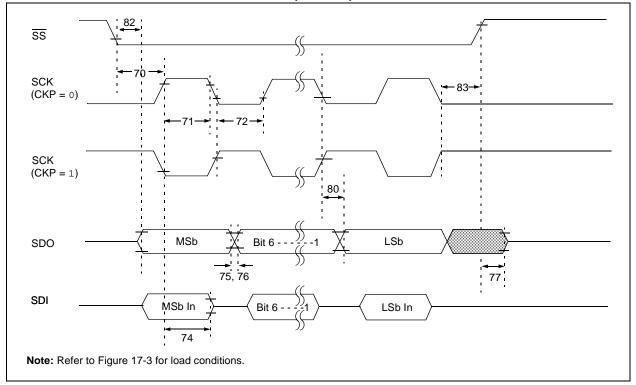


FIGURE 17-14: SPI SLAVE MODE TIMING (CKE = 1)

Param No.	Symbol	Characteristic		Min	Тур†	Мах	Units	Conditions
70*	TssL2scH, TssL2scL	$\overline{SS} \downarrow$ to SCK \downarrow or SCK \uparrow Input		Тсү	_	—	ns	
71*	TscH	SCK Input High Time (Slave mode)		TCY + 20	_	-	ns	
72*	TscL	SCK Input Low Time (Slave mode)	TCY + 20	_	-	ns		
73*	TDIV2scH, TDIV2scL	Setup Time of SDI Data Input to SC	100	_	_	ns		
74*	TscH2DIL, TscL2DIL	Hold Time of SDI Data Input to SCK	100	—	—	ns		
75*	TDOR	SDO Data Output Rise Time	Standard(F) Extended(LF)		10 25	25 50	ns ns	
76*	TDOF	SDO Data Output Fall Time		_	10	25	ns	
77*	TssH2doZ	SS ↑ to SDO Output High-Impedance	ce	10		50	ns	
78*	TscR	SCK Output Rise Time (Master mode)	Standard(F) Extended(LF)		10 25	25 50	ns ns	
79*	TscF	SCK Output Fall Time (Master mode	e)	_	10	25	ns	
80*	TscH2doV, TscL2doV	SDO Data Output Valid after SCK Edge	Standard(F) Extended(LF)		_	50 145	ns	
81*	TDOV2scH, TDOV2scL	SDO Data Output Setup to SCK Edg	Тсү	—	—	ns		
82*	TssL2doV	SDO Data Output Valid after $\overline{SS} \downarrow E$	_	_	50	ns		
83*	TscH2ssH, TscL2ssH	SS ↑ after SCK Edge	1.5 Tcy + 40	_	—	ns		

TABLE 17-9: SPI MODE REQUIREMENTS

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

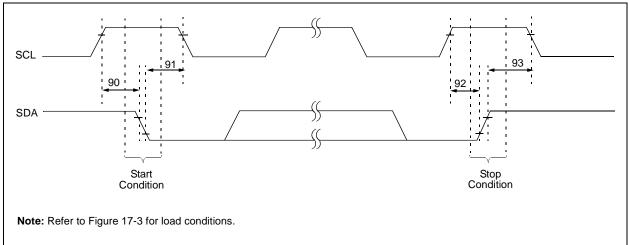
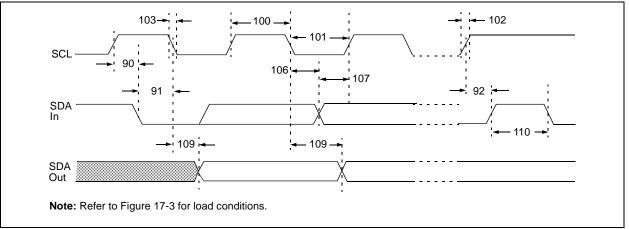


FIGURE 17-15: I²C BUS START/STOP BITS TIMING

Param No.	Symbol	Chara	octeristic	Min	Тур	Max	Units	Conditions
90	TSU:STA	Start condition	100 kHz mode	4700	_	—	ns	Only relevant for Repeated Start
		Setup time	400 kHz mode	600	_	_		condition
91	THD:STA	Start condition	100 kHz mode	4000	_	_	ns	After this period, the first clock pulse
		Hold time	400 kHz mode	600		_		is generated
92	Tsu:sto	Stop condition	100 kHz mode	4700		_	ns	
		Setup time	400 kHz mode	600		_		
93	THD:STO	Stop condition	100 kHz mode	4000		-	ns	
		Hold time	400 kHz mode	600	_	-		

TABLE 17-10: I²C BUS START/STOP BITS REQUIREMENTS

FIGURE 17-16: I²C BUS DATA TIMING



Param No.	Sym	Characte	eristic	Min	Max	Units	Conditions
100	Тнідн	Clock High Time	100 kHz mode	4.0	_	μs	
			400 kHz mode	0.6		μs	
			SSP Module	0.5 TCY			
101	TLOW	Clock Low Time	100 kHz mode	4.7	—	μs	
			400 kHz mode	1.3	—	μs	
			SSP Module	0.5 TCY			
102	Tr	SDA and SCL Rise	100 kHz mode	—	1000	ns	
		Time	400 kHz mode	20 + 0.1 Cв	300	ns	Cb is specified to be from 10 to 400 pF
103	TF	SDA and SCL Fall	100 kHz mode	—	300	ns	
		Time	400 kHz mode	20 + 0.1 Cв	300	ns	CB is specified to be from 10 to 400 pF
90	TSU:STA	Start Condition Setup	100 kHz mode	4.7	_	μs	Only relevant for Repeated Start
		Time	400 kHz mode	0.6		μs	condition
91	THD:STA	Start Condition Hold	100 kHz mode	4.0	_	μs	After this period, the first clock
		Time	400 kHz mode	0.6	—	μs	pulse is generated
106	THD:DAT	Data Input Hold Time	100 kHz mode	0	_	ns	
			400 kHz mode	0	0.9	μs	1
107	TSU:DAT	Data Input Setup Time	100 kHz mode	250	_	ns	(Note 2)
			400 kHz mode	100	_	ns	
92	Tsu:sto	Stop Condition Setup	100 kHz mode	4.7	—	μs	
		Time	400 kHz mode	0.6	_	μs	
109	ΤΑΑ	Output Valid from	100 kHz mode	—	3500	ns	(Note 1)
		Clock	400 kHz mode	_	_	ns	
110	TBUF	Bus Free Time	100 kHz mode	4.7	—	μs	Time the bus must be free before
			400 kHz mode	1.3	_	μs	a new transmission can start
	Св	Bus Capacitive Loading]	_	400	pF	

TABLE 17-11: I²C BUS DATA REQUIREMENTS

Note 1: As a transmitter, the device must provide this internal minimum delay time to bridge the undefined region (min. 300 ns) of the falling edge of SCL to avoid unintended generation of Start or Stop conditions.

2: A fast mode (400 kHz) I²C bus device can be used in a standard mode (100 kHz) I²C bus system, but the requirement that, TSU:DAT ≥ 250 ns, must then be met. This will automatically be the case if the device does not stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal. If such a device does stretch the LOW period of the SCL signal, it must output the next data bit to the SDA line, TR MAX. + TSU:DAT = 1000 + 250 = 1250 ns (according to the standard mode I²C bus specification), before the SCL line is released.

FIGURE 17-17: USART SYNCHRONOUS TRANSMISSION (MASTER/SLAVE) TIMING

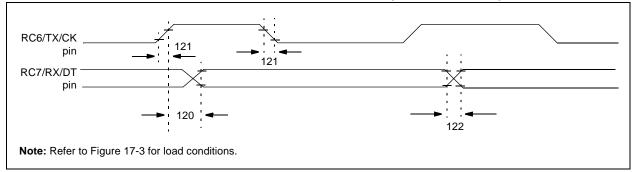


TABLE 17-12: USART SYNCHRONOUS TRANSMISSION REQUIREMENTS

Param No.	Symbol	Characteristic	Characteristic				Units	Conditions
120	TCKH2DTV	SYNC XMIT (MASTER & SLAVE) Clock High to Data Out Valid	Standard(F)		_	80	ns	
			Extended(LF)	_	—	100	ns	
121	TCKRF	Clock Out Rise Time and Fall Time	Standard(F)	_	_	45	ns	
	(Master mode)		Extended(LF)	_	_	50	ns	
122	TDTRF	Data Out Rise Time and Fall Time	Standard(F)	_	—	45	ns	
			Extended(LF)	_	—	50	ns	

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

FIGURE 17-18: USART SYNCHRONOUS RECEIVE (MASTER/SLAVE) TIMING

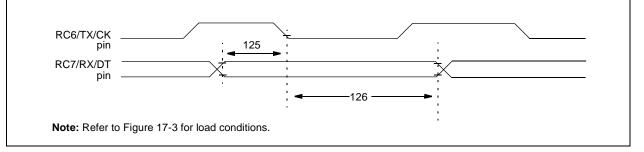


TABLE 17-13: USART SYNCHRONOUS RECEIVE REQUIREMENTS

Param No.	Symbol	Characteristic	Min	Тур†	Max	Units	Conditions
125	TDTV2CKL	SYNC RCV (MASTER & SLAVE)					
		Data Setup before CK \downarrow (DT setup time)	15	—	—	ns	
126	TCKL2DTL	Data Hold after CK \downarrow (DT hold time)	15	_	_	ns	

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

TABLE 17-14: A/D CONVERTER CHARACTERISTICS:PIC16F873A/874A/876A/877A (INDUSTRIAL) PIC16LF873A/874A/876A/877A (INDUSTRIAL)

Param No.	Sym	Characte	eristic	Min	Тур†	Max	Units	Conditions		
A01	NR	Resolution		_		10-bits	bit	VREF = VDD = 5.12V, VSS \leq VAIN \leq VREF		
A03	EIL	Integral Linearity Er	ror	—	—	< ± 1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$		
A04	Edl	Differential Linearity Error		_	—	< ± 1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$		
A06	EOFF	Offset Error	Offset Error		Offset Error		—	< ± 2	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$
A07	Egn	Gain Error		—	—	< ± 1	LSb	$\label{eq:VREF} \begin{array}{l} VREF = VDD = 5.12V,\\ VSS \leq VAIN \leq VREF \end{array}$		
A10	_	Monotonicity		—	guaranteed ⁽³⁾	—	_	$VSS \leq VAIN \leq VREF$		
A20	Vref	Reference Voltage (VREF+ - VREF-)		2.0	—	Vdd + 0.3	V			
A21	Vref+	Reference Voltage High		AVDD - 2.5V		AVDD + 0.3V	V			
A22	VREF-	Reference Voltage	Low	AVss - 0.3V		VREF+ - 2.0V	V			
A25	VAIN	Analog Input Voltag	е	Vss - 0.3V	—	VREF + 0.3V	V			
A30	ZAIN	Recommended Imp Analog Voltage Sou		—	—	2.5	kΩ	(Note 4)		
A40	IAD	A/D Conversion	PIC16F87XA	—	220	_	μΑ	Average current		
		Current (VDD)	PIC16LF87XA	—	90	—	μA	consumption when A/D is on (Note 1)		
A50	IREF	VREF Input Current	(Note 2)		_	5	μΑ	During VAIN acquisition. Based on differential of VHOLD to VAIN to charge CHOLD, see Section 11.1 "A/D Acquisition Requirements". During A/D conversion		
						100	μΛ	cycle		

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

Note 1: When A/D is off, it will not consume any current other than minor leakage current. The power-down current spec includes any such leakage from the A/D module.

2: VREF current is from RA3 pin or VDD pin, whichever is selected as reference input.

3: The A/D conversion result never decreases with an increase in the input voltage and has no missing codes.

4: Maximum allowed impedance for analog voltage source is 10 kΩ. This requires higher acquisition time.



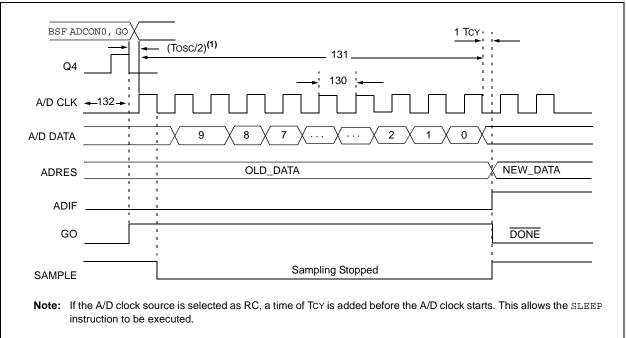


TABLE 17-15:	A/D CONVERSION REQUIREMENTS
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Param No.	Symbol	Charact	eristic	Min	Тур†	Max	Units	Conditions
130	TAD	A/D Clock Period	PIC16F87XA	1.6	_	_	μs	Tosc based, VREF \geq 3.0V
			PIC16LF87XA	3.0		_	μs	Tosc based, VREF $\geq 2.0V$
			PIC16F87XA	2.0	4.0	6.0	μs	A/D RC mode
			PIC16LF87XA	3.0	6.0	9.0	μs	A/D RC mode
131	TCNV	Conversion Time (not (Note 1)		—	12	TAD		
132	TACQ	Acquisition Time	(Note 2)	40	-	μs		
				10*	_	_	μs	The minimum time is the amplifier settling time. This may be used if the "new" input volt- age has not changed by more than 1 LSb (i.e., 20.0 mV @ 5.12V) from the last sampled voltage (as stated on CHOLD).
134	TGO	Q4 to A/D Clock Start	_	Tosc/2 §		_	If the A/D clock source is selected as RC, a time of Tcy is added before the A/D clock starts. This allows the SLEEP instruction to be executed.	

* These parameters are characterized but not tested.

† Data in "Typ" column is at 5V, 25°C unless otherwise stated. These parameters are for design guidance only and are not tested.

§ This specification ensured by design.

- **Note 1:** ADRES register may be read on the following TCY cycle.
 - 2: See Section 11.1 "A/D Acquisition Requirements" for minimum conditions.

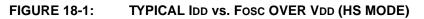
PIC16F87XA

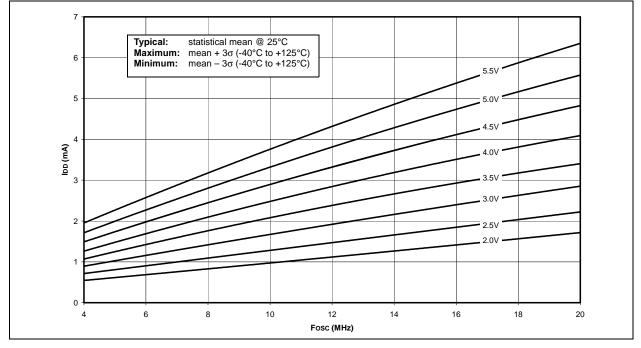
NOTES:

18.0 DC AND AC CHARACTERISTICS GRAPHS AND TABLES

Note: The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and therefore, outside the warranted range.

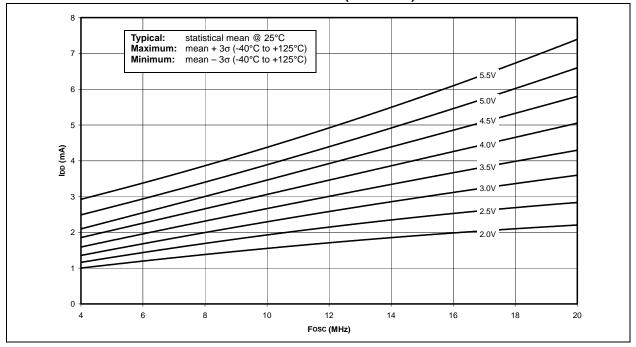
"Typical" represents the mean of the distribution at 25°C. "Maximum" or "minimum" represents (mean + 3σ) or (mean - 3σ) respectively, where σ is a standard deviation, over the whole temperature range.







MAXIMUM IDD vs. Fosc OVER VDD (HS MODE)



PIC16F87XA

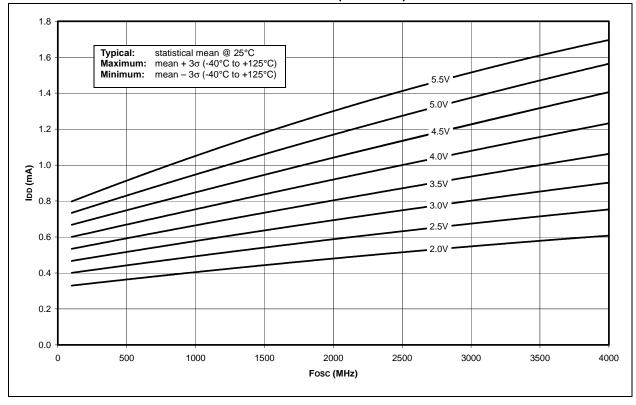
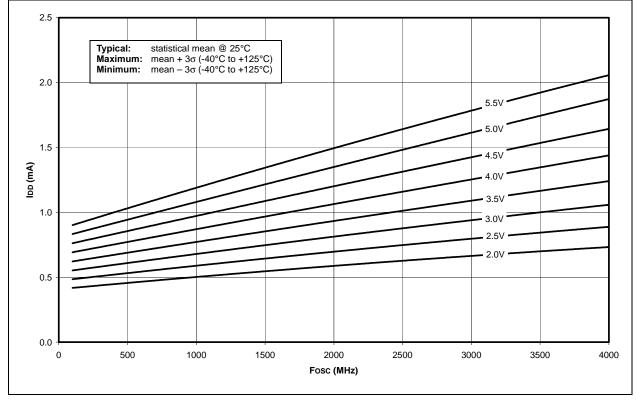


FIGURE 18-3: TYPICAL IDD vs. Fosc OVER VDD (XT MODE)





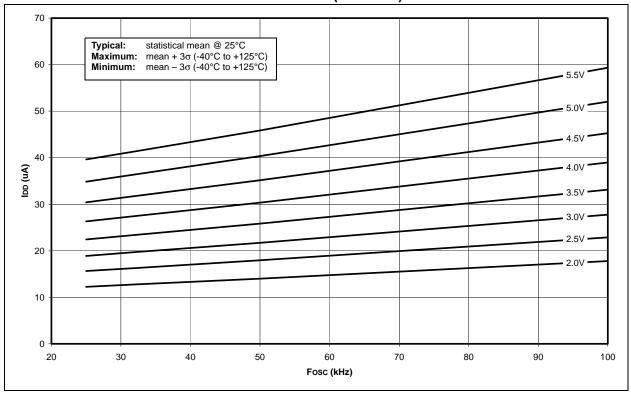
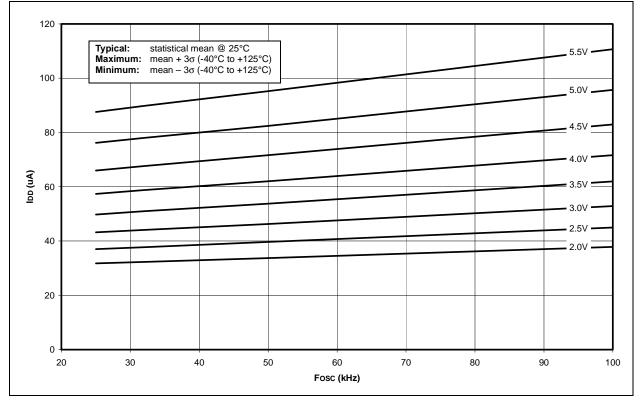


FIGURE 18-5: TYPICAL IDD vs. Fosc OVER VDD (LP MODE)







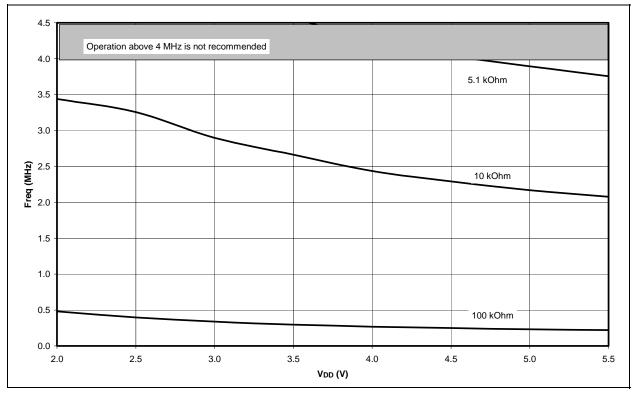
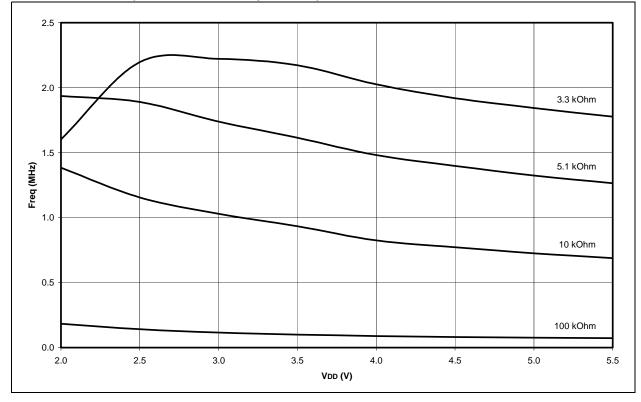
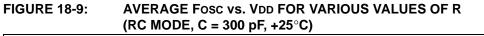
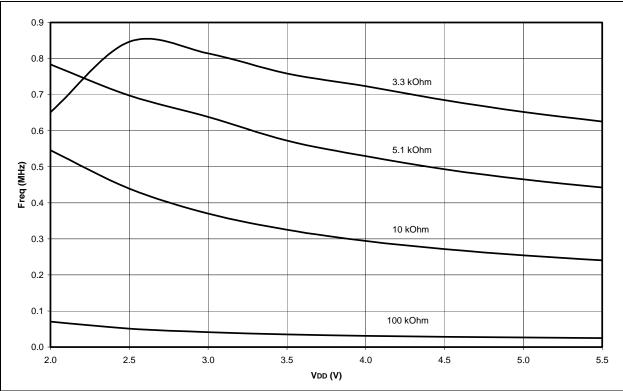


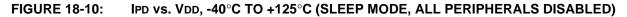
FIGURE 18-8: AVERAGE FOSC vs. VDD FOR VARIOUS VALUES OF R (RC MODE, C = 100 pF, +25°C)



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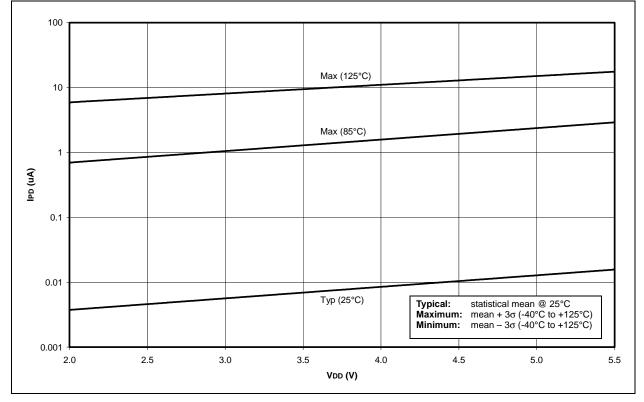
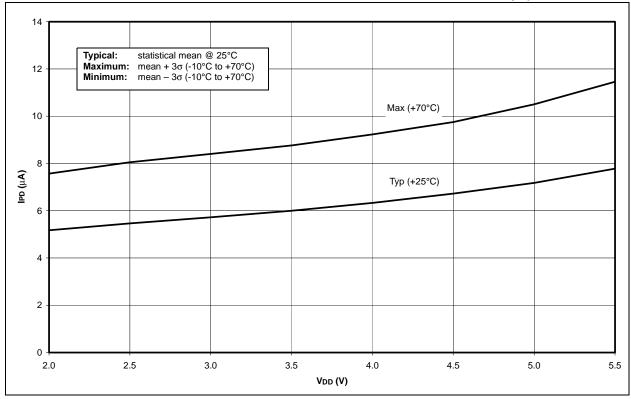
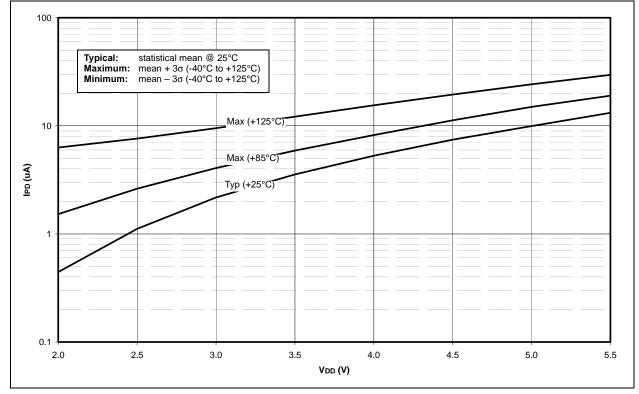


FIGURE 18-11: TYPICAL AND MAXIMUM \triangle ITMR1 vs. VDD OVER TEMPERATURE (-10°C TO +70°C, TIMER1 WITH OSCILLATOR, XTAL = 32 kHz, C1 AND C2 = 47 pF)







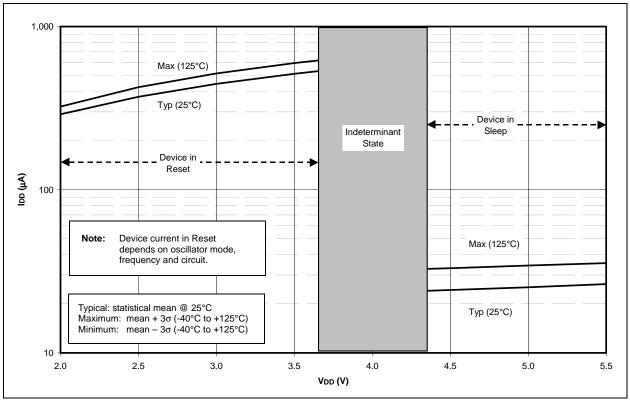
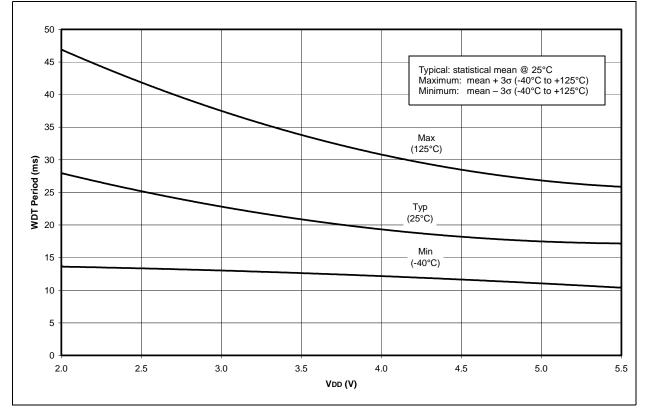


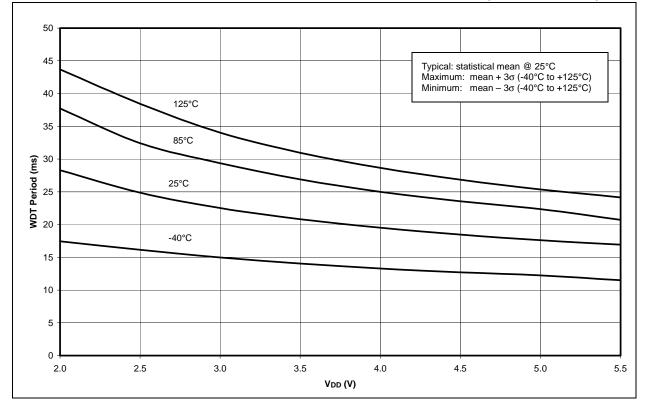
FIGURE 18-13: △IBOR vs. VDD OVER TEMPERATURE

FIGURE 18-14: TYPICAL, MINIMUM AND MAXIMUM WDT PERIOD vs. VDD (-40°C TO +125°C)



PIC16F87XA





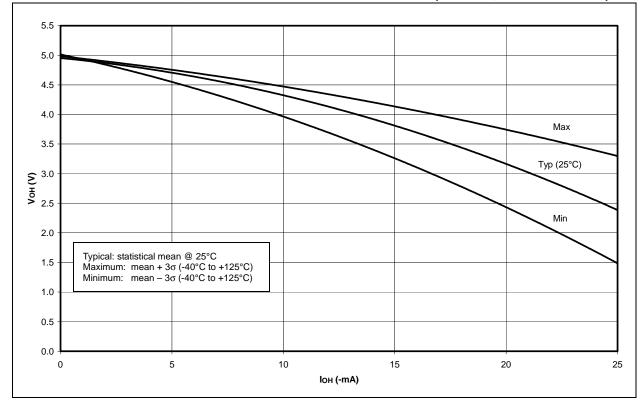


FIGURE 18-16: TYPICAL, MINIMUM AND MAXIMUM VOH vs. IOH (VDD = 5V, -40°C TO +125°C)

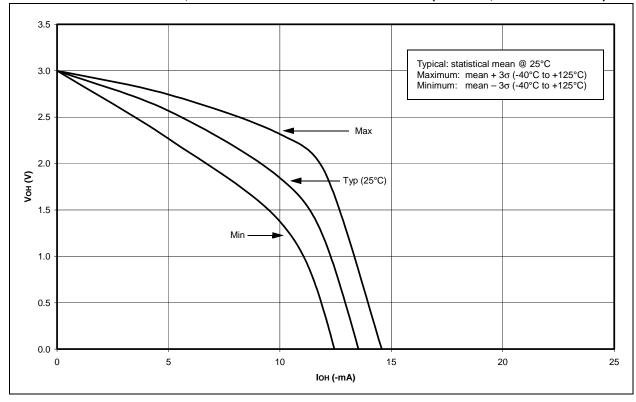
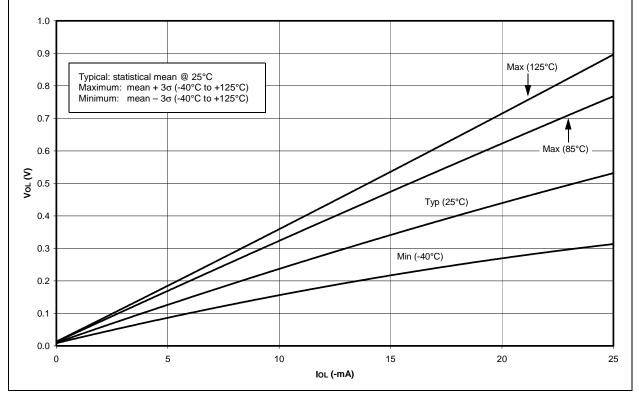
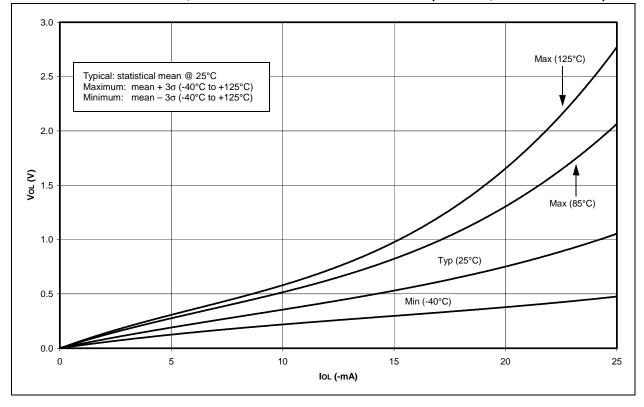


FIGURE 18-17: TYPICAL, MINIMUM AND MAXIMUM VOH vs. IOH (VDD = 3V, -40°C TO +125°C)

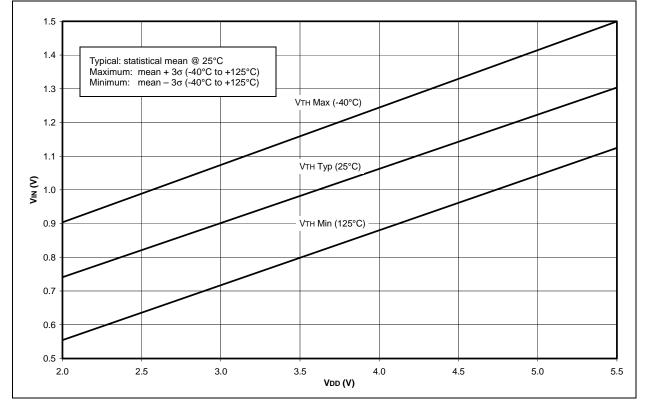












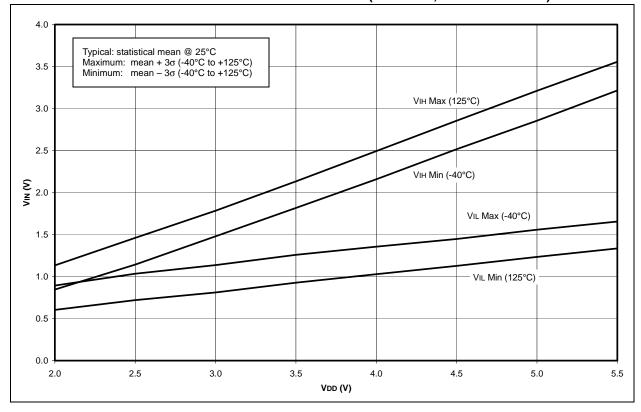
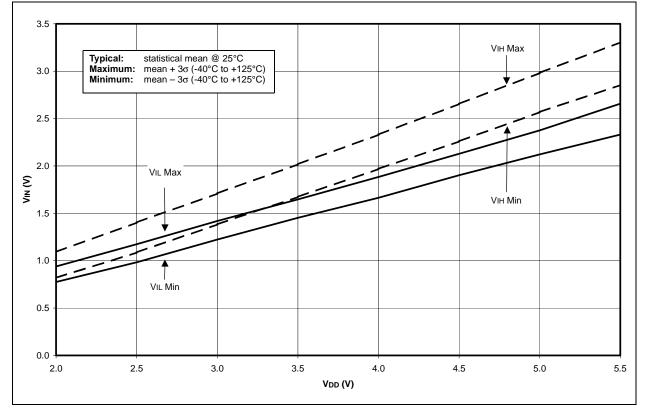
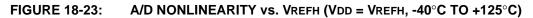
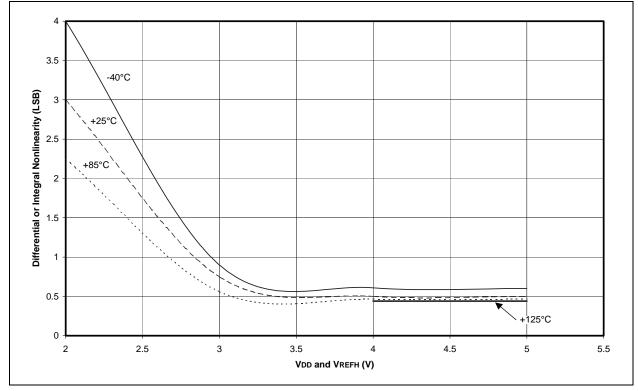


FIGURE 18-21: MINIMUM AND MAXIMUM VIN vs. VDD (ST INPUT, -40°C TO +125°C)

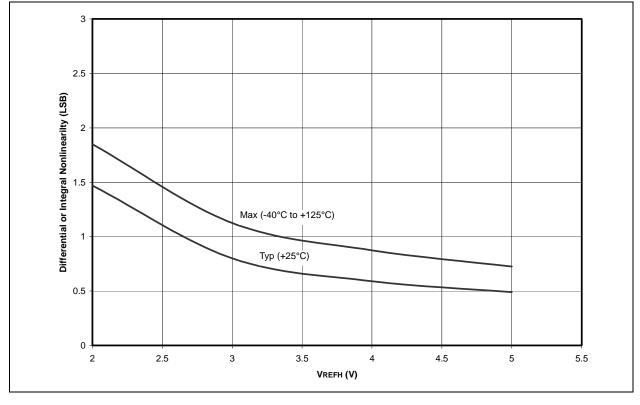








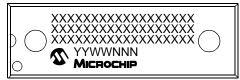




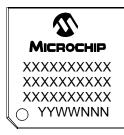
19.0 PACKAGING INFORMATION

19.1 Package Marking Information

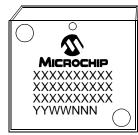
40-Lead PDIP



44-Lead TQFP



44-Lead PLCC



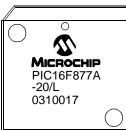
Example



Example



Example



Legen	d: XXX Y YY WW NNN	Customer specific information* Year code (last digit of calendar year) Year code (last 2 digits of calendar year) Week code (week of January 1 is week '01') Alphanumeric traceability code
Note:	be carried	nt the full Microchip part number cannot be marked on one line, it will over to the next line thus limiting the number of available characters her specific information.

* Standard PICmicro device marking consists of Microchip part number, year code, week code, and traceability code. For PICmicro device marking beyond this, certain price adders apply. Please check with your Microchip Sales Office. For QTP devices, any special marking adders are included in QTP price.

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28-Lead QFN



28-Lead SSOP



-I/ML

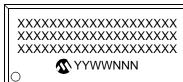
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Ο



Example

Example



28-Lead SOIC











PIC16F876A/SO

10310017

Example



44-Lead QFN

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XXXXXXXXXXX

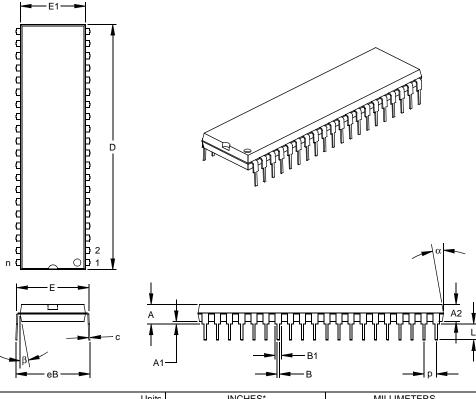
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YYWWNNN

28-Lead PDIP (Skinny DIP)

Package Marking Information (Cont'd)

40-Lead Plastic Dual In-line (P) - 600 mil (PDIP)



Dimension Limits MIN NOM MAX MIN NOM Number of Pins n 40 40 40 40 Pitch P .100 2.54 40 Top to Seating Plane A .160 .175 .190 4.06 4.45 Molded Package Thickness A2 .140 .150 .160 3.56 3.81 Base to Seating Plane A1 .015 0.38 Shoulder to Shoulder Width E .595 .600 .625 15.11 15.24 Molded Package Width E1 .530 .545 .560 13.46 13.84 Overall Length D 2.045 2.058 2.065 51.94 52.26 Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0		1ILLIMETERS	N	INCHES*			Units	
Pitch P .100 2.54 Top to Seating Plane A .160 .175 .190 4.06 4.45 Molded Package Thickness A2 .140 .150 .160 3.56 3.81 Base to Seating Plane A1 .015 0.38 Shoulder to Shoulder Width E .595 .600 .625 15.11 15.24 Molded Package Width E1 .530 .545 .560 13.46 13.84 Overall Length D 2.045 2.058 2.065 51.94 52.26 Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46	MAX	NOM	MIN	MAX	NOM	MIN	Limits	Dimension
Top to Seating Plane A .160 .175 .190 4.06 4.45 Molded Package Thickness A2 .140 .175 .190 4.06 4.45 Molded Package Thickness A2 .140 .150 .160 3.56 3.81 Base to Seating Plane A1 .015 0.38 Shoulder to Shoulder Width E .595 .600 .625 15.11 15.24 Molded Package Width E1 .530 .545 .560 13.46 13.84 Overall Length D 2.045 2.058 2.065 51.94 52.26 Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46 <		40			40		n	Number of Pins
Molded Package Thickness A2 .140 .150 .160 3.56 3.81 Base to Seating Plane A1 .015 0.38 Shoulder to Shoulder Width E .595 .600 .625 15.11 15.24 Molded Package Width E1 .530 .545 .560 13.46 13.84 Overall Length D 2.045 2.058 2.065 51.94 52.26 Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46		2.54			.100		р	Pitch
Base to Seating Plane A1 .015 0.38 Shoulder to Shoulder Width E .595 .600 .625 15.11 15.24 Molded Package Width E1 .530 .545 .560 13.46 13.84 Overall Length D 2.045 2.058 2.065 51.94 52.26 Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46	4.83	4.45	4.06	.190	.175	.160	Α	Top to Seating Plane
Shoulder to Shoulder Width E .595 .600 .625 15.11 15.24 Molded Package Width E1 .530 .545 .560 13.46 13.84 Overall Length D 2.045 2.058 2.065 51.94 52.26 Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46	4.06	3.81	3.56	.160	.150	.140	A2	Molded Package Thickness
Molded Package Width E1 .530 .545 .560 13.46 13.84 Overall Length D 2.045 2.058 2.065 51.94 52.26 Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46			0.38			.015	A1	Base to Seating Plane
Overall Length D 2.045 2.058 2.065 51.94 52.26 Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness C .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46	15.88	15.24	15.11	.625	.600	.595	Е	Shoulder to Shoulder Width
Tip to Seating Plane L .120 .130 .135 3.05 3.30 Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46	14.22	13.84	13.46	.560	.545	.530	E1	Molded Package Width
Lead Thickness c .008 .012 .015 0.20 0.29 Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46	52.45	52.26	51.94	2.065	2.058	2.045	D	Overall Length
Upper Lead Width B1 .030 .050 .070 0.76 1.27 Lower Lead Width B .014 .018 .022 0.36 0.46	3.43	3.30	3.05	.135	.130	.120	L	Tip to Seating Plane
Lower Lead Width B .014 .018 .022 0.36 0.46	0.38	0.29	0.20	.015	.012	.008	С	Lead Thickness
	1.78	1.27	0.76	.070	.050	.030	B1	Upper Lead Width
Overall Row Spacing § eB .620 .650 .680 15.75 16.51	0.56	0.46	0.36	.022	.018	.014	В	Lower Lead Width
	17.27	16.51	15.75	.680	.650	.620	eB	Overall Row Spacing §
Mold Draft Angle Top α 5 10 15 5 10	15	10	5	15	10	5	α	Mold Draft Angle Top
Mold Draft Angle Bottom β 5 10 15 5 10	15	10	5	15	10	5	β	Mold Draft Angle Bottom

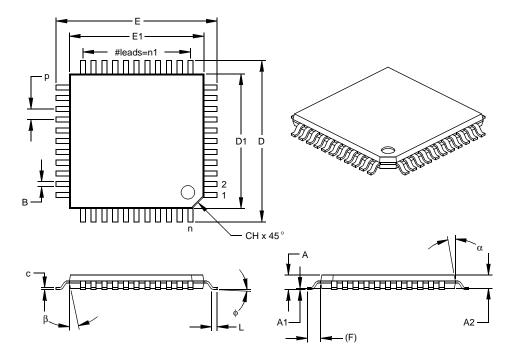
* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MO-011

Drawing No. C04-016

44-Lead Plastic Thin Quad Flatpack (PT) 10x10x1 mm Body, 1.0/0.10 mm Lead Form (TQFP)



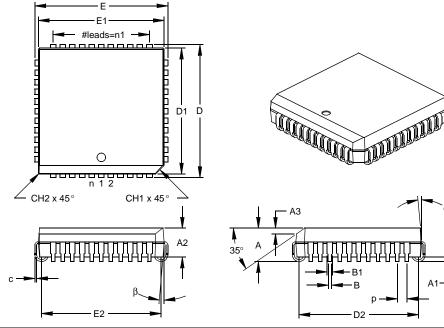
	Units		INCHES		М	ILLIMETERS	*
Dimensio	on Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		44			44	
Pitch	р		.031			0.80	
Pins per Side	n1		11			11	
Overall Height	Α	.039	.043	.047	1.00	1.10	1.20
Molded Package Thickness	A2	.037	.039	.041	0.95	1.00	1.05
Standoff §	A1	.002	.004	.006	0.05	0.10	0.15
Foot Length	L	.018	.024	.030	0.45	0.60	0.75
Footprint (Reference)	(F)		.039		1.00		
Foot Angle	¢	0	3.5	7	0	3.5	7
Overall Width	Е	.463	.472	.482	11.75	12.00	12.25
Overall Length	D	.463	.472	.482	11.75	12.00	12.25
Molded Package Width	E1	.390	.394	.398	9.90	10.00	10.10
Molded Package Length	D1	.390	.394	.398	9.90	10.00	10.10
Lead Thickness	С	.004	.006	.008	0.09	0.15	0.20
Lead Width	В	.012	.015	.017	0.30	0.38	0.44
Pin 1 Corner Chamfer	СН	.025	.035	.045	0.64	0.89	1.14
Mold Draft Angle Top	α	5	10	15	5	10	15
Mold Draft Angle Bottom	β	5	10	15	5	10	15

* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D1 and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-026 Drawing No. C04-076





	Units		INCHES*		MILLIMETERS			
Dimensior	1 Limits	MIN	NOM	MAX	MIN	NOM	MAX	
Number of Pins	n		44			44		
Pitch	р		.050			1.27		
Pins per Side	n1		11			11		
Overall Height	А	.165	.173	.180	4.19	4.39	4.57	
Molded Package Thickness	A2	.145	.153	.160	3.68	3.87	4.06	
Standoff §	A1	.020	.028	.035	0.51	0.71	0.89	
Side 1 Chamfer Height	A3	.024	.029	.034	0.61	0.74	0.86	
Corner Chamfer 1	CH1	.040	.045	.050	1.02	1.14	1.27	
Corner Chamfer (others)	CH2	.000	.005	.010	0.00	0.13	0.25	
Overall Width	E	.685	.690	.695	17.40	17.53	17.65	
Overall Length	D	.685	.690	.695	17.40	17.53	17.65	
Molded Package Width	E1	.650	.653	.656	16.51	16.59	16.66	
Molded Package Length	D1	.650	.653	.656	16.51	16.59	16.66	
Footprint Width	E2	.590	.620	.630	14.99	15.75	16.00	
Footprint Length	D2	.590	.620	.630	14.99	15.75	16.00	
Lead Thickness	С	.008	.011	.013	0.20	0.27	0.33	
Upper Lead Width	B1	.026	.029	.032	0.66	0.74	0.81	
Lower Lead Width	В	.013	.020	.021	0.33	0.51	0.53	
Mold Draft Angle Top	α	0	5	10	0	5	10	
Mold Draft Angle Bottom	β	0	5	10	0	5	10	

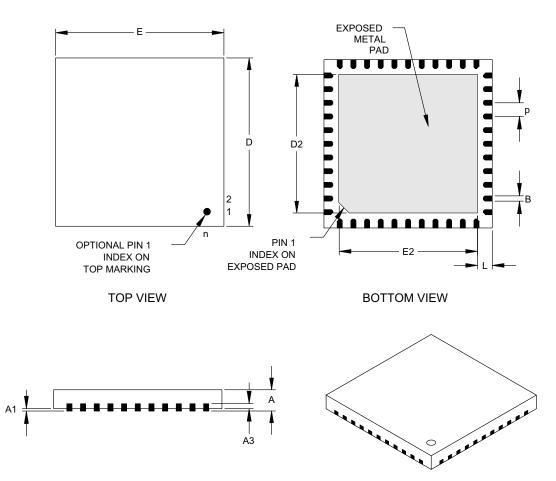
* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MO-047

Drawing No. C04-048

44-Lead Plastic Quad Flat No Lead Package (ML) 8x8 mm Body (QFN)



	Units	Units INCHES			MILLIMETERS*		
Din	nension Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		44			44	
Pitch	р	.026 BSC			0.65 BSC		
Overall Height	A	.031	.035	.039	0.80	0.90	1.00
Standoff	A1	.000	.001	.002	0	0.02	0.05
Base Thickness	A3	.010 REF			0.25 REF		
Overall Width	E	.315 BSC			8.00 BSC		
Exposed Pad Width	E2	.262	.268	.274	6.65	6.80	6.95
Overall Length	D	.315 BSC			8.00 BSC		
Exposed Pad Length	D2	.262	.268	.274	6.65	6.80	6.95
Lead Width	В	.012	.013	.013	0.30	0.33	0.35
Lead Length	L	.014	.016	.018	0.35	0.40	0.45

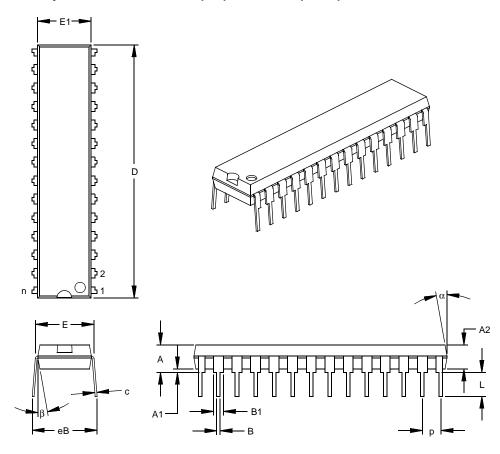
*Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC equivalent: M0-220

Drawing No. C04-103

28-Lead Skinny Plastic Dual In-line (SP) – 300 mil (PDIP)

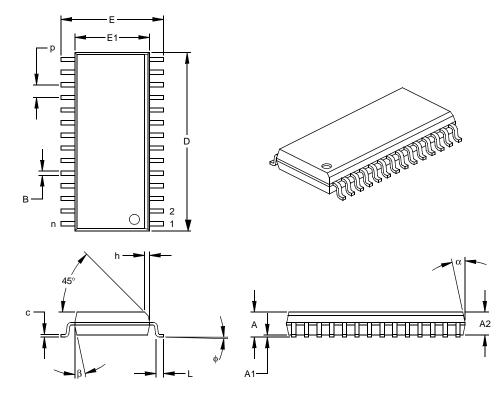


	Un	its	INCHES*			MILLIMETERS		
Dimen	sion Limits	;	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n			28			28	
Pitch	р			.100			2.54	
Top to Seating Plane	A		.140	.150	.160	3.56	3.81	4.06
Molded Package Thickness	A	2	.125	.130	.135	3.18	3.30	3.43
Base to Seating Plane	A	1	.015			0.38		
Shoulder to Shoulder Width	E		.300	.310	.325	7.62	7.87	8.26
Molded Package Width	E	1	.275	.285	.295	6.99	7.24	7.49
Overall Length	D		1.345	1.365	1.385	34.16	34.67	35.18
Tip to Seating Plane	L		.125	.130	.135	3.18	3.30	3.43
Lead Thickness	С		.008	.012	.015	0.20	0.29	0.38
Upper Lead Width	B	1	.040	.053	.065	1.02	1.33	1.65
Lower Lead Width	В		.016	.019	.022	0.41	0.48	0.56
Overall Row Spacing	§ eE	3	.320	.350	.430	8.13	8.89	10.92
Mold Draft Angle Top	α		5	10	15	5	10	15
Mold Draft Angle Bottom	β		5	10	15	5	10	15

* Controlling Parameter § Significant Characteristic Notes:

Dimension D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MO-095 Drawing No. C04-070

28-Lead Plastic Small Outline (SO) - Wide, 300 mil (SOIC)



	Units	nits INCHES*			MILLIMETERS		
Dimension	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		28			28	
Pitch	р		.050			1.27	
Overall Height	Α	.093	.099	.104	2.36	2.50	2.64
Molded Package Thickness	A2	.088	.091	.094	2.24	2.31	2.39
Standoff §	A1	.004	.008	.012	0.10	0.20	0.30
Overall Width	Е	.394	.407	.420	10.01	10.34	10.67
Molded Package Width	E1	.288	.295	.299	7.32	7.49	7.59
Overall Length	D	.695	.704	.712	17.65	17.87	18.08
Chamfer Distance	h	.010	.020	.029	0.25	0.50	0.74
Foot Length	L	.016	.033	.050	0.41	0.84	1.27
Foot Angle Top	¢	0	4	8	0	4	8
Lead Thickness	С	.009	.011	.013	0.23	0.28	0.33
Lead Width	В	.014	.017	.020	0.36	0.42	0.51
Mold Draft Angle Top	α	0	12	15	0	12	15
Mold Draft Angle Bottom	β	0	12	15	0	12	15

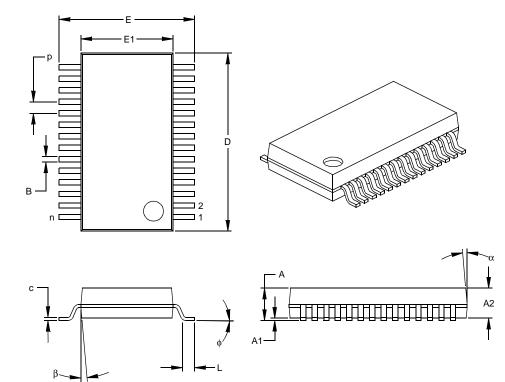
* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side.

JEDEC Equivalent: MS-013 Drawing No. C04-052

28-Lead Plastic Shrink Small Outline (SS) - 209 mil, 5.30 mm (SSOP)



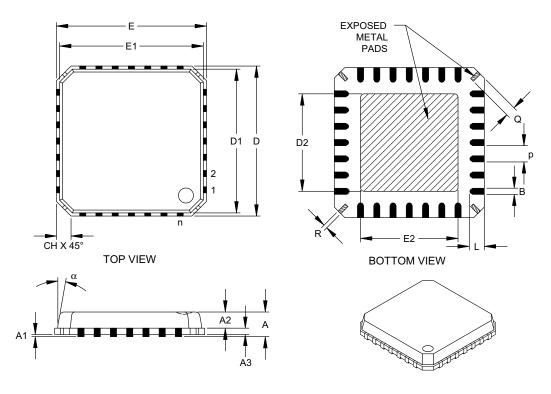
	Units	INCHES			MILLIMETERS*		
Dimensio	n Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		28			28	
Pitch	р		.026			0.65	
Overall Height	Α	.068	.073	.078	1.73	1.85	1.98
Molded Package Thickness	A2	.064	.068	.072	1.63	1.73	1.83
Standoff §	A1	.002	.006	.010	0.05	0.15	0.25
Overall Width	Е	.299	.309	.319	7.59	7.85	8.10
Molded Package Width	E1	.201	.207	.212	5.11	5.25	5.38
Overall Length	D	.396	.402	.407	10.06	10.20	10.34
Foot Length	L	.022	.030	.037	0.56	0.75	0.94
Lead Thickness	С	.004	.007	.010	0.10	0.18	0.25
Foot Angle	φ	0	4	8	0.00	101.60	203.20
Lead Width	В	.010	.013	.015	0.25	0.32	0.38
Mold Draft Angle Top	α	0	5	10	0	5	10
Mold Draft Angle Bottom	β	0	5	10	0	5	10

* Controlling Parameter § Significant Characteristic

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC Equivalent: MS-150 Drawing No. C04-073

28-Lead Plastic Quad Flat No Lead Package (ML) 6x6 mm Body, Punch Singulated (QFN)



	Units	INCHES			MILLIMETERS*		
Dimensior	Limits	MIN	NOM	MAX	MIN	NOM	MAX
Number of Pins	n		28			28	
Pitch	р		.026 BSC			0.65 BSC	
Overall Height	A		.033	.039		0.85	1.00
Molded Package Thickness	A2		.026	.031		0.65	0.80
Standoff	A1	.000	.0004	.002	0.00	0.01	0.05
Base Thickness	A3	.008 REF			0.20 REF		
Overall Width	E		.236 BSC			6.00 BSC	
Molded Package Width	E1		.226 BSC			5.75 BSC	
Exposed Pad Width	E2	.140	.146	.152	3.55	3.70	3.85
Overall Length	D		.236 BSC			6.00 BSC	
Molded Package Length	D1		.226 BSC			5.75 BSC	
Exposed Pad Length	D2	.140	.146	.152	3.55	3.70	3.85
Lead Width	В	.009	.011	.014	0.23	0.28	0.35
Lead Length	L	.020	.024	.030	0.50	0.60	0.75
Tie Bar Width	R	.005	.007	.010	0.13	0.17	0.23
Tie Bar Length	Q	.012	.016	.026	0.30	0.40	0.65
Chamfer	СН	.009	.017	.024	0.24	0.42	0.60
Mold Draft Angle Top	α			12°			12°

*Controlling Parameter

Notes:

Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed .010" (0.254mm) per side. JEDEC equivalent: mMO-220

Drawing No. C04-114

APPENDIX A: REVISION HISTORY

Revision A (November 2001)

Original data sheet for PIC16F87XA devices. The devices presented are enhanced versions of the PIC16F87X microcontrollers discussed in the *"PIC16F87X Data Sheet"* (DS30292).

Revision B (October 2003)

This revision includes the DC and AC Characteristics Graphs and Tables. The Electrical Specifications in **Section 17.0 "Electrical Characteristics"** have been updated and there have been minor corrections to the data sheet text.

APPENDIX B: DEVICE DIFFERENCES

The differences between the devices in this data sheet are listed in Table B-1.

	PIC16F873A	PIC16F874A	PIC16F876A	PIC16F877A
Flash Program Memory (14-bit words)	4K	4K	8K	8K
Data Memory (bytes)	192	192	368	368
EEPROM Data Memory (bytes)	128	128	256	256
Interrupts	14	15	14	15
I/O Ports	Ports A, B, C	Ports A, B, C, D, E	Ports A, B, C	Ports A, B, C, D, E
Serial Communications	MSSP, USART	MSSP, USART	MSSP, USART	MSSP, USART
Parallel Slave Port	No	Yes	No	Yes
10-bit Analog-to-Digital Module	5 input channels	8 input channels	5 input channels	8 input channels
Packages	28-pin PDIP 28-pin SOIC 28-pin SSOP 28-pin QFN	40-pin PDIP 44-pin PLCC 44-pin TQFP 44-pin QFN	28-pin PDIP 28-pin SOIC 28-pin SSOP 28-pin QFN	40-pin PDIP 44-pin PLCC 44-pin TQFP 44-pin QFN

TABLE B-1: DIFFERENCES BETWEEN DEVICES IN THE PIC16F87XA FAMILY

APPENDIX C: CONVERSION CONSIDERATIONS

Considerations for converting from previous versions of devices to the ones listed in this data sheet are listed in Table C-1.

TABLE C-1: CONVERSION CONSIDERATION

Characteristic	PIC16C7X	PIC16F87X	PIC16F87XA
Pins	28/40	28/40	28/40
Timers	3	3	3
Interrupts	11 or 12	13 or 14	14 or 15
Communication	PSP, USART, SSP (SPI, I ² C Slave)	PSP, USART, SSP (SPI, I ² C Master/Slave)	PSP, USART, SSP (SPI, I ² C Master/Slave)
Frequency	20 MHz	20 MHz	20 MHz
Voltage	2.5V-5.5V	2.2V-5.5V	2.0V-5.5V
A/D	8-bit, 4 conversion clock selects	10-bit, 4 conversion clock selects	10-bit, 7 conversion clock selects
CCP	2	2	2
Comparator	—	—	2
Comparator Voltage Reference	_	_	Yes
Program Memory	4K, 8K EPROM	4K, 8K Flash (Erase/Write on single-word)	4K, 8K Flash (Erase/Write on four-word blocks)
RAM	192, 368 bytes	192, 368 bytes	192, 368 bytes
EEPROM Data	None	128, 256 bytes	128, 256 bytes
Code Protection	On/Off	Segmented, starting at end of program memory	On/Off
Program Memory Write Protection	_	On/Off	Segmented, starting at beginning of program memory
Other	—	In-Circuit Debugger, Low-Voltage Programming	In-Circuit Debugger, Low-Voltage Programming

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Device	PIC16F87XA ⁽¹⁾ , PIC16F87XAT ⁽²⁾ ; VDD range 4.0V to 5.5V PIC16LF87XA ⁽¹⁾ , PIC16LF87XAT ⁽²⁾ ; VDD range 2.0V to 5.5V	 c) PIC16F877A-I/P = Industrial temp., PDIP package, 10 MHz, normal VDD limits.
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